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## MAX77646/MAX77647

## Ultra-Configurable SIMO PMIC Featuring 3-Output Buck-Boost, 1-LDO for Long Battery Life Primary Cell Applications

### General Description

The MAX77646/MAX77647 provides power supply solutions for low-power applications where size and efficiency are critical. The IC features a Single Inductor Multiple Output (SIMO) buck-boost regulator that provides three independently programmable power rails from a single inductor to minimize total solution size. A 150mA LDO provides ripple rejection for audio and other noise-sensitive applications. The LDO can also be configured as a load switch to manage power consumption by disconnecting external blocks when not required.

The MAX77646/MAX77647 operates with battery voltages from 1.8V with coin-cell, dual silver oxide, dual alkaline, and LiSOCL<sub>2</sub> primary batteries.

The MAX77646 (R<sub>SEL</sub> variant) SIMO and LDO output voltages are individually programmable through resistors. A peak current limit input is used to set both the inductor's peak current limits of the device with a single resistor. Individual enable pins combined with the flexible resistor programmability allows the device to be tailored for many applications.

The MAX77647 (I<sup>2</sup>C variant) SIMO and LDO output voltages are individually programmable through I<sup>2</sup>C and in addition, includes two GPIOs with alternate modes for scalability. A bidirectional I<sup>2</sup>C serial interface allows for configuring and checking the status of the devices. An internal on/off controller provides a controlled startup sequence for the regulators and provides supervisory functionality while they are on. Numerous factory programmable options allow the device to be tailored for many applications, enabling faster time to market.

### Applications

- GNSS Module
- IoT Enabled Wireless Sensor Nodes
- Home Safety and Security Monitors
- Factory Safety and Security Monitors
- Fitness, Health, and Activity Trackers

### Benefits and Features

- 1.8V to 3.3V Input Range
- Highly Integrated
  - 3x Output, Single Inductor Multiple Output (SIMO) Buck-Boost Regulators
    - Supports Wide Output Voltage Range from 0.5V to 5.1V for all SIMO Channels
  - 1x 150mA LDO
    - 100mA in LSW Mode
  - 2x GPIOs (MAX77647)
  - Watchdog Timer (MAX77647)
- Ultra Low-Power SIMO
  - 1.3μA Operating Current (3x SIMO Channels)
  - 0.1μA Shutdown Current
  - 85% Efficiency in Buck-Only Mode ( $V_{IN} = 2.4V$ ,  $V_{OUT} = 0.8V$ , 10mA,  $I_{PK} = 0.5A$ )
  - 86% Efficiency in Buck-Boost Mode ( $V_{IN} = 2.4V$ ,  $V_{OUT} = 1.8V$ , 10mA,  $I_{PK} = 0.5A$ )
  - Less Than 20mVpp Output Ripple at  $V_{OUT} = 1.8V$
  - Automatic Low-Power Mode to Normal-Power Mode Transition
- Flexible and Configurable
  - Ultra-Configurable Resistor Programmable Output Voltages (MAX77646)
  - I<sup>2</sup>C-Programmable Output Voltages (MAX77647)
- Small Size
  - 5mm<sup>2</sup> Wafer-Level Package (WLP)
  - 20-Bump, 0.5mm Pitch, 5 x 4 Array

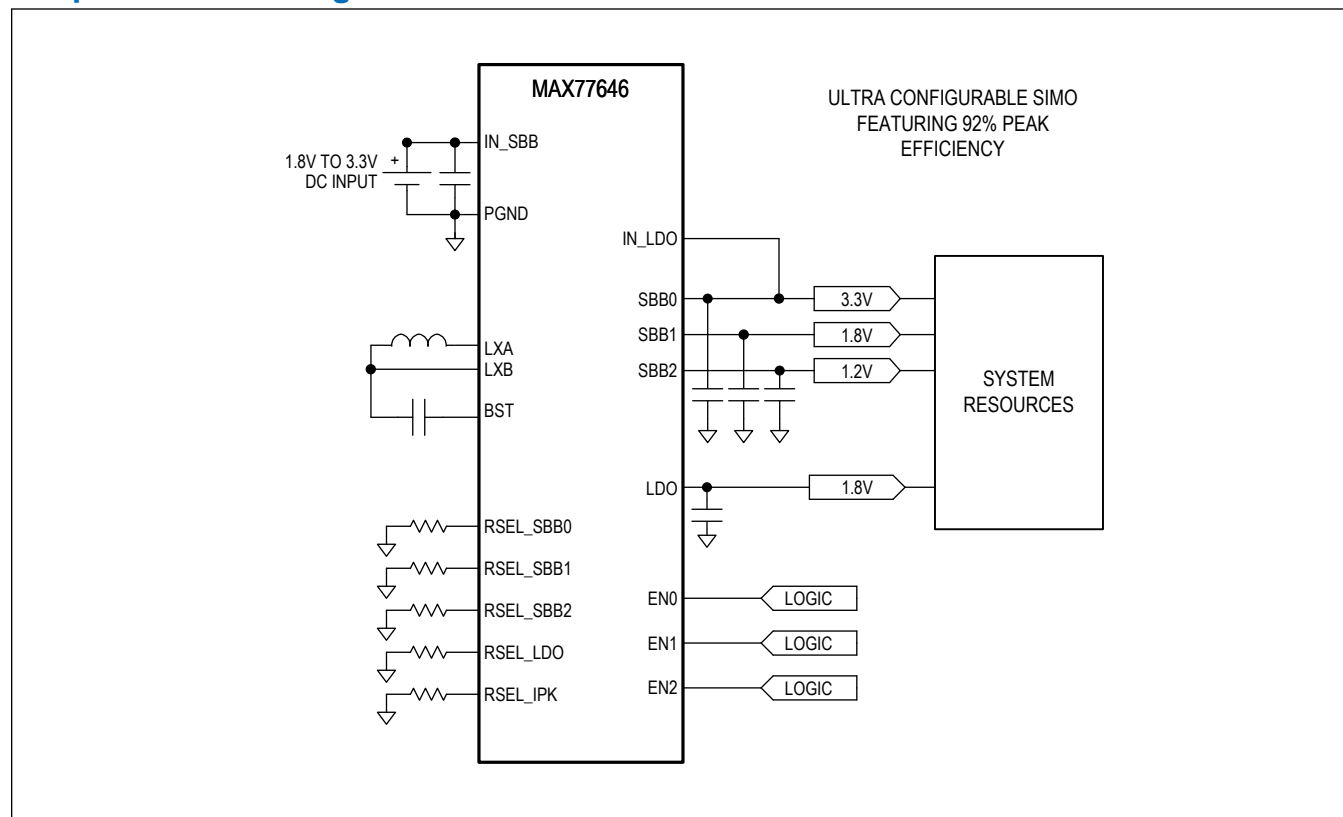
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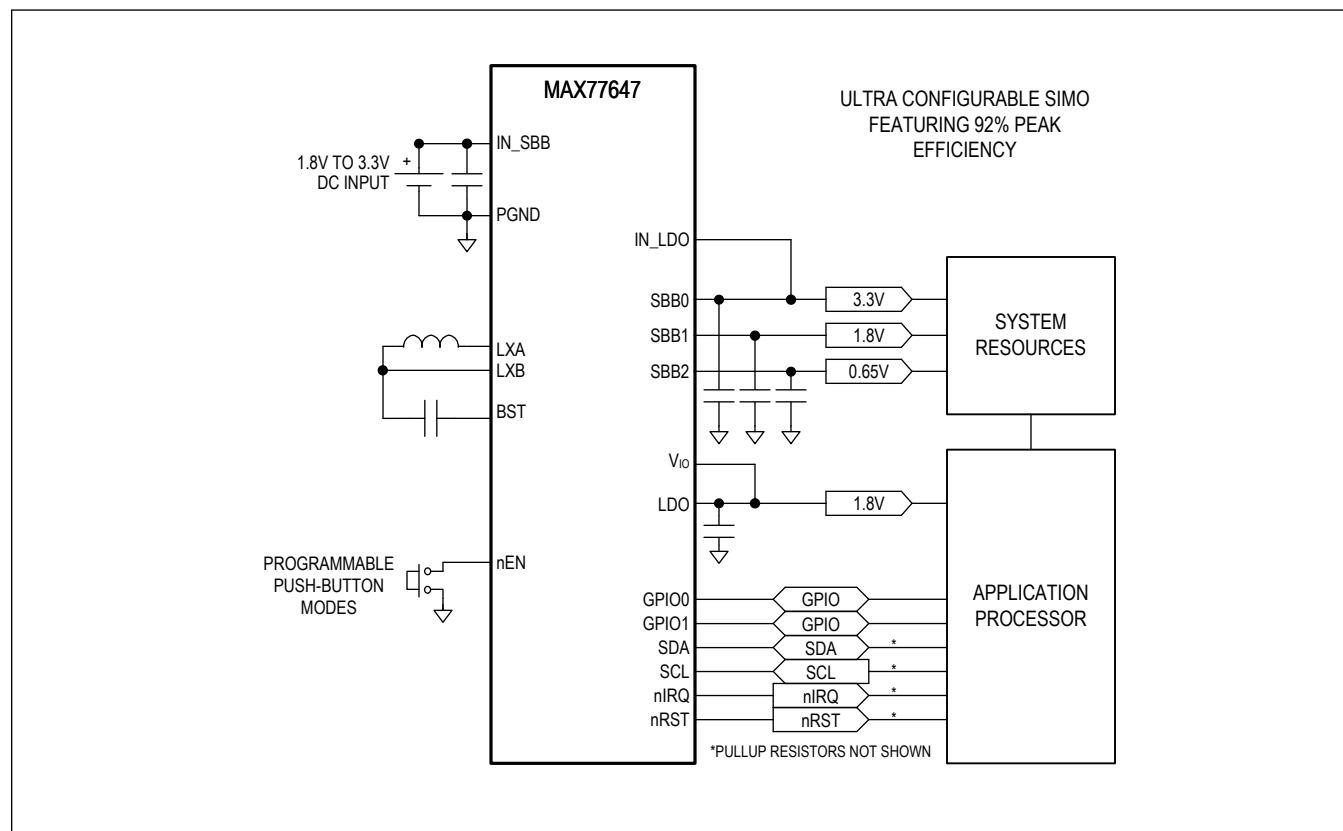
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## Simplified Block Diagrams





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## Absolute Maximum Ratings

nEN, nIRQ, nRST to GND .....	-0.3V to $V_{IN\_SBB} + 0.3V$	BST to LXB .....	-0.3V to +6.0V
$V_{IO}$ to GND .....	-0.3V to $V_{IN\_SBB} + 0.3V$	LXA Continuous Current (Note 2) .....	1.2A <sub>RMS</sub>
SCL, SDA to GND .....	-0.3V to +6.0V	LXB Continuous Current (Note 2) .....	1.2A <sub>RMS</sub>
GPIO0, GPIO1 to GND .....	-0.3V to +6.0V	IN_LDO to PGND .....	-0.3V to +6.0V
nIRQ, nRST, SDA, GPIO Continuous Current .....	±20mA	LDO to PGND .....	-0.3V to $V_{IN\_LDO} + 0.3V$
RSEL_SBB0, RSEL_SBB1, RSEL_SBB2, RSEL_LDO, RSEL_IPK, EN0, EN1, EN2 to GND .....	-0.3V to $V_{IN\_SBB} + 0.3V$	PGND to GND .....	-0.3V to +0.3V
IN_SBB to PGND .....	-0.3V to +6.0V	Operating Temperature Range .....	-40°C to +125°C
IN_SBB Continuous Current (Note 1) .....	1.2A <sub>RMS</sub>	Junction Temperature .....	+150°C
SBB0, SBB1, SBB2 to PGND .....	-0.3V to +6.0V	Storage Temperature Range .....	-65°C to +150°C
SBB0, SBB1, SBB2 Short-Circuit Duration .....	Continuous	Soldering Temperature (reflow) .....	+260°C
BST to IN_SBB .....	-0.3V to +6.0V	Continuous Power Dissipation (Multilayer Board, $T_A = +70^\circ\text{C}$ , derate 20.4mW/°C above +70°C) .....	1632mW

**Note 1:** Do not repeatedly hot-plug a source to the IN terminals (IN\_SBB or IN\_LDO) at a rate greater than 10Hz. Hot plugging low impedance sources results in an ~8A momentary (~2μs) current spike.

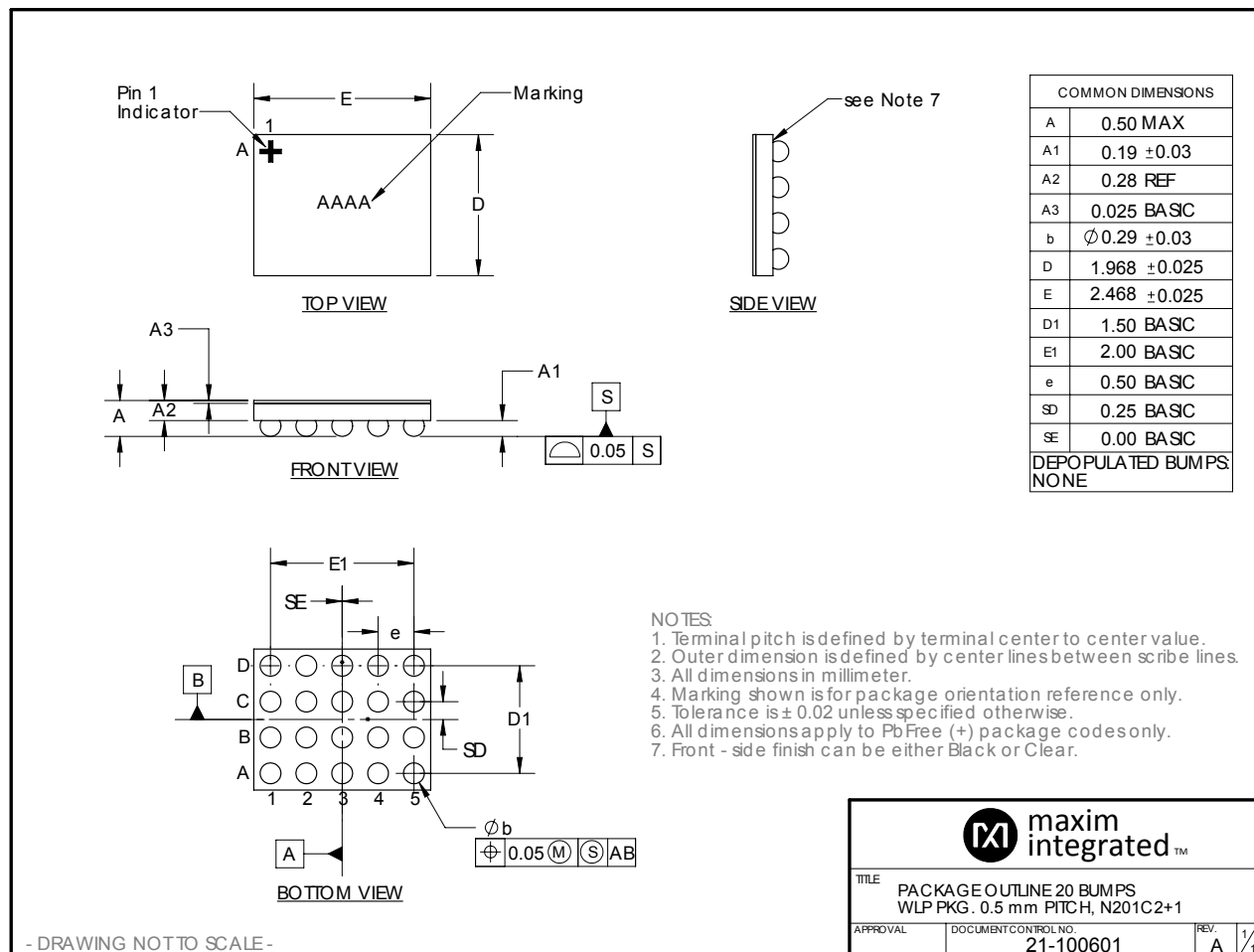
**Note 2:** Do not externally bias LXA or LXB. LXA has internal clamping diodes to PGND and IN\_SBB. LXB has an internal low-side clamping diode to PGND and an internal high-side clamping diode that dynamically connects to a selected SIMO output. It is normal for these diodes to briefly conduct during switching events. When the SIMO regulator is disabled, the LXB to PGND absolute maximum voltage is -0.3V to  $V_{SBB0} + 0.3V$ .

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

## Package Information

### WLP

Package Code	N201C2+1
Outline Number	<a href="#">21-100601</a>
Land Pattern Number	Refer to <a href="#">Application Note 1891</a>
<b>Thermal Resistance, Four-Layer Board:</b>	
Junction to Ambient ( $\theta_{JA}$ )	46.84



For the latest package outline information and land patterns (footprints), go to [www.maximintegrated.com/packages](http://www.maximintegrated.com/packages). Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a four-layer board. For detailed information on package thermal considerations, refer to [www.maximintegrated.com/thermal-tutorial](http://www.maximintegrated.com/thermal-tutorial).

## Electrical Characteristics

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ ,  $V_{IO} = 1.8V$ , limits are 100% production tested at  $T_A = +25^\circ C$ . Limits over the operating temperature range ( $T_A = -20^\circ C$  to  $+85^\circ C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Operating Voltage Range	$V_{IN}$		1.8		3.3	V
Shutdown Supply Current	$I_{SHDN}$	Current measured into $IN\_SBB$ ; $IN\_LDO$ connected to $IN\_SBB$ ; all resources OFF (LDO, SBB0, SBB1, SBB2); $T_A = +25^\circ C$		0.1	0.5	$\mu A$

**Electrical Characteristics (continued)**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ ,  $V_{IO} = 1.8V$ , limits are 100% production tested at  $T_A = +25^{\circ}C$ . Limits over the operating temperature range ( $T_A = -20^{\circ}C$  to  $+85^{\circ}C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Quiescent Supply Current	$I_Q$	Current measured into IN_SBB; IN_LDO connected to IN_SBB; all resources ON (LDO, SBB0, SBB1, SBB2) at no load; watchdog timer disabled	Main bias is in low-power mode (CNFG_GLBL0.SBI A_LPM = 1); SIMO low power enabled		3	6	$\mu A$
Quiescent SIMO Supply Current	$I_{Q-SIMO}$	Current measured into IN_SBB; IN_LDO disconnected from IN_SBB; all SIMO resources ON (SBB0, SBB1, SBB2) at no load; LDO OFF; watchdog timer disabled	Main bias in low-power mode (CNFG_GLBL0.SBI A_LPM = 1); SIMO low power enabled		1.3	3.5	$\mu A$
SIMO Supply Current	$I_{IN-SIMO\_SBB0}$	Current measured into IN_SBB, IN_LDO disconnected, SBB0 ON at no load, SBB1/2 OFF	Main bias ON in Low-power mode (CNFG_GLBL0.SBI A_LPM = 0); SIMO low power disabled (MAX77646)		30	50	$\mu A$
			Main bias ON in normal-power mode (CNFG_GLBL0.SBI A_LPM = 0); SIMO low power disabled (MAX77647)		45	75	

**Electrical Characteristics—Global Resources**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ , limits are 100% production tested at  $T_A = +25^{\circ}C$ , limits over the operating temperature range ( $T_A = -20^{\circ}C$  to  $+85^{\circ}C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>GENERAL CHARACTERISTICS</b>						
Main Bias Enable Time	$t_{SBIAS\_EN}$			1		ms
<b>VOLTAGE MONITORS / POWER-ON RESET (POR)</b>						
POR Threshold	$V_{POR}$	$V_{IN\_SBB}$ falling		1.316		V
POR Threshold Hysteresis				40		mV
<b>VOLTAGE MONITORS / UNDERVOLTAGE LOCKOUT (UVLO)</b>						
UVLO Threshold	$V_{IN\_SBB-UVLO}$	$V_{IN\_SBB}$ falling	1.46	1.6	1.71	V

**Electrical Characteristics—Global Resources (continued)**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ , limits are 100% production tested at  $T_A = +25^{\circ}C$ , limits over the operating temperature range ( $T_A = -20^{\circ}C$  to  $+85^{\circ}C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
UVLO Threshold Hysteresis	V <sub>INSBB-UVLO-HYS</sub>				150		mV
VOLTAGE MONITORS / OVERVOLTAGE LOCKOUT (OVLO)							
OVLO Threshold	V <sub>INSBB-OVLO</sub>	V <sub>INSBB-OVLO</sub> rising		5.65	5.85	6.00	V
THERMAL MONITORS							
Overtemperature-Lockout Threshold	T <sub>OTLO</sub>	T <sub>J</sub> rising			145		°C
Thermal Alarm Temperature 1	T <sub>JAL1</sub>	T <sub>J</sub> rising			80		°C
Thermal Alarm Temperature 2	T <sub>JAL2</sub>	T <sub>J</sub> rising			100		°C
Thermal Alarm Temperature Hysteresis					15		°C
ENABLE INPUT (nEN, ENx)							
nEN Internal Pullup	R <sub>nEN-PU</sub>	Pullup to V <sub>IN_SBB</sub>	CNFG_GLBL0.PU_DIS = 0	200			kΩ
			CNFG_GLBL0.PU_DIS = 1	10000			
EN Internal Pulldown	R <sub>EN-PD</sub>	Pulldown to GND (MAX77646)		50			nA
nEN Input Leakage Current	I <sub>nEN_LKG</sub>	V <sub>nEN</sub> = V <sub>IN_SBB</sub> = V <sub>IN_LDO</sub> = 3.3V	T <sub>A</sub> = +25°C	-1	±0.001	+1	μA
		V <sub>nEN</sub> = V <sub>IN_SBB</sub> = V <sub>IN_LDO</sub> = 3.3V	T <sub>A</sub> = +85°C	±0.01			
nEN Input Falling Threshold	V <sub>TH_nEN_F</sub>	nEN falling		0.4			V
nEN Input Rising Threshold	V <sub>TH_nEN_R</sub>	nEN rising		1.4			V
EN Input Threshold, High	V <sub>IH</sub>	Voltage threshold, rising (MAX77646)		1.4			V
EN Input Threshold, Low	V <sub>IL</sub>	Voltage threshold, falling (MAX77646)		0.4			V
Debounce Time	t <sub>DBNC_nEN</sub>	CNFG_GLBL0.DBEN_nEN = 0		1			ms
		CNFG_GLBL0.DBEN_nEN = 1		30			
Manual Reset Time	t <sub>MRST</sub>	CNFG_GLBL0.T_MRST = 1		3	4	5	s
		CNFG_GLBL0.T_MRST = 0		7	8	10.5	
OPEN-DRAIN INTERRUPT OUTPUT (nIRQ)							
Output Voltage Low	V <sub>OL</sub>	I <sub>SINK</sub> = 2mA		0.4			V
Output Falling Edge Time	t <sub>f_nIRQ</sub>	C <sub>IRQ</sub> = 25pF		2			ns

**Electrical Characteristics—Global Resources (continued)**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ , limits are 100% production tested at  $T_A = +25^\circ C$ , limits over the operating temperature range ( $T_A = -20^\circ C$  to  $+85^\circ C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
Leakage Current	I <sub>nIRQ_LKG</sub>	V <sub>IN_SBB</sub> = V <sub>IO</sub> = 3.3V nIRQ is high impedance (no interrupts) V <sub>nIRQ</sub> = 0V and 3.3V	T <sub>A</sub> = +25°C	-1	±0.001	+1	µA
			T <sub>A</sub> = +85°C	±0.01			
OPEN-DRAIN RESET OUTPUT (nRST)							
Output Voltage Low	V <sub>OL</sub>	I <sub>SINK</sub> = 2mA		0.4			V
Output Falling Edge Time	t <sub>f_nRST</sub>	C <sub>RST</sub> = 25pF		2			ns
nRST Deassert Delay Time	t <sub>RSTODD</sub>			5.12			ms
nRST Assert Delay Time	t <sub>RSTOAD</sub>			10.24			ms
Leakage Current	I <sub>nRST_LKG</sub>	V <sub>IN_SBB</sub> = V <sub>IO</sub> = 3.3V nRST is high impedance (no reset) V <sub>nRST</sub> = 0V and 3.3V	T <sub>A</sub> = +25°C	-1	±0.001	+1	µA
		V <sub>IN_SBB</sub> = V <sub>IO</sub> = 3.3V nRST is high impedance (no reset) V <sub>nRST</sub> = 0V and 3.3V	T <sub>A</sub> = +85°C	±0.01			
GENERAL PURPOSE INPUT/OUTPUT (GPIO)							
GPIO Supply Voltage Range				V <sub>IO</sub>			V
Input Voltage Low	V <sub>IL</sub>	V <sub>IO</sub> = 1.8V		0.3 x V <sub>IO</sub>			V
Input Voltage High	V <sub>IH</sub>	V <sub>IO</sub> = 1.8V		0.7 x V <sub>IO</sub>			V
Input Leakage Current	I <sub>GPI_LKG</sub>	CNFG_GPIOx.DIR = 1 V <sub>IO</sub> = 3.3V V <sub>GPIO</sub> = 0V and 3.3V	T <sub>A</sub> = +25°C	-1	±0.001	+1	µA
			T <sub>A</sub> = +85°C	±0.01			
Output Voltage Low	V <sub>OL</sub>	I <sub>SINK</sub> = 2mA		0.4			V
Output Voltage High	V <sub>OH</sub>	I <sub>SOURCE</sub> = 1mA		0.8 x V <sub>IO</sub>			V
Input Debounce Time	t <sub>DBNC_GPI</sub>	CNFG_GPIOx.DBEN_GPI = 1		30			ms
Output Falling Edge Time	t <sub>f_GPIO</sub>	C <sub>GPIO</sub> = 25pF		3			ns
Output Rising Edge Time	t <sub>r_GPIO</sub>	C <sub>GPIO</sub> = 25pF		3			ns

**Electrical Characteristics—Global Resources (continued)**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ , limits are 100% production tested at  $T_A = +25^{\circ}C$ , limits over the operating temperature range ( $T_A = -20^{\circ}C$  to  $+85^{\circ}C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
FLEXIBLE POWER SEQUENCER							
FPS Startup Delay	t <sub>FPS_DLY</sub>			1.43			ms
Power-Up Event Periods	t <sub>EN</sub>	See <a href="#">Figure 10</a>		1.28			ms
Power-Down Event Periods	t <sub>DIS</sub>	See <a href="#">Figure 10</a>		2.56			ms
R <sub>SEL</sub> (MAX77646)							
Select Resistor Detection Time	t <sub>RSEL</sub>	Total time to detect all five resistors	V <sub>IN_SBB</sub> = 2.4V, C <sub>RSEL</sub> < 2pF	4			ms
Required Select Resistor Accuracy	R <sub>SEL_TOL</sub>	Use the nearest ±1% resistor from the <a href="#">R<sub>SEL</sub> Selection Table</a>		±1			%

**Note 3:** Programmed at Analog Devices factory.

**Electrical Characteristics—SIMO Buck-Boost**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ ,  $C_{SBBx} = 10\mu F$ ,  $L = 1.5\mu H$ , limits are 100% production tested at  $T_A = +25^{\circ}C$ , limits over the operating temperature range ( $T_A = -20^{\circ}C$  to  $+85^{\circ}C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
GENERAL CHARACTERISTICS / OUTPUT VOLTAGE RANGE (SBB0/1/2)							
Programmable Output Voltage Range				0.5		5.1	V
Output DAC Bits					8		bits
Output DAC LSB Size		0.5V to 5.1V			25		mV
OUTPUT VOLTAGE ACCURACY							
Output Voltage Accuracy		V <sub>SBBx</sub> falling, threshold where LXA switches high; specified as a percentage of target output voltage	T <sub>A</sub> = -40°C to +85°C	-2.0		+2.0	%
OUT Over-Regulation Threshold	V <sub>OV</sub>	T <sub>A</sub> = +25°C			1.7	3.3	%
TIMING CHARACTERISTICS							
Enable Delay		Delay time from the SIMO receiving its first enable signal to when it begins to switch in order to service that output			10		µs
Soft-Start Slew Rate	dV/dt <sub>SS</sub>	I <sub>PK</sub> = 1A, C <sub>OUT</sub> = 10µF			5.0		mV/µs
POWER STAGE CHARACTERISTICS							
LXA Leakage Current		SBB0, SBB1, SBB2 are disabled, V <sub>IN_SBB</sub> = 3.3V, V <sub>LXA</sub> = 0V, or 3.3V	T <sub>A</sub> = +25°C	-1.0	±0.1	+1.0	µA
			T <sub>A</sub> = +85°C		±1.0		

**Electrical Characteristics—SIMO Buck-Boost (continued)**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4\text{ V}$ ,  $C_{SBBx} = 10\mu\text{F}$ ,  $L = 1.5\mu\text{H}$ , limits are 100% production tested at  $T_A = +25^\circ\text{C}$ , limits over the operating temperature range ( $T_A = -20^\circ\text{C}$  to  $+85^\circ\text{C}$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS		MIN	TYP	MAX	UNITS
LXB Leakage Current		SBB0, SBB1, SBB2 are disabled, $V_{IN\_SBB} = 3.3V$ , $V_{LXA} = 0V$ or $3.3V$ , all $V_{SBBx} = 3.3V$	$T_A = +25^{\circ}C$	-1.0	$\pm 0.1$	+1.0	$\mu A$
			$T_A = +85^{\circ}C$	$\pm 1.0$			
BST Leakage Current		$V_{IN\_SBB} = 3.3V$ , $V_{LXB} = 3.3V$ , $V_{BST} = 9.3V$	$T_A = +25^{\circ}C$		+0.01	+1.0	$\mu A$
			$T_A = +85^{\circ}C$	+0.1			
Disabled Output Leakage Current		SBB0, SBB1, SBB2 are disabled, active-discharge disabled, $V_{SBBx} = 3.3V$ , $V_{LXB} = 0V$ , $V_{IN\_SBB} = V_{BST} = 3.3V$	$T_A = +25^{\circ}C$		+0.1	+1.0	$\mu A$
			$T_A = +85^{\circ}C$	+0.2			
Active Discharge Resistance	$R_{AD\_SBBx}$	SBB0, SBB1, SBB2 are disabled, active-discharge enabled (CNFG_SBBx_B.ADE_SBBx = 1)		60	120	180	$\Omega$
CONTROL SCHEME							
Peak Current Limit	$I_{P\_SBB}$ (Note 4)			-8%	0.500	+8%	A
				-8%	0.700	+8%	
				-8%	0.910	+8%	

**Note 4:** Typical values align with bench observations using the stated conditions with an inductor. Minimum and maximum values are tested in production with DC currents without an inductor. See the *Typical Operating Characteristics* SIMO switching waveforms to gain more insight on this specification.

**Electrical Characteristics—Low Dropout Linear Regulator (LDO)/Load Switch (LSW)**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4\text{V}$ , limits are 100% production tested at  $T_A = +25^\circ\text{C}$ , limits over the operating temperature range ( $T_A = -20^\circ\text{C}$  to  $+85^\circ\text{C}$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
<b>LDO</b>						
Input Voltage Range	$V_{IN\_LDO}$	LDO mode	1.6		5.5	V
		Switch mode	1.1		5.5	
Quiescent Supply Current	$I_{IN\_LDO}$	$I_{OUT\_LDO} = 0$		1.4	2.4	$\mu\text{A}$
		$I_{OUT\_LDO} = 0$ , switch mode		0.5	1.2	
Quiescent Supply Current in Dropout	$I_{IN\_DRP\_LDO}$	$I_{OUT\_LDO} = 0$ , $V_{IN\_LDO} = 2.9\text{V}$ , $V_{LDO} = 3\text{V}$		2.1	4.6	$\mu\text{A}$
Maximum Output Current	$I_{OUT\_LDO}$	$V_{IN\_LDO} > 1.8\text{V}$	150			mA
		$V_{IN\_LDO} = 1.8\text{V}$ or lower	100			
Output Voltage (MAX77646)	$V_{OUT\_LDO}$		0.5		4	V
Output Voltage (MAX77647)	$V_{OUT\_LDO}$		0.5		5.0	V

**Electrical Characteristics—Low Dropout Linear Regulator (LDO)/Load Switch (LSW)  
(continued)**

( $V_{IN\_SBB} = V_{IN\_LDO} = 2.4V$ , limits are 100% production tested at  $T_A = +25^\circ C$ , limits over the operating temperature range ( $T_A = -20^\circ C$  to  $+85^\circ C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
Output Accuracy		$V_{IN\_LDO} = (V_{OUT\_LDO} + 0.5V)$ or higher, $I_{OUT\_LDO} = 1mA$	-3.1		+3.1	%
Dropout Voltage	$V_{DRP\_LDO}$	$V_{IN\_LDO} = 3V$ , LDO programmed to 3V, $I_{OUT\_LDO} = 100mA$			100	mV
Line Regulation		$V_{IN\_LDO} = (V_{OUT\_LDO} + 0.5V)$ to 5.5V	-0.5		+0.5	%/V
Load Regulation		$V_{IN\_LDO} = 1.8V$ or higher, $I_{OUT\_LDO} =$ 100 $\mu A$ to 100mA		0.001	0.005	%/mA
Line Transient		$V_{IN\_LDO} = 4V$ to 5V, 5 $\mu s$ rise time		$\pm 35$		mV
Load Transient		$I_{OUT\_LDO} = 100\mu A$ to 10mA, 200ns rise time		100		mV
		$I_{OUT\_LDO} = 100\mu A$ to 100mA, 200ns rise time		200		
Active-Discharge Resistance	$R_{AD\_LDO}$		42	80	200	$\Omega$
Switch Mode On- Resistance	$R_{ON\_LDO}$	$V_{IN\_LDO} = 2.7V$ , $I_{OUT\_LDO} = 100mA$			0.5	$\Omega$
		$V_{IN\_LDO} = 1.8V$ , $I_{OUT\_LDO} = 50mA$			0.8	
		$V_{IN\_LDO} = 1.2V$ , $I_{OUT\_LDO} = 5mA$			1.2	
Slew Rate		$I_{OUT\_LDO} = 0mA$ , time from 10% to 90% of final register value, $C_{OUT\_LDO} = 1\mu F$		2.2		V/ms
		$I_{OUT\_LDO} = 0mA$ , time from 10% to 90% of final register value, $C_{OUT\_LDO} = 1\mu F$ , switch mode		2.2		
Short-Circuit Current Limit		$V_{IN\_LDO} = 2.7V$ , $V_{OUT\_LDO} = GND$	190	550	910	mA
		$V_{IN\_LDO} = 2.7V$ , $V_{OUT\_LDO} = GND$ , switch mode	40	255		
Output Noise		10Hz to 100kHz, $V_{IN\_LDO} = 5V$ , $V_{OUT\_LDO} = 3.3V$		150		$\mu V_{RMS}$
		10Hz to 100kHz, $V_{IN\_LDO} = 5V$ , $V_{OUT\_LDO} = 2.5V$		125		
		10Hz to 100kHz, $V_{IN\_LDO} = 5V$ , $V_{OUT\_LDO} = 1.2V$		90		
		10Hz to 100kHz, $V_{IN\_LDO} = 5V$ , $V_{OUT\_LDO} = 0.9V$		80		
Output DAC Bits				8		bits
Output DAC LSB Size				25		mV



**Electrical Characteristics—I<sup>2</sup>C Serial Communication**

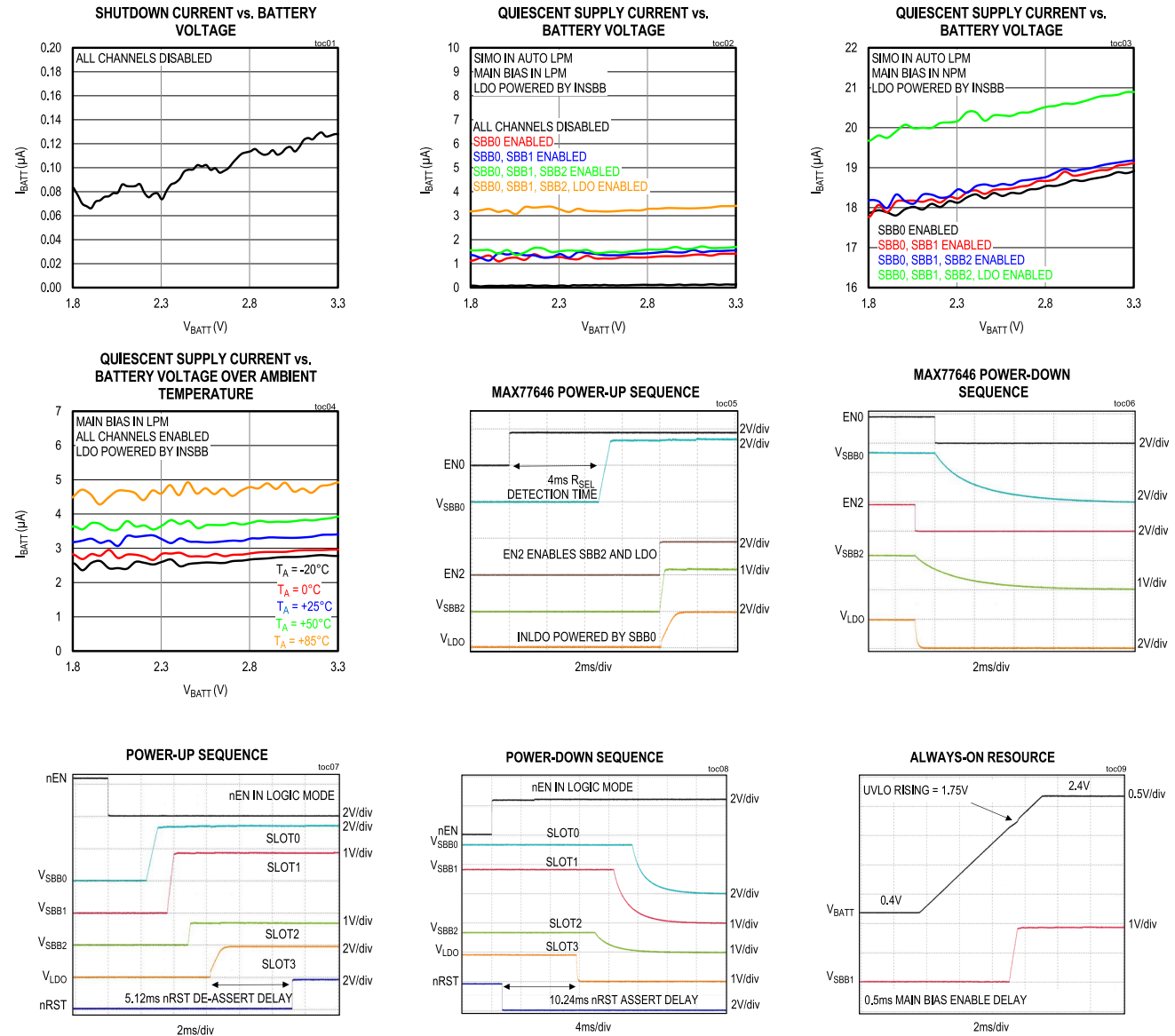
( $V_{IN} = 2.4V$ ,  $V_{IO} = 1.8V$ , limits are 100% production tested at  $T_A = +25^{\circ}C$ , limits over the operating temperature range ( $T_A = -20^{\circ}C$  to  $+85^{\circ}C$ ) are guaranteed by design and characterization, unless otherwise noted.)

PARAMETER	SYMBOL	CONDITIONS	MIN	TYP	MAX	UNITS
POWER SUPPLY						
V <sub>IO</sub> Voltage Range	V <sub>IO</sub>		1.7	1.8	3.6	V
V <sub>IO</sub> Bias Current		V <sub>IO</sub> = 3.6V, V <sub>SDA</sub> = V <sub>SCL</sub> = 0V or 3.6V, T <sub>A</sub> = +25°C	-1	0	+1	μA
		V <sub>IO</sub> = 1.7V, V <sub>SDA</sub> = V <sub>SCL</sub> = 0V or 1.7V	-1	0	+1	
SDA AND SCL I/O STAGE						
SCL, SDA Input High Voltage	V <sub>IH</sub>	V <sub>IO</sub> = 1.7V to 3.6V	0.7 x V <sub>IO</sub>			V
SCL, SDA Input Low Voltage	V <sub>IL</sub>	V <sub>IO</sub> = 1.7V to 3.6V	0.3 x V <sub>IO</sub>			V
SCL, SDA Input Hysteresis	V <sub>HYS</sub>		0.05 x V <sub>IO</sub>			V
SCL, SDA Input Leakage Current	I <sub>I</sub>	V <sub>IO</sub> = 3.6V, V <sub>SCL</sub> = V <sub>SDA</sub> = 0V and 3.6V	-10		+10	μA
SDA Output Low Voltage	V <sub>OL</sub>	Sinking 20mA	0.4			V
SCL, SDA Pin Capacitance	C <sub>I</sub>		10			pF
Max Output Fall Time from V <sub>IH</sub> to V <sub>IL</sub>	t <sub>OF</sub> (Note 5)		120			ns
I <sup>2</sup> C COMPATIBLE INTERFACE TIMING (STANDARD) (Note 5)						
Clock Frequency	f <sub>SCL</sub>		0		1000	kHz
Hold Time (REPEATED) START Condition	t <sub>HD_STA</sub>		0.26			μs
SCL Low Period	t <sub>LOW</sub>		0.5			μs
SCL High Period	t <sub>HIGH</sub>		0.26			μs
Setup Time REPEATED START Condition	t <sub>SU_STA</sub>		0.26			μs
Data Hold Time	t <sub>HD_DAT</sub>		0			μs
Data Setup Time	t <sub>SU_DAT</sub>		50			ns
Setup Time for STOP Condition	t <sub>SU_STO</sub>		0.26			μs
Bus Free Time between STOP and START Condition	t <sub>BUF</sub>		0.5			μs
Pulse Width of Suppressed Spikes	t <sub>SP</sub>	Maximum pulse width of spikes that must be suppressed by the input filter	50			ns

**Note 5:** Design guidance only. Not production tested.

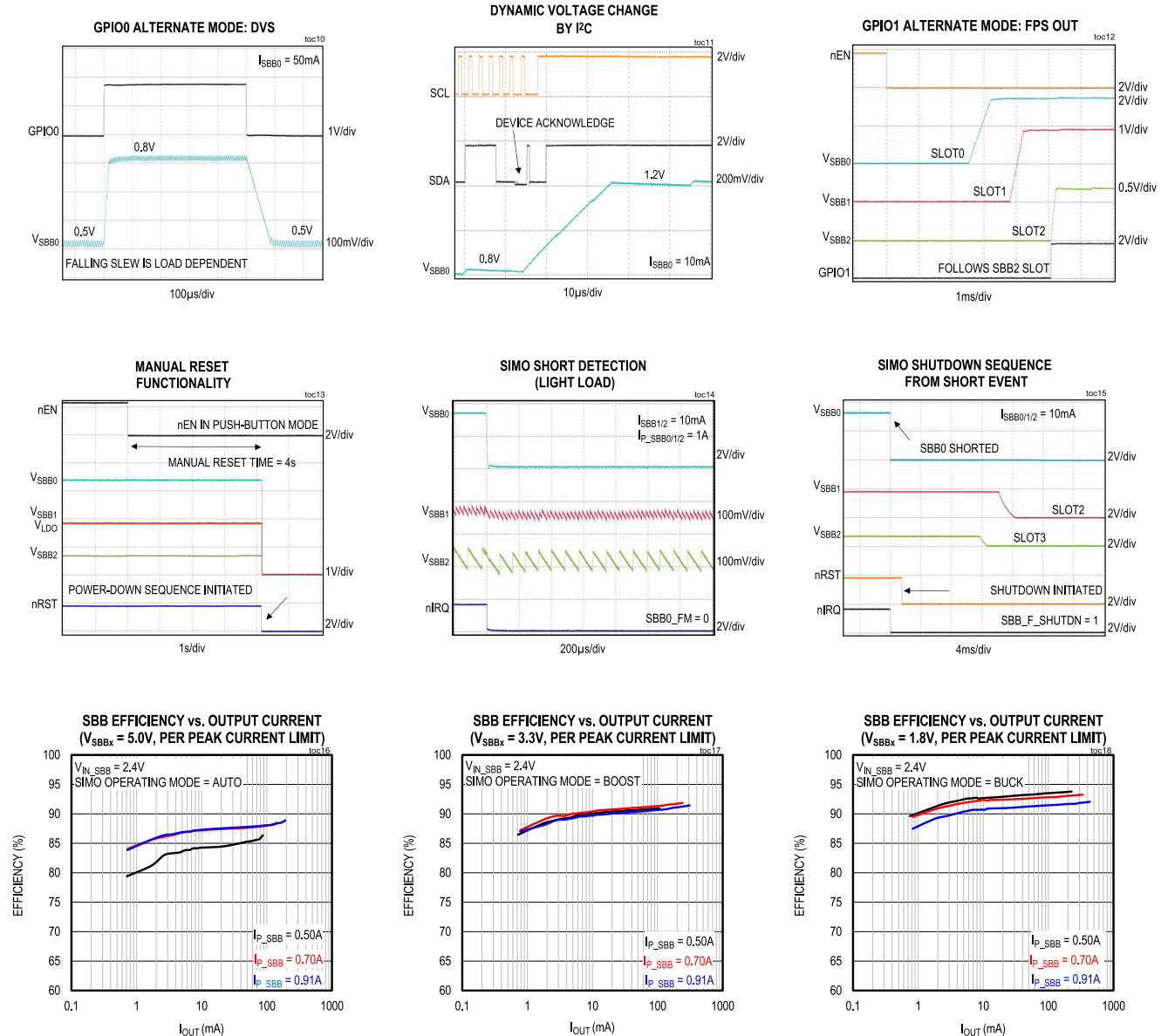
## Typical Operating Characteristics

(Typical Applications Circuit.  $V_{IN\_SBB} = 2.4V$ ,  $V_{IO} = 1.8V$ ,  $T_A = +25^\circ C$ , SIMO in automatic operation mode,  $V_{SBB0} = 3.3V$ ,  $I_{P\_SBB0} = 0.5A$ ,  $V_{SBB1} = 1.8V$ ,  $I_{P\_SBB1} = 0.5A$ ,  $V_{SBB2} = 0.65V$ ,  $I_{P\_SBB2} = 0.5A$ , unless otherwise noted. Inductor = DFE201612E-1R5M, 1.5 $\mu H$ , 72m $\Omega$ )



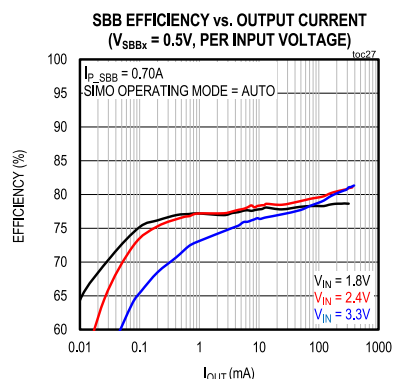
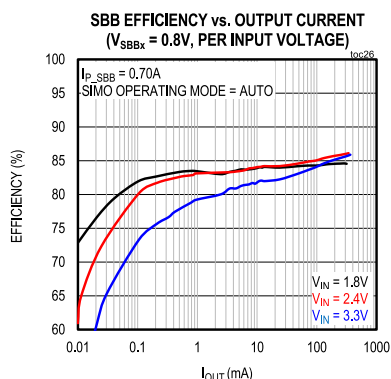
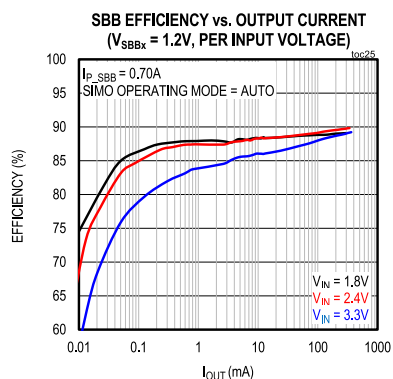
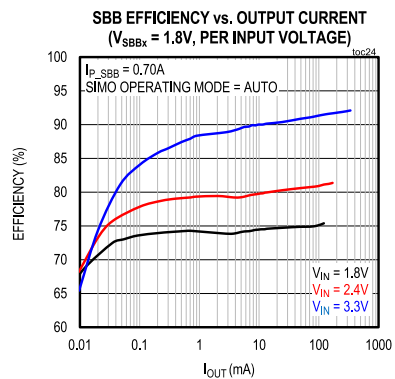
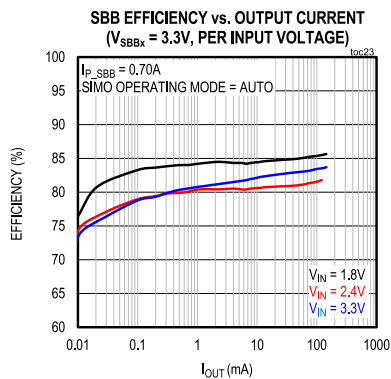
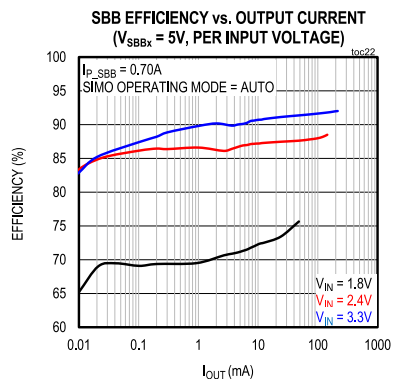
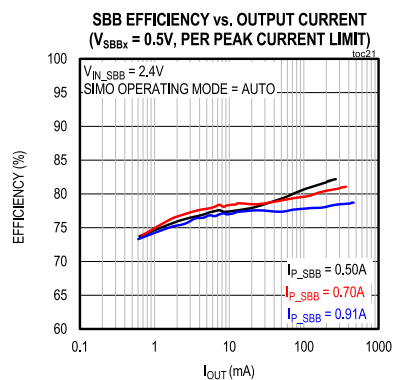
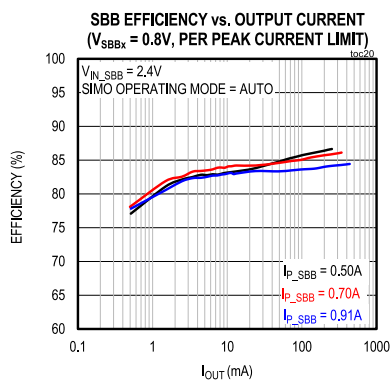
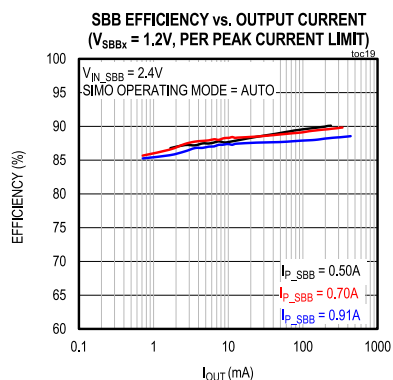
## Typical Operating Characteristics (continued)

(Typical Applications Circuit.  $V_{IN\_SBB} = 2.4V$ ,  $V_{IO} = 1.8V$ ,  $T_A = +25^\circ C$ , SIMO in automatic operation mode,  $V_{SBB0} = 3.3V$ ,  $I_{P\_SBB0} = 0.5A$ ,  $V_{SBB1} = 1.8V$ ,  $I_{P\_SBB1} = 0.5A$ ,  $V_{SBB2} = 0.65V$ ,  $I_{P\_SBB2} = 0.5A$ , unless otherwise noted. Inductor = DFE201612E-1R5M,  $1.5\mu H$ ,  $72m\Omega$ )



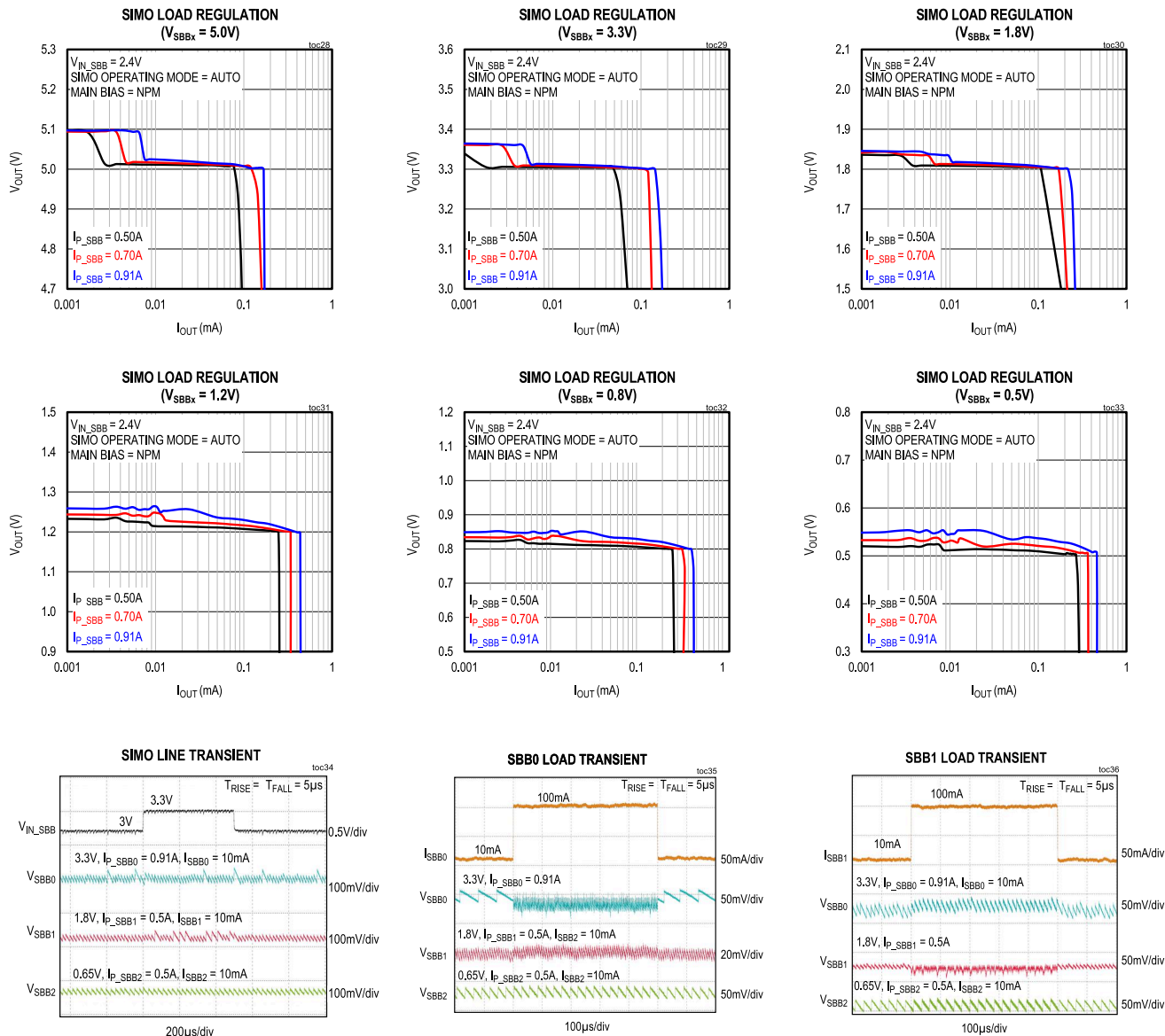
## Typical Operating Characteristics (continued)

(Typical Applications Circuit.  $V_{IN\_SBB} = 2.4V$ ,  $V_{IO} = 1.8V$ ,  $T_A = +25^\circ C$ , SIMO in automatic operation mode,  $V_{SBB0} = 3.3V$ ,  $I_{P\_SBB0} = 0.5A$ ,  $V_{SBB1} = 1.8V$ ,  $I_{P\_SBB1} = 0.5A$ ,  $V_{SBB2} = 0.65V$ ,  $I_{P\_SBB2} = 0.5A$ , unless otherwise noted. Inductor = DFE201612E-1R5M,  $1.5\mu H$ ,  $72m\Omega$ )



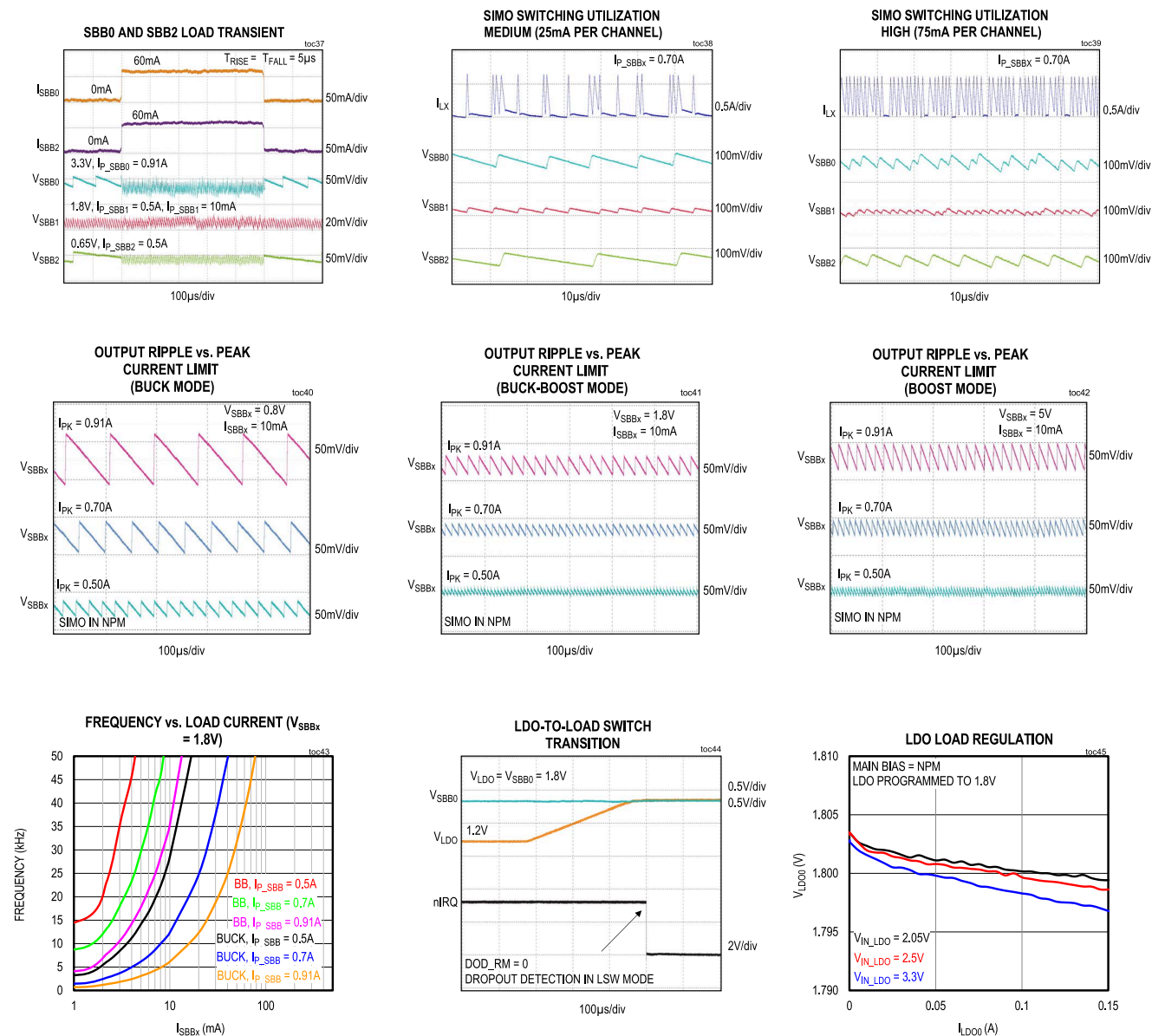
## Typical Operating Characteristics (continued)

(Typical Applications Circuit.  $V_{IN\_SBB} = 2.4V$ ,  $V_{IO} = 1.8V$ ,  $T_A = +25^\circ C$ , SIMO in automatic operation mode,  $V_{SBB0} = 3.3V$ ,  $I_{P\_SBB0} = 0.5A$ ,  $V_{SBB1} = 1.8V$ ,  $I_{P\_SBB1} = 0.5A$ ,  $V_{SBB2} = 0.65V$ ,  $I_{P\_SBB2} = 0.5A$ , unless otherwise noted. Inductor = DFE201612E-1R5M, 1.5 $\mu H$ , 72m $\Omega$ )



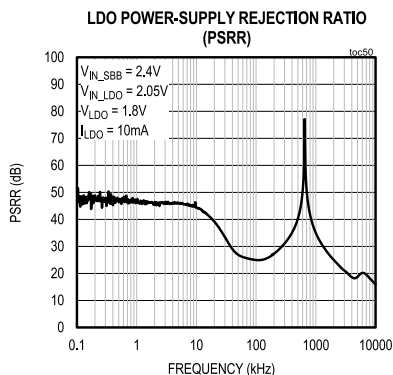
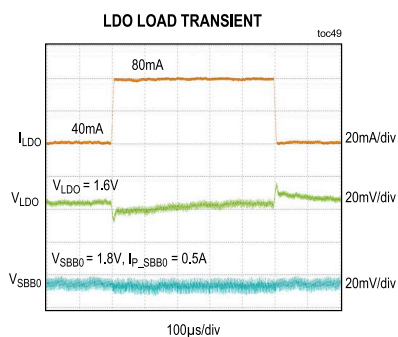
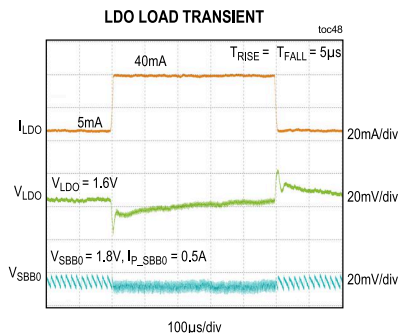
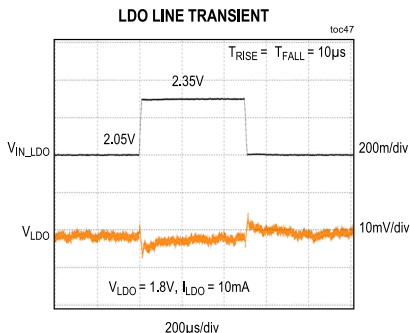
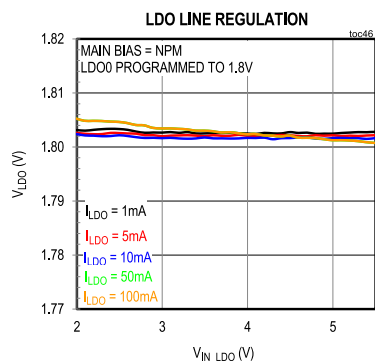
## Typical Operating Characteristics (continued)

(Typical Applications Circuit.  $V_{IN\_SBB} = 2.4V$ ,  $V_{IO} = 1.8V$ ,  $T_A = +25^\circ C$ , SIMO in automatic operation mode,  $V_{SBB0} = 3.3V$ ,  $I_{P\_SBB0} = 0.5A$ ,  $V_{SBB1} = 1.8V$ ,  $I_{P\_SBB1} = 0.5A$ ,  $V_{SBB2} = 0.65V$ ,  $I_{P\_SBB2} = 0.5A$ , unless otherwise noted. Inductor = DFE201612E-1R5M, 1.5 $\mu H$ , 72m $\Omega$ )



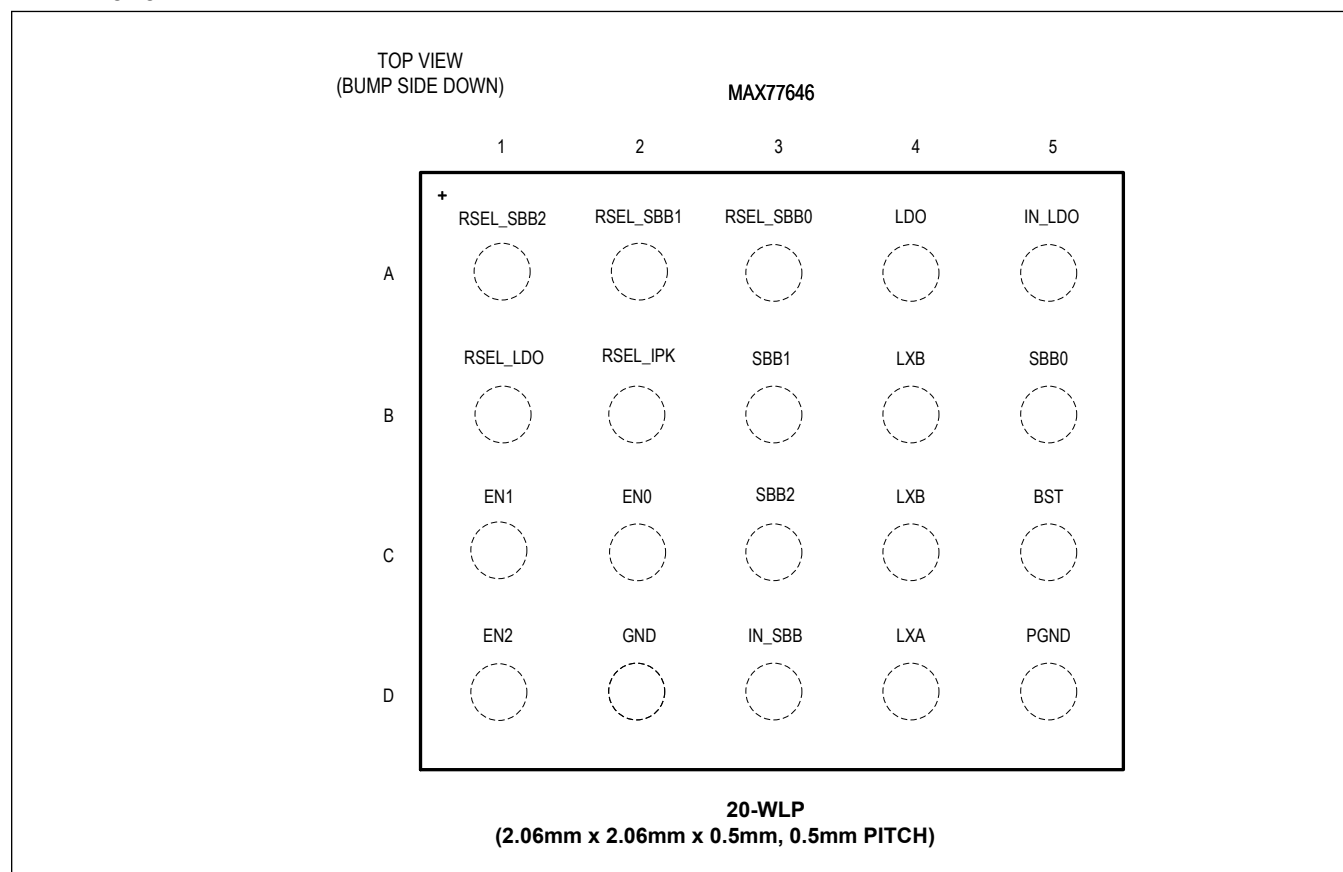
### Typical Operating Characteristics (continued)

(Typical Applications Circuit.  $V_{IN\_SBB} = 2.4V$ ,  $V_{IO} = 1.8V$ ,  $T_A = +25^\circ C$ , SIMO in automatic operation mode,  $V_{SBB0} = 3.3V$ ,  $I_{P\_SBB0} = 0.5A$ ,  $V_{SBB1} = 1.8V$ ,  $I_{P\_SBB1} = 0.5A$ ,  $V_{SBB2} = 0.65V$ ,  $I_{P\_SBB2} = 0.5A$ , unless otherwise noted. Inductor = DFE201612E-1R5M, 1.5 $\mu H$ , 72m $\Omega$ )



## Pin Configurations

### MAX77646

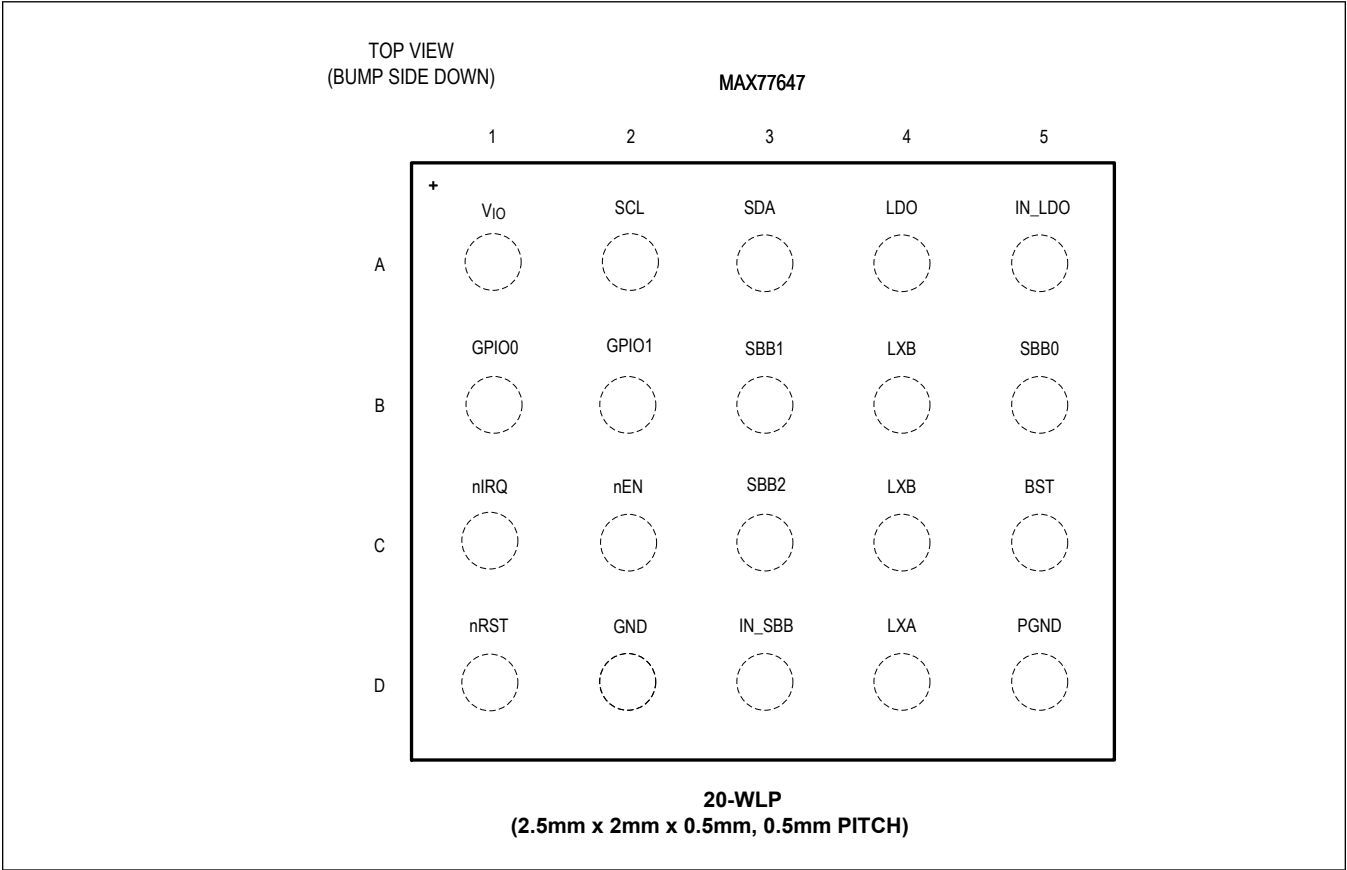




MAX77646/MAX77647

Ultra-Configurable SIMO PMIC Featuring 3-Output Buck-Boost, 1-LDO for Long Battery Life Primary Cell Applications

MAX77647



## Pin Description

PIN		NAME	FUNCTION	TYPE
MAX77646	MAX77647			
TOP LEVEL				
—	A1	V <sub>IO</sub>	I <sup>2</sup> C Interface and GPIO Driver Power	Power Input
—	A2	SCL	I <sup>2</sup> C Clock	Digital Input
—	A3	SDA	I <sup>2</sup> C Data	Digital I/O
—	B1	GPIO0	General Purpose Input/Output. The GPIO I/O stage is internally biased with V <sub>IO</sub> .	Digital I/O
—	B2	GPIO1	General Purpose Input/Output. The GPIO I/O stage is internally biased with V <sub>IO</sub> .	Digital I/O
—	C1	nIRQ	Active-Low, Open-Drain Interrupt Output. Connect a 100kΩ pullup resistor between nIRQ and a voltage equal to or less than V <sub>IN_SBB</sub> .	Digital Output
—	C2	nEN	Active-Low Enable Input. nEN supports push-button or slide-switch configurations. If not used, connect nEN to SYS and use the CNFG_SBBx_B.EN_SBBx[2:0] and CNFG_LDO_B.EN_LDO[2:0] bitfields to enable channels.	Digital Input
—	D1	nRST	Active-Low, Open-Drain Reset Output. Connect a 100kΩ pullup resistor between nRST and a voltage equal to or less than V <sub>IN_SBB</sub> .	Digital Output
A1	—	RSEL_SBB2	Select Resistor Pin SBB2. Connect a resistor from this pin to GND using the value to configure the output voltage of SBB2.	Resistor Input
A2	—	RSEL_SBB1	Select Resistor Pin SBB1. Connect a resistor from this pin to GND using the value to configure the output voltage of SBB1.	Resistor Input
A3	—	RSEL_SBB0	Select Resistor Pin SBB0. Connect a resistor from this pin to GND using the value to configure the output voltage of SBB0.	Resistor Input
B1	—	RSEL_LDO	Select Resistor Pin LDO. Connect a resistor from this pin to GND using the value to configure the output voltage of LDO.	Resistor Input
B2	—	RSEL_IPK	Select Resistor Pin IPK. Connect a resistor from this pin to GND using the value to configure the peak inductor current.	Resistor Input
C1	—	EN1	Enable Input for SBB1. Hold high to enable output regulation. Hold low to disable the output.	Digital Input
C2	—	EN0	Enable Input for SBB0. Hold high to enable output regulation. Hold low to disable the output.	Digital Input
D1	—	EN2	Enable Input for SBB2 and LDO. Hold high to enable output regulation. Hold low to disable the output.	Digital Input
D2	D2	GND	Quiet Ground. Connect GND to PGND, and the low-impedance ground plane of the PCB.	Ground
SIMO BUCK-BOOST				
B3	B3	SBB1	SIMO Buck-Boost Output 1. SBB1 is the power output for channel 1 of the SIMO buck-boost. Bypass SBB1 to PGND with a 22μF ceramic capacitor. If not used, see the <a href="#">Unused Outputs</a> section.	Power Output
B4, C4	B4, C4	LXB	Switching Node B. LXB is driven between PGND and SBBx when SBBx is enabled. LXB is driven to PGND when all SIMO channels are disabled. Connect a 1.5μH inductor between LXA and LXB.	Power Input
B5	B5	SBB0	SIMO Buck-Boost Output 0. SBB0 is the power output for channel 0 of the SIMO buck-boost. Bypass SBB0 to PGND with a 22μF ceramic capacitor. If not used, see the <a href="#">Unused Outputs</a> section.	Power Output

## Pin Description (continued)

PIN		NAME	FUNCTION	TYPE
MAX77646	MAX77647			
C3	C3	SBB2	SIMO Buck-Boost Output 2. SBB2 is the power output for channel 2 of the SIMO buck-boost. Bypass SBB0 to PGND with a 22 $\mu$ F ceramic capacitor. If not used, see the <a href="#">Unused Outputs</a> section.	Power Output
C5	C5	BST	SIMO Power Input for the High-Side Output NMOS Drivers. Connect a 10nF ceramic capacitor between BST and LXB.	Power Input
D3	D3	IN_SBB	SIMO Power Input. Bypass to PGND with a minimum of 10 $\mu$ F ceramic capacitor as close as possible to the IN_SBB pin.	Power Input
D4	D4	LXA	Switching Node A. LXA is driven between PGND and IN_SBB when any SIMO channel is enabled. LXA is driven to PGND when all SIMO channels are disabled. Connect a 1.5 $\mu$ H inductor between LXA and LXB.	Power I/O
D5	D5	PGND	Power Ground for the SIMO Low-Side FETs. Connect PGND to GND and the low-impedance ground plane of the PCB.	Ground
<b>LDO</b>				
A4	A4	LDO	Linear Regulator Output. Bypass with a 1.0 $\mu$ F ceramic capacitor to GND. If not used, disable LDO and connect this pin to ground or leave unconnected.	Power Output
A5	A5	IN_LDO	Linear Regulator Input. If connected to a SIMO output with a short trace, IN_LDO can share the output's capacitor. Otherwise, bypass with a 2.2 $\mu$ F ceramic capacitor to ground. If not used, connect to ground or leave unconnected.	Power Input

## Detailed Description

The MAX77646/MAX77647 provide highly-integrated power management solutions for low-power applications.

Four regulators are integrated within this device (see [Table 1](#)). A single-inductor, multiple output (SIMO) buck-boost regulator efficiently provides three independently programmable power rails. A 150mA low-dropout linear regulator (LDO) provides ripple rejection for audio and other noise-sensitive applications.

The MAX77647 includes other features such as two GPIOs with alternate modes for various system requirements. A bidirectional I<sup>2</sup>C serial interface allows for configuring and checking the status of the device. An internal on/off controller provides regulator sequencing and supervisory functionality for the device.

**Table 1. Regulator Summary**

REGULATOR NAME	REGULATOR TOPOLOGY	MAXIMUM I <sub>OUT</sub> (mA)	V <sub>IN</sub> RANGE (V)	MAX77647 V <sub>OUT</sub> RANGE/ RESOLUTION
SBB0	SIMO	Up to 500*	1.8 to 3.3	0.5V to 5.1V in 25mV steps
SBB1	SIMO		1.8 to 3.3	0.5V to 5.1V in 25mV steps
SBB2	SIMO		1.8 to 3.3	0.5V to 5.1V in 25mV steps
LDO	PMOS LDO	150	1.8 to 3.3	0.5V to 5.0V in 25mV steps

\*Shared capacity with other SBBx channels. See the [SIMO Available Output Current](#) section for more information.

## Part Number Decoding

The MAX77647 has different one-time programmable (OTP) options and variants to support a variety of applications. The OTP options set default settings such as output voltage. Variants are versions of the MAX77647 with different features. See the [Part Number Decode](#) section for how to identify these. [Table 2](#) and [Table 3](#) list all available variants and OTP options respectively. Refer to [Analog Devices Products Naming Convention](#) for more details.

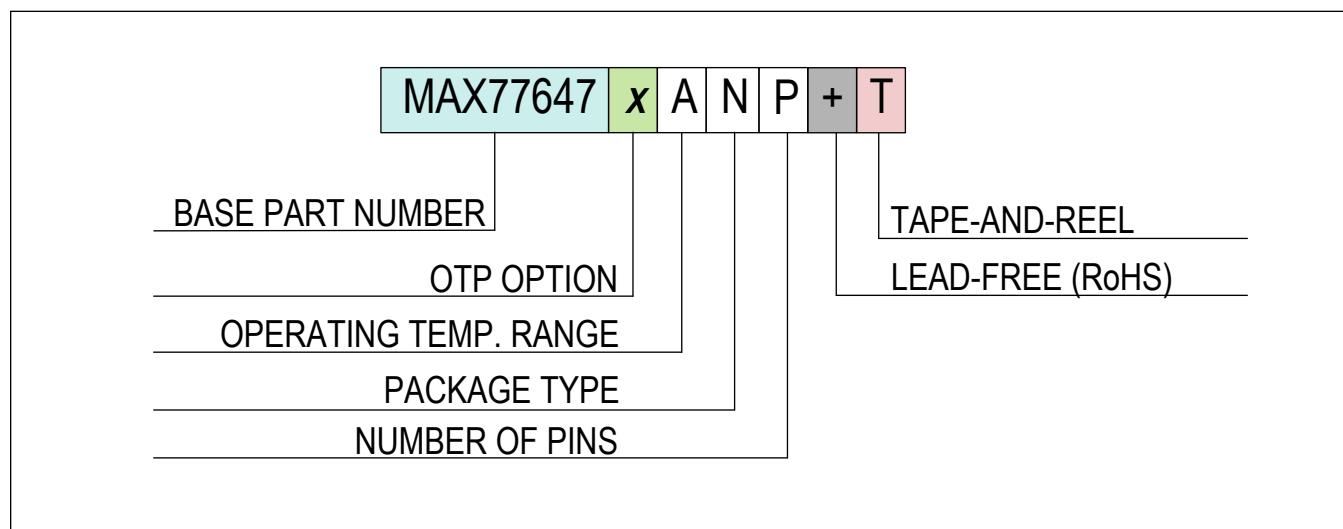


Figure 1. Part Number Decode

**Table 2. Variants Table**

PART NUMBER	MAX77646	MAX77647
Resistor Programmable Output Voltages	Yes	No
Resistor Programmable Peak Current	Yes	No

**Table 2. Variants Table (continued)**

Supports LDO/LSW	Yes	Yes
Include GPIOs	No	Yes
Individual Enable Pins for LDO and SIMO Outputs	Yes	No
Programmable Flexible Power Sequencer	No	Yes
Watchdog Timer	No	Yes
Supports I <sup>2</sup> C Serial Communication	No	Yes

**Table 3. OTP Options Table**

			OTP LETTER AND SETTINGS
BLOCK	BIT FIELD NAME	SETTING NAME	A
Global	SBIA_LPM	Bias Low-Power Mode	LPM
	DBEN_nEN	nEN Debounce Time	500μs
	nEN_MODE	nEN Mode	Push-Button
	T_MRST	Manual Reset Time	4s
	GPIO_LPM	SIMO Power Mode	LPM
	ALT_GPIO0	GPIO0 Mode	GPIO
	ALT_GPIO1	GPIO1 Mode	GPIO
	ADDR	I <sup>2</sup> C Address (7-Bit)	0x48
	CID[4:0]	Chip ID	0x01
Watchdog	WDT_LOCK	Watchdog Timer Disable Control	Unlocked
	WDT_EN	Watchdog Timer Enable	Disabled

Table 3. OTP Options Table (continued)

SIMO	SBB_F_SHUTDOWN	SBB Shutdown from SBB Faults	Disabled
	TV_SBB0[7:0]	SBB0 $V_{OUT}$	3.300V
	IP_SBB0[1:0]	SBB0 Inductor Current Peak Limit	0.500A
	OP_MODE[1:0] (SBB0)	SBB0 Operating Mode	Automatic
	ADE_SBB0	Active-Discharge Resistor Enable	Enabled
	EN_SBB0[2:0]	SBB0 Enable Control	OFF
	TV_SBB1[7:0]	SBB1 $V_{OUT}$	1.800V
	IP_SBB1[1:0]	SBB1 Inductor Current Peak Limit	0.500A
	OP_MODE[1:0] (SBB1)	SBB1 Operating Mode	Automatic
	ADE_SBB1	Active-Discharge Resistor Enable	Enabled
	EN_SBB1[2:0]	SBB1 Enable Control	FPS Slot 1
	TV_SBB2[7:0]	SBB2 $V_{OUT}$	0.650V
	IP_SBB2[1:0]	SBB2 Inductor Current Peak Limit	0.500A
	OP_MODE[1:0] (SBB2)	SBB2 Operating Mode	Automatic
	ADE_SBB2	Active-Discharge Resistor Enable	Enabled
	EN_SBB2[2:0]	SBB2 Enable Control	OFF
LDO	TV_OFS_LDO and TV_LDO[6:0]	LDO $V_{OUT}$	1.800V
	LDO_MD	LDO or LSW Mode	LDO
	ADE_LDO	Active-Discharge Resistor Enable	Enabled
	EN_LDO[2:0]	LDO Enable Control	OFF

## Detailed Description—Global Resources

The global resources encompass a set of circuits that serve the entire device and ensure safe, consistent, and reliable operation.

### Features and Benefits

- Voltage Monitors
  - IN power-on reset (POR) comparator generates a reset signal upon power-up
  - IN undervoltage ensures repeatable behavior when power is applied to and removed from the device
  - IN overvoltage monitor inhibits operation with overvoltage power sources to ensure reliability in faulty environments
- Thermal Monitors
  - +145°C junction temperature shutdown
- Manual Reset
  - 4s or 8s period
- Wake-Up Events
  - nEN input assertion
- Interrupt Handler
  - Interrupt output (nIRQ)
  - All interrupts are maskable
- Push-Button/Slide-Switch/Logic Mode On-Key (nEN)
  - Configurable push-button/slide-switch functionality
  - 500 $\mu$ s or 30ms debounce timer interfaces directly with mechanical switches
- On/Off Controller
  - Startup/shutdown sequencing
  - Programmable sequencing delay
- GPIO, nRST Digital I/Os

### Voltage Monitors

The device monitors the input voltage (IN\_SBB and IN\_LDO) to ensure proper operation using three comparators (POR, UVLO, and OVLO). These comparators include hysteresis to prevent their outputs from toggling between states during noisy system transitions.

#### IN\_SBB POR Comparator

The IN\_SBB POR comparator monitors  $V_{IN\_SBB}$  and generates a power-on reset (POR) signal. When  $V_{IN\_SBB}$  is below  $V_{POR}$ , the device is held in reset. When  $V_{IN\_SBB}$  rises above  $V_{POR}$ , internal signals and on-chip memory stabilize and the device is released from reset.

#### IN\_SBB Undervoltage-Lockout Comparator

The IN\_SBB undervoltage-lockout (UVLO) comparator monitors  $V_{IN\_SBB}$  and generates a  $INSBB\_UVLO$  signal when the  $V_{IN\_SBB}$  falls below the UVLO threshold. The  $INSBB\_UVLO$  signal is provided to the top-level digital controller. See [Figure 6](#), [Table 5](#), [Figure 7](#), and [Table 6](#) for additional information regarding the UVLO comparator.

#### IN\_SBB Overvoltage-Lockout Comparator

The device is rated for 5.5V maximum operating voltage ( $V_{IN\_SBB}$ ) with an absolute maximum input voltage of 6.0V. An overvoltage-lockout monitor increases the robustness of the device by inhibiting operation when the supply voltage is greater than  $V_{INSBB\_OVLO}$ . See [Figure 6](#), [Table 5](#), [Figure 7](#), and [Table 6](#) for additional information regarding the OVLO comparator.

## Thermal Monitors

The MAX77647 has three global on-chip thermal sensors:

- Junction Temperature Alarm 1 → 80°C
- Junction Temperature Alarm 2 → 100°C
- Junction Temperature Shutdown → 145°C

The junction temperature alarms have maskable rising interrupts as well as status bits (see the [Register Map](#) section for more information). Unmasking these thermal alarms is recommended for all systems. If the first alarm is triggered, the system software should attempt to lower system power dissipation. If the second alarm is triggered, then attempts to lower the power dissipation were unsuccessful and the system software should turn the device off. Finally, if the junction temperature rises to junction temperature shutdown, then the MAX77647 sets the ERCFLAG.TOVLD bit and automatically turns itself off.

After a junction temperature shutdown event, the system can be enabled again. The system software can read the ERCFLAG register during initialization to see ERCFLAG.TOVLD = 1 and log that an extreme thermal event has occurred.

## Chip Identification

The MAX77647 offers different one-time-programmable (OTP) options to, for example, set the default output voltages. These options are identified by the chip identification number, which can be read in the CID register.

## nEN Enable Input (MAX77647)

The nEN is an active-low internally debounced digital input that typically comes from the system's on-key. The debounce time is programmable with CNFG\_GLBLx.DBEN\_nEN. The primary purpose of this input is to generate a wake-up signal for the PMIC that turns on the regulators. Maskable rising/falling interrupts are available for nEN (INT\_GLBLx.nEN\_R and INT\_GLBLx.nEN\_F) for alternate functionality.

The nEN input can be configured to work either with a push-button (CNFG\_GLBL0.nEN\_MODE = 0b00), a slide-switch (CNFG\_GLBL0.nEN\_MODE = 0b01), or logic mode (CNFG\_GLBL0.nEN\_MODE = 0b10). See [Figure 3](#) for more information. In both push-button mode and slide-switch mode, the on/off controller looks for a falling edge on the nEN input to initiate a power-up sequence.

## EN Enable Input (MAX77646)

The MAX77646 features three enable pins to individually control the on/off state of the SIMO outputs. Drive each enable pin high to turn its respective SIMO output on. Each enable pin has a 50nA pulldown current to ground (as shown in [Figure 2](#)). The pins can be tied high for always on-applications. Do not leave the EN pins unconnected.

EN2 controls the ON/OFF state of both SBB2 output and LDO regulator together.

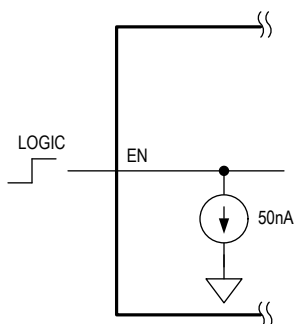


Figure 2. EN Pulldown



**nEN Manual Reset (MAX77647)**

The nEN works as a manual reset input when the on/off controller is in the "Resource-On" state. The manual reset function is useful for forcing a power-down in case communication with the processor fails. When nEN is configured for push-button mode and the input is asserted (nEN = LOW) for an extended period ( $t_{MRST}$ ), the on/off controller initiates a power-down sequence and goes to shutdown mode. When nEN is configured for slide-switch mode and the input is deasserted (nEN = HIGH) for an extended period ( $t_{MRST}$ ), the on/off controller initiates a power-down sequence and goes to standby mode. When nEN is configured as a logic mode, the on/off controller initiates a power-up sequence and goes into resource ON mode when the input is asserted (nEN = LOW). When the input is deasserted (nEN = HIGH), the on/off controller initiates a power-down sequence and goes into shutdown mode.

**nEN Triple-Functionality: Push-Button vs. Slide-Switch vs. Logic (MAX77647)**

The nEN digital input can be configured to work with a push-button, a slide-switch, or a logic input. [Figure 3](#) shows nEN's triple functionality for power-on sequencing and manual reset. The default configuration of the device is push-button mode (CNFG\_GLBL0.nEN\_MODE = 0b00) and no additional programming is necessary. Applications that use a slide-switch on-key and logic configuration must set CNFG\_GLBL0.nEN\_MODE = 0b01 and CNFG\_GLBL0.nEN\_MODE = 0b10 respectively within  $t_{MRST}$ .

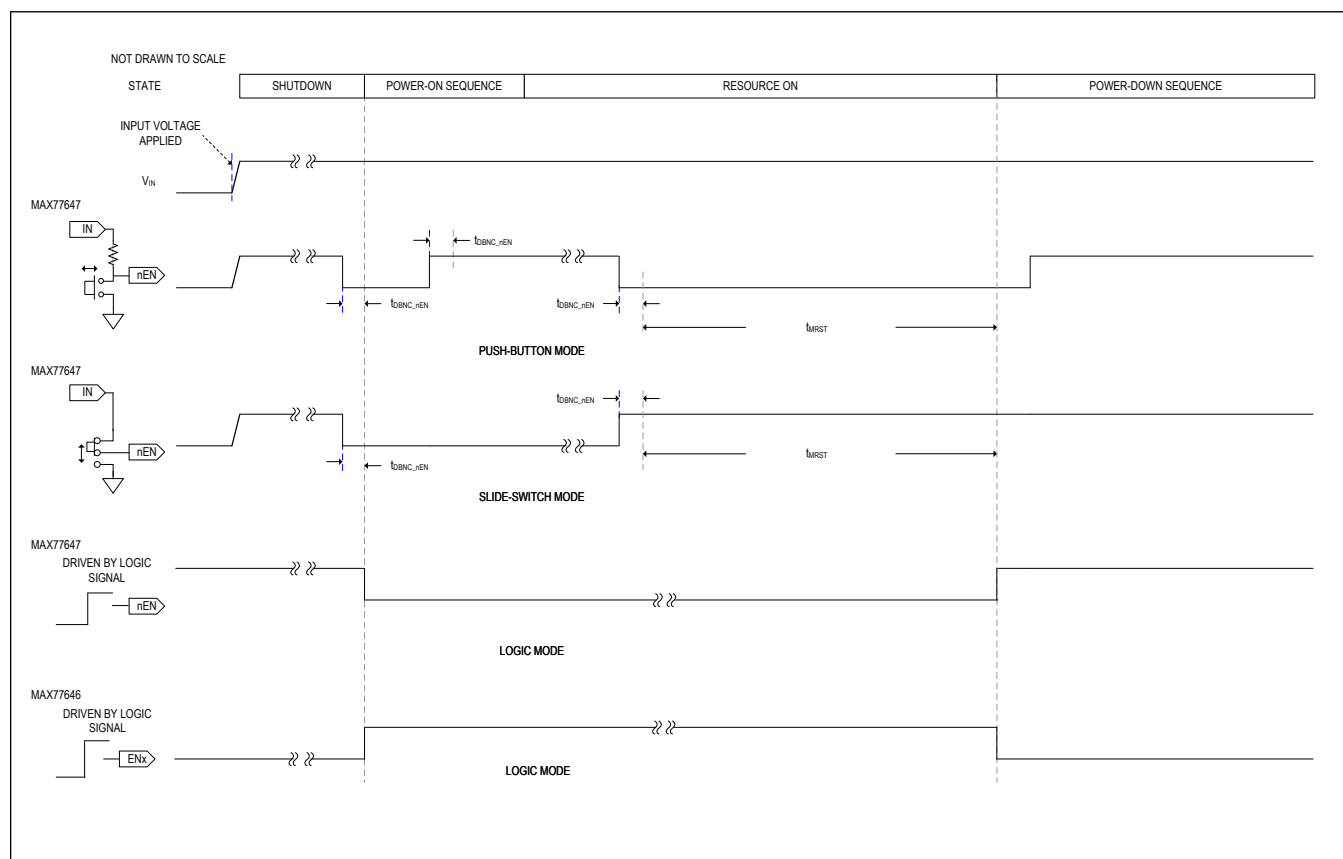


Figure 3. nEN Usage Timing Diagram

**nEN Internal Pullup Resistors to  $V_{IN\_SBB}$** 

The nEN logic thresholds are referenced to  $V_{IN\_SBB}$ . There are internal pullup resistors between nEN and  $V_{CCINT}$  ( $R_{nEN\_PU}$ ), which can be configured with the CNFG\_GLBL0.PU\_DIS bit. See [Figure 4](#). While PU\_DIS = 0, the pullup value is approximately 200kΩ. While PU\_DIS = 1, the pullup value is 10MΩ.

Applications using a slide-switch on-key or push-pull digital output connected to nEN can reduce quiescent current

consumption by changing pullup strength to 10MΩ. Applications using normally-open, momentary, and push-button on-keys (as shown in [Figure 4](#)) do not create this leakage path and should use the stronger 200kΩ pullup option.

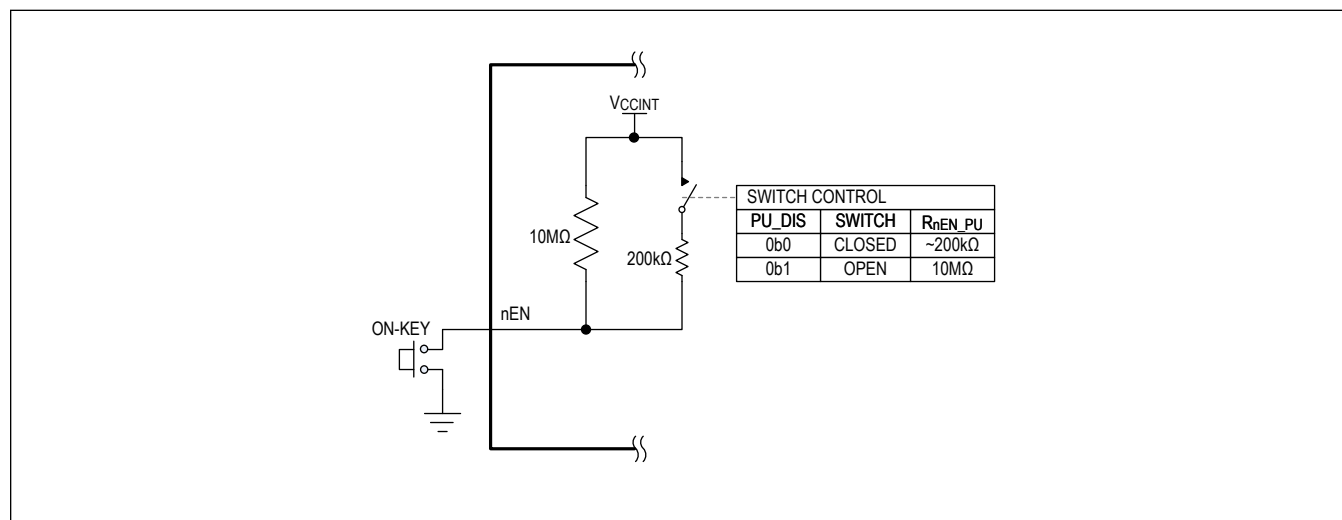


Figure 4. nEN Pullup Resistor Configuration

### Interrupts (nIRQ) (MAX77647)

The nIRQ (MAX77647) is an active-low, open-drain output that is typically routed to the host processor's interrupt input to signal an important change in device status. See the [Register Map](#) section for a comprehensive list of all interrupt bits and status registers.

A pullup resistor to a voltage less than or equal to  $V_{IN\_SBB}$  is required for this node. The nIRQ is the logical *NOR* of all unmasked interrupt bits in the register map.

All interrupts are masked by default. Masked interrupt bits do not cause the nIRQ pin to change. Unmask the interrupt bits to allow the nIRQ to assert.

### Reset Output (nRST) (MAX77647)

The nRST (MAX77647) is an open-drain, active-low output that is typically used to hold the processor in a reset state when the device is powered down. During a power-up sequence, the nRST deasserts after the last regulator in the power-up chain is enabled ( $t_{RSTODD}$ ). During a power-down sequence, the nRST output asserts before any regulator is powered down ( $t_{RSTOAD}$ ).

A pullup resistor to a voltage less than or equal to  $V_{IN\_SBB}$  is required for this node.

### General-Purpose Input Output (GPIO) (MAX77647)

The MAX77647 provides two general-purpose input/output (GPIO) pins to increase system flexibility. See [Figure 5](#) for more details.

Clear CNFG\_GPIOx.DIR = 0b0 to configure GPIO as a general-purpose output (GPO). The GPO can either be in push-pull mode (CNFG\_GPIOx.DRV = 1) or open-drain mode (CNFG\_GPIOx.DRV = 0).

- The push-pull output mode is ideal for applications that need fast (~2ns) edges and low power consumption.
- The open-drain mode requires an external pullup resistor (typically 10kΩ to 100kΩ). Connect the external pullup resistor to a bias voltage that is less than or equal to  $V_{IO}$ .
  - The open-drain mode can be used to communicate to different logic domains. For example, to send a signal from the GPO on a 1.8V logic domain ( $V_{IO} = 1.8V$ ) to a device on a 1.2V logic domain, connect the external pullup resistor to 1.2V.
  - The open-drain mode can be used to connect several open-drain (or open-collector) devices together on the same

bus to create wired logic (wired AND logic is positive-true; wired OR logic is negative-true).

- The general-purpose input (GPI) functions are still available while the pin is configured as a GPO. In other words, the CNFG\_GPIOx.DI (input status) bit still functions and does not collide with the state of the CNFG\_GPIOx.DIR bit.

Set CNFG\_GPIOx.DIR to have the GPIO function as a GPI. The GPI features a 30ms debounce timer ( $t_{\text{DBNC\_GPI}}$ ) that can be enabled or disabled with DBEN\_GPIO.

- Enable the debounce timer (CNFG\_GPIOx.DBEN\_GPIO = 1) if the GPI is connected to a device that can bounce or chatter, like a mechanical switch.
- If the GPI is connected to a circuit with clean logic transitions and no risk of bounce, disable the debounce timer (CNFG\_GPIOx.DBEN\_GPIO = 0) to eliminate logic delays. With no debounce timer, the GPI input logic propagates to nIRQ in 10ns.

A dedicated internal oscillator is used to create the 30ms ( $t_{\text{DBNC\_GPI}}$ ) debounce timer. To obtain low  $V_{\text{IO}}$  supply current, ensure the GPIO voltage is either logic-high or logic-low. If the GPIO pin is unconnected (either as a GPI or an open-drain GPO) and  $V_{\text{IO}}$  is powered, the GPIO voltage trends towards the logic level gray area ( $0.3 \times V_{\text{IO}} < V_{\text{GPIO}} < 0.7 \times V_{\text{IO}}$ ). If  $V_{\text{GPIO}}$  is in the gray area,  $V_{\text{IO}}$  current can be more than 10 $\mu\text{A}$ .

The GPI features edge detectors that feed into the top-level interrupt system of the chip. This allows software to use interrupts to service events associated with a GPI change instead of polling for these changes.

- If the application wants nIRQ to go low **only on a GPI rising edge**, then it should **clear** the GPI rising edge interrupt mask bit (INTM\_GLBLx.GPI\_RM = 0) and **set** the GPI falling edge interrupt mask bit (INTM\_GLBLx.GPI\_FM = 1).
- If the application wants nIRQ to go low **only on a GPI falling edge**, then it should **set** the GPI rising edge interrupt mask bit (INTM\_GLBLx.GPI\_RM = 1) and **clear** the GPI falling edge interrupt mask bit (INTM\_GLBLx.GPI\_FM = 0).
- If the application wants nIRQ to go low **on both GPI falling and rising edges**, then it should **clear** the GPI rising edge interrupt mask bit (INTM\_GLBLx.GPI\_RM = 0) and **clear** the GPI falling edge interrupt mask bit (INTM\_GLBLx.GPI\_FM = 0).

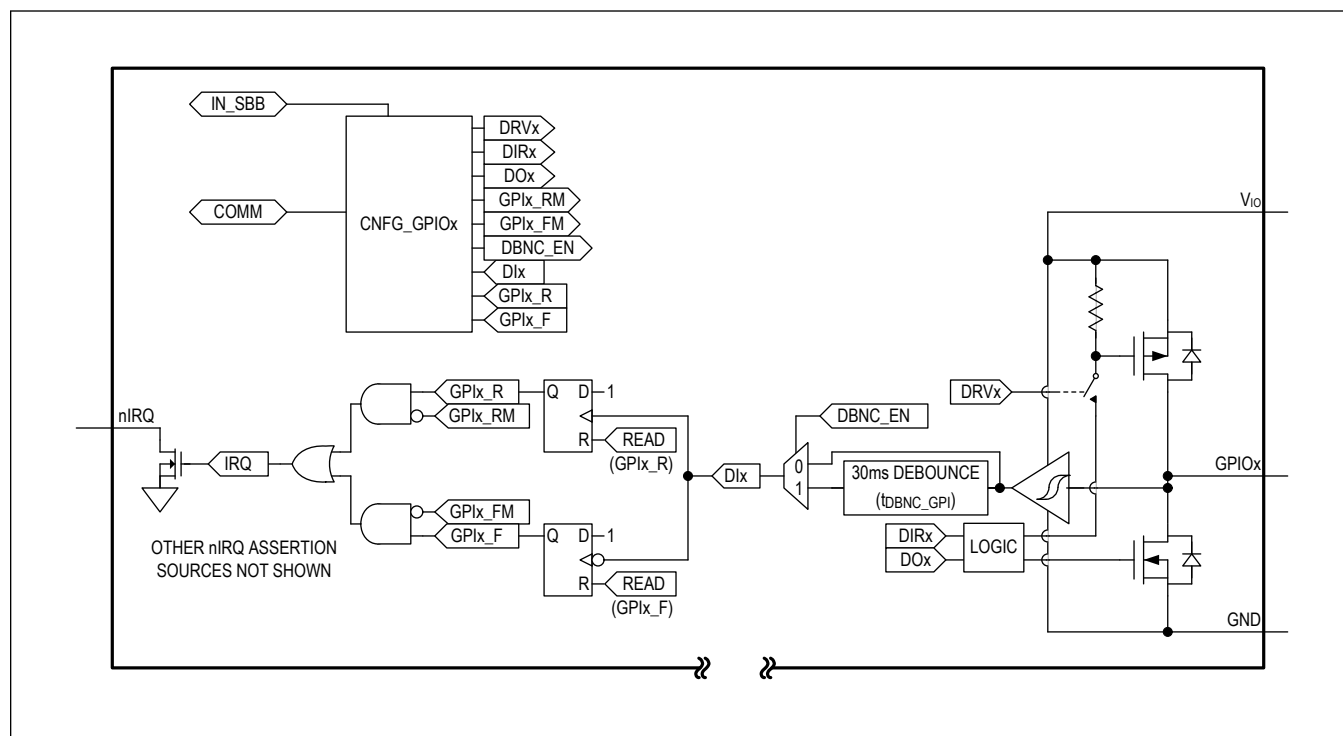


Figure 5. GPIOx Block Diagram

**Alternate Mode (MAX77647)**

The GPIO in the MAX77647 can be configured to have a different function if required. Whether the GPIO is in standard GPIO mode or alternate mode can be checked by reading the CNFG\_GPIOx.ALT\_GPIOx bit and the CNFG\_SBB\_TOP.GPIOx\_LPM bit simultaneously. [Table 4](#) summarizes the alternate functionality of each GPIO pin.

**Table 4. GPIO MODE**

GPIOx	CNFG_GPIOx REGISTER		
	ALT_GPIOx = 0	ALT_GPIOx = 1	
		GPIOx_LPM = 0	GPIOx_LPM = 1
GPIO0	Standard GPIO	Disable all SBBx Low-Power Mode	Active-high input controls the DVS feature for SBB0
GPIO1			Active-high output of SBB2's flexible power sequencer (FPS) slot

**On/Off Controller**

The on/off controller monitors multiple power-up (wake-up) and power-down (shutdown) conditions to enable or disable resources that are necessary for the system and its processor to move between its operating modes.

Many systems have one power management controller and one processor and rely on the on/off controller to be the master controller. In this case, the on/off controller receives wake-up events and enables some or all of the regulators to power up a processor. That processor then manages the system. To conceptualize this master operation, see [Figure 7](#) and [Table 6](#). A typical path through the on/off controller is:

1. Apply a battery and start in the shutdown state.
2. Press the system's on-key (nEN = LOW) and follow transitions 3 and 4 to the resource-on state. If any resources are on the FPS, transitions 5A and 5B are followed.
3. The device performs its desired functions in the resource-on state. When it is ready to turn off, a manual reset first drives the transition through transitions 6A and 6B for FPS power down then through the 7 and 0 states to the shutdown state.

Some systems have several power management blocks, a main processor, and subprocessors. These systems can use this device as a subpower management block for a peripheral portion of circuitry as long as there is an I<sup>2</sup>C port available from a higher level processor. To conceptualize this operation, see [Figure 7](#) and [Table 6](#). A typical path through the on/off controller used in this way is:

1. Apply a battery to the system and start in the shutdown state.
2. The higher level processor can now control this device's resources with I<sup>2</sup>C commands (e.g., turn on/off regulators).
3. When the higher level processor is ready to turn this device off, it turns off everything through I<sup>2</sup>C to transition along path 7 to the shutdown state.

Note that in this style of operation, the CNFG\_GLBL0.SFT\_CTRL[1:0] bits should not be used to turn the device off. The CNFG\_GLBL0.SFT\_CTRL[1:0] bits establish directives to the on/off controller itself that does not make sense in this subpower management block operation. If the processor uses I<sup>2</sup>C commands to enable the device's resources, the processor should also use I<sup>2</sup>C commands to disable them.

## Top Level On/Off Controller (MAX77646)

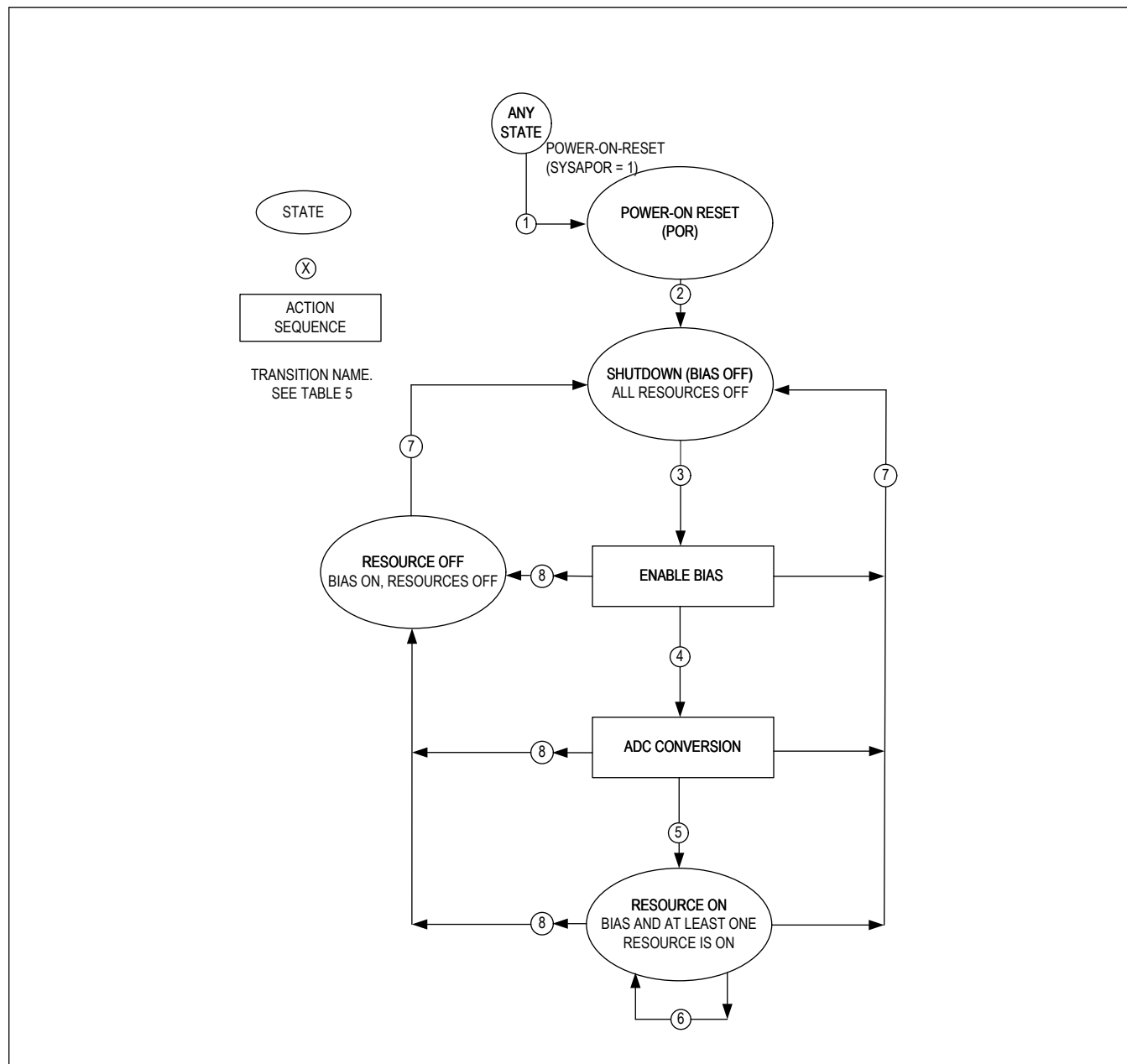


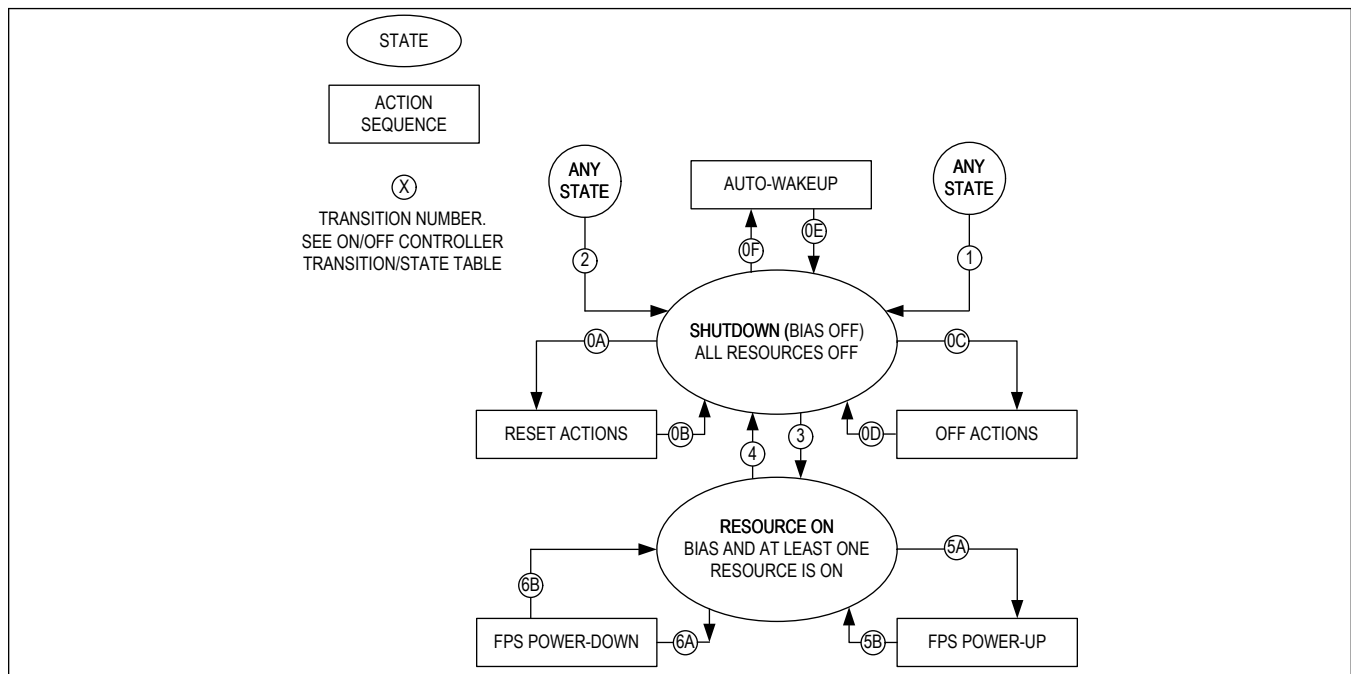
Figure 6. Top Level On/Off Controller State Diagram (MAX77646)

Table 5. On/Off Controller Transition/State (MAX77646)

TRANSITION	CONDITION (TRANSITION HAPPENS WHEN...)
1	IN_SBB voltage is above the POR threshold ( $V_{IN\_SBB} > V_{POR}$ )
2	$V_{IN\_SBB} \geq V_{INSBB\_UVLO\_R}$
3	First ENx transitions from low to high

**Table 5. On/Off Controller Transition/State (MAX77646) (continued)**

4	Main Bias enable time expires ( $t_{\text{SBias\_EN}}$ )
5	ADC detection time expires ( $t_{\text{RSEL}}$ )
6	Additional ENx transitions from low to high <b>OR</b> ENx transitions from high to low (as long as one SIMO channel or the LDO is enabled)
7	Last ENx transitions from high to low (all SIMO channels and the LDO are disabled)
8	Chip overtemperature lockout ( $T_J > T_{\text{OTLO}}$ ) <b>OR</b> IN_SBB undervoltage lockout ( $V_{\text{IN\_SBB}} < V_{\text{INSBB\_UVLO}}$ ) <b>OR</b> IN_SBB overvoltage lockout ( $V_{\text{IN\_SBB}} > V_{\text{INSBB\_OVLO}}$ )

**Top Level On/Off Controller (MAX77647)****Figure 7. Top Level On/Off Controller State Diagram (MAX77647)****Table 6. On/Off Controller Transition/State (MAX77647)**

TRANSITION	CONDITION (TRANSITION HAPPENS WHEN...)
0A	Software cold reset (CNFG_GLBL0.SFT_CTRL[1:0] = 0b01) <b>OR</b> Watchdog timer expired and caused reset (ERCFLAG.WDT_RST = 1, CNFG_WDT.WDT_MODE = 1)
0B	Reset actions completed
0C	Software power-off (CNFG_GLBL0.SFT_CTRL[1:0] = 0b10) <b>OR</b> Watchdog expired and caused power-off (ERCFLAG.WDT_OFF = 1, CNFG_WDT.WDT_MODE = 0) <b>OR</b> Chip overtemperature lockout ( $T_J > T_{\text{OTLO}}$ ) <b>OR</b> IN_SBB undervoltage lockout ( $V_{\text{IN\_SBB}} < V_{\text{INSBB\_UVLO}} + V_{\text{INSBB\_UVLO\_HYS}}$ ) <b>OR</b> IN_SBB overvoltage lockout ( $V_{\text{IN\_SBB}} > V_{\text{INSBB\_OVLO}}$ ) <b>OR</b> Manual reset occurred (ERCFLAG.MRST = 1)
0D	Off actions completed
0E	Software auto wake-up (CNFG_GLBL0.SFT_CTRL = 0b11)
0F	Auto wake-up actions completed

**Table 6. On/Off Controller Transition/State (MAX77647) (continued)**

<b>1</b>	IN_SBB voltage is above the POR threshold ( $V_{IN\_SBB} > V_{POR}$ )
<b>2</b>	Chip overtemperature lockout ( $T_J > T_{OTLO}$ ) <b>OR</b> IN_SBB undervoltage lockout ( $V_{IN\_SBB} < V_{INSBB\_UVLO} + V_{INSBB\_UVLO\_HYS}$ ) <b>OR</b> IN_SBB overvoltage lockout ( $V_{IN\_SBB} > V_{INSBB\_OVLO}$ )
<b>3</b>	Any resources force enabled <b>OR</b> Internal wake-up flags are set (see the <a href="#">Internal Wake-Up Flags</a> (MAX77647) section)
<b>4</b>	NOT (Transition 5) Software cold reset (CNFG_GLBL0.SFT_CTRL[1:0] = 0b01) <b>OR</b> Software power-off (CNFG_GLBL0.SFT_CTRL[1:0] = 0b10) <b>OR</b> Software auto wake-up (CNFG_GLBL0.SFT_CTRL = 0b11) <b>OR</b> Watchdog timer expired <b>OR</b> Manual reset occurred (ERCFLAG.MRT = 1)
<b>5A</b>	FPS power-up sequence has not happened yet <b>AND</b> Resources are not forced off <b>AND</b> Internal wake-up flags are set (see the <a href="#">Internal Wake-Up Flags</a> (MAX77647) section)
<b>5B</b>	FPS power-up sequence done
<b>6A</b>	FPS power-up sequence completed <b>AND</b> All resources are force disabled <b>OR</b> Software cold reset (CNFG_GLBL0.SFT_CTRL[1:0] = 0b01) <b>OR</b> Software power-off (CNFG_GLBL0.SFT_CTRL[1:0] = 0b10) <b>OR</b> Software auto wake-up (CNFG_GLBL0.SFT_CTRL = 0b11) <b>OR</b> Watchdog timer expired <b>OR</b> Manual reset occurred (ERCFLAG.MRT = 1) <b>OR</b> CNFG_GLBL1.SBB_F_SHUTDN = 0b1 <b>AND</b> SBB fault occurs
<b>6B</b>	FPS power-down sequence finished

**Internal Wake-Up Flags (MAX77647)**

After transitioning to the shutdown state because of a reset, to allow the device to power up again, internal wake-up flags are set to remember the wake-up request. In [Figure 7](#) and [Table 6](#), these internal wake-up flags trigger transition 3. The internal wake-up flags are set when any of the following happen:

- nEN is debounced (see the [nEN Enable Input \(MAX77647\)](#) section)
  - For example, after a push-button is pressed or a slide-switch switched to HIGH.
- Software cold reset command sent (CNFG\_GLBL0.SFT\_CTRL[1:0] = 0b01)

Reset, Off, and Auto Wake-Up Sequences (MAX77647)

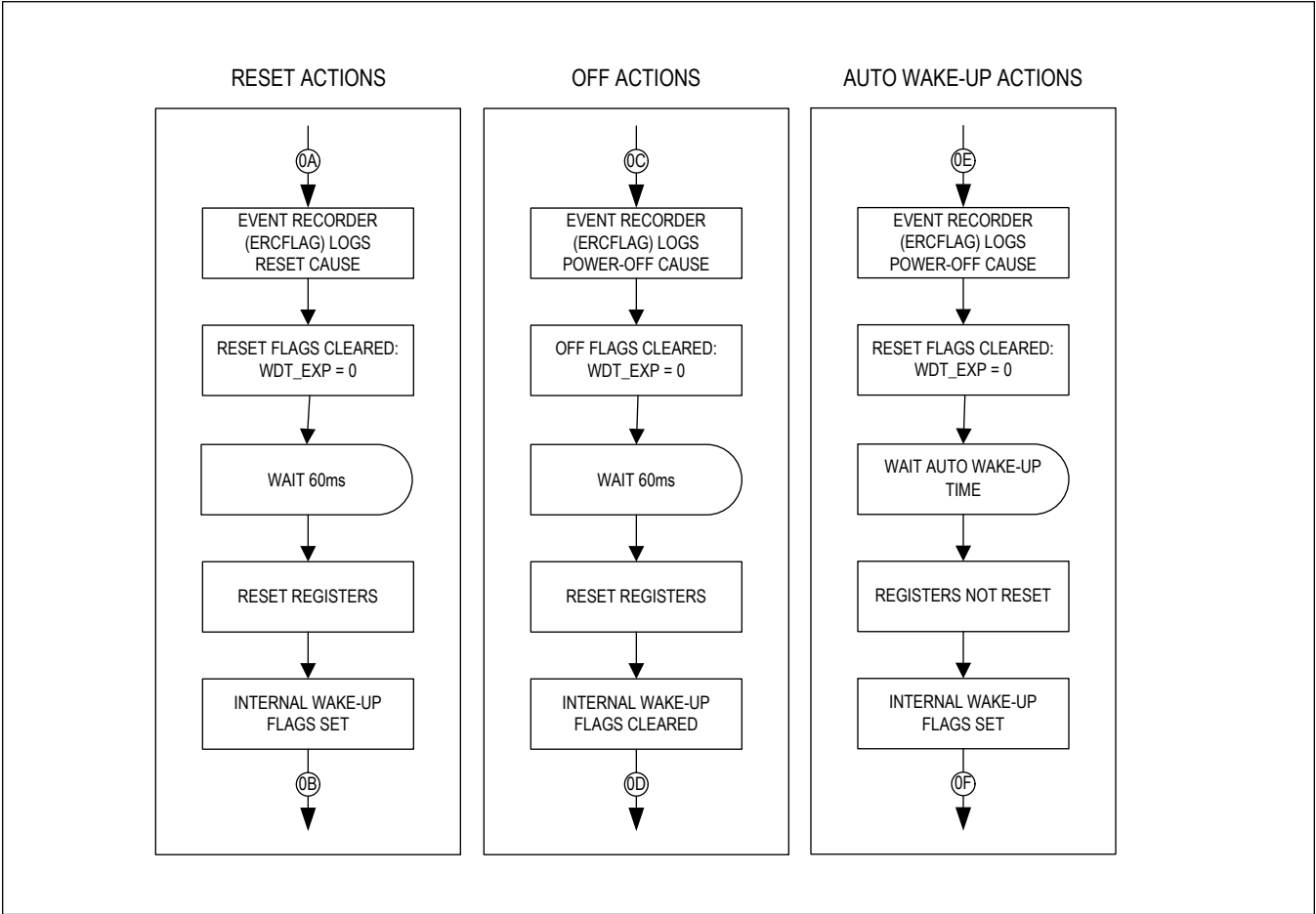


Figure 8. On/Off Controller Reset and Off-Action Sequences



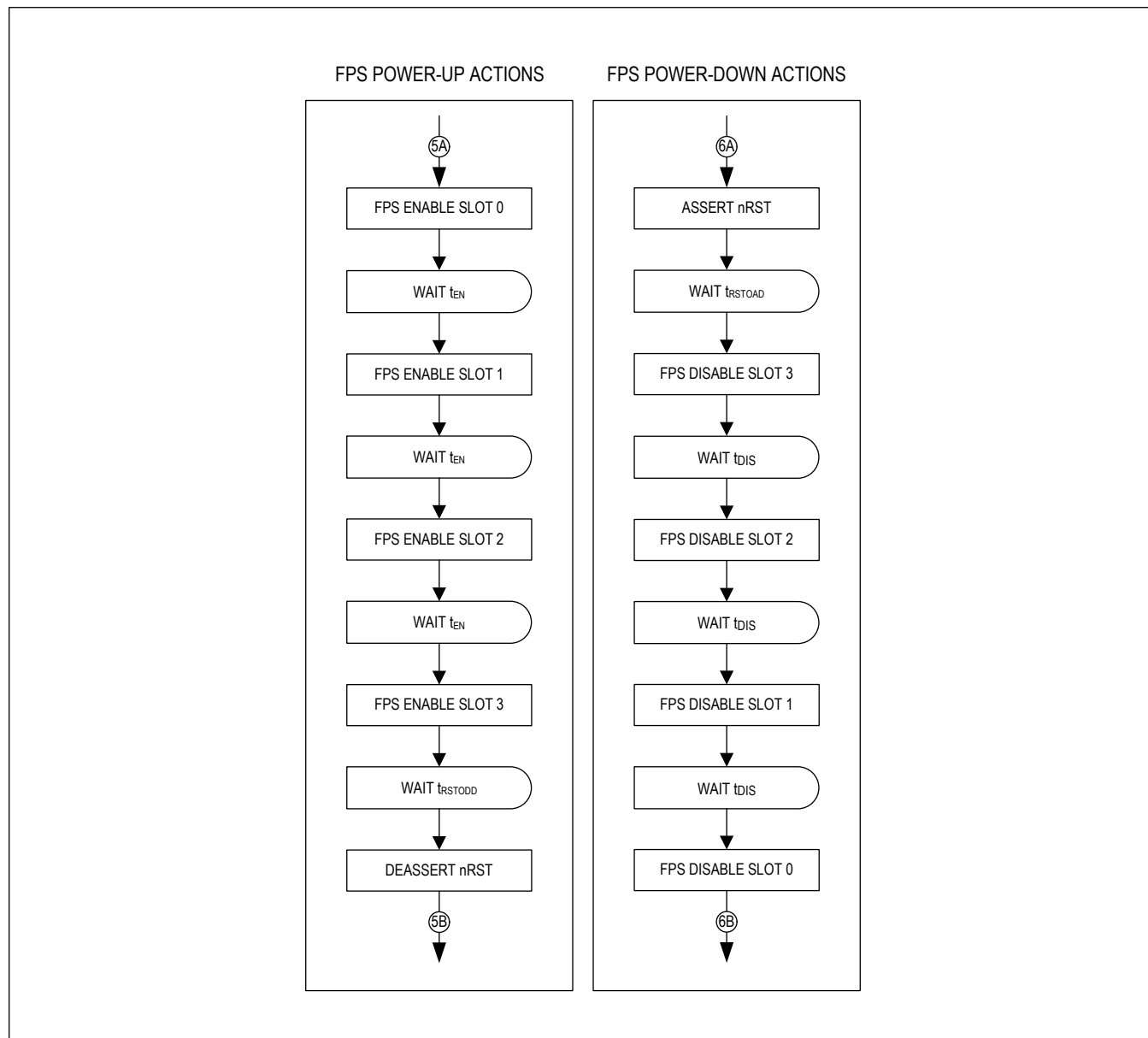
**Power-Up/Down Sequence (MAX77647)**

Figure 9. Power-Up/Down Sequence

**Flexible Power Sequencer (FPS) (MAX77647)**

The FPS allows resources to power up under hardware or software control. Additionally, each resource can power up independently or among a group of other regulators with adjustable power-up/down delays (sequencing). [Figure 10](#) shows four resources powering up under the control of the flexible power sequencer.

The flexible sequencing structure consists of one master sequencing timer and four slave resources (SBB0, SBB1, SBB2 and LDO). When the FPS is enabled, a master timer generates four sequencing events for device power-up/down.

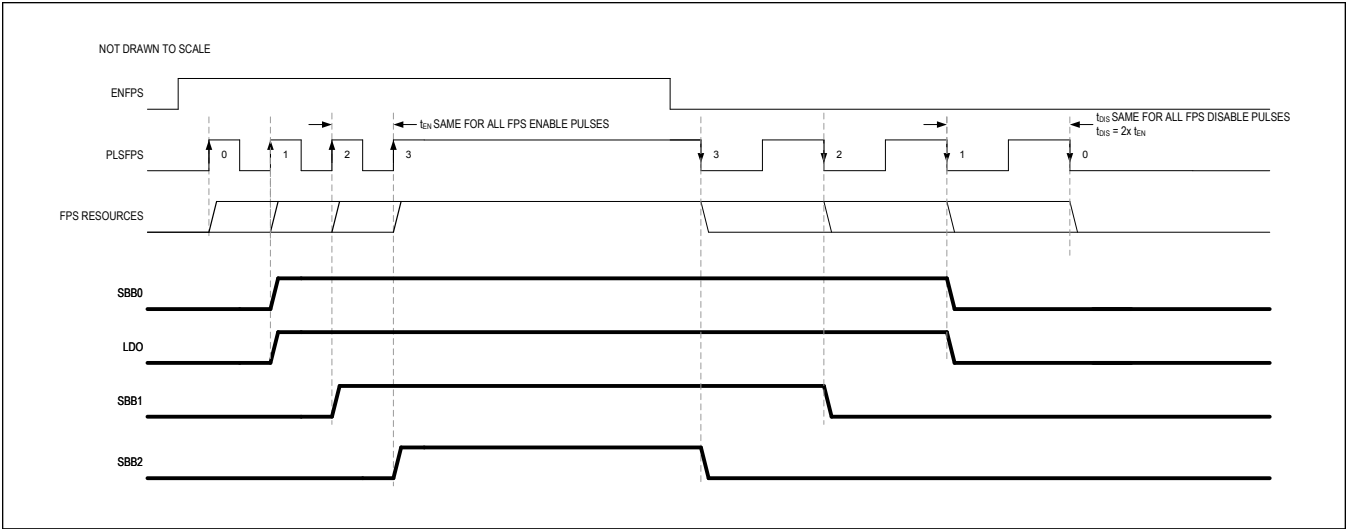


Figure 10. Flexible Power Sequencer Basic Timing Diagram

Startup Timing Diagram Due to nEN (MAX77647)

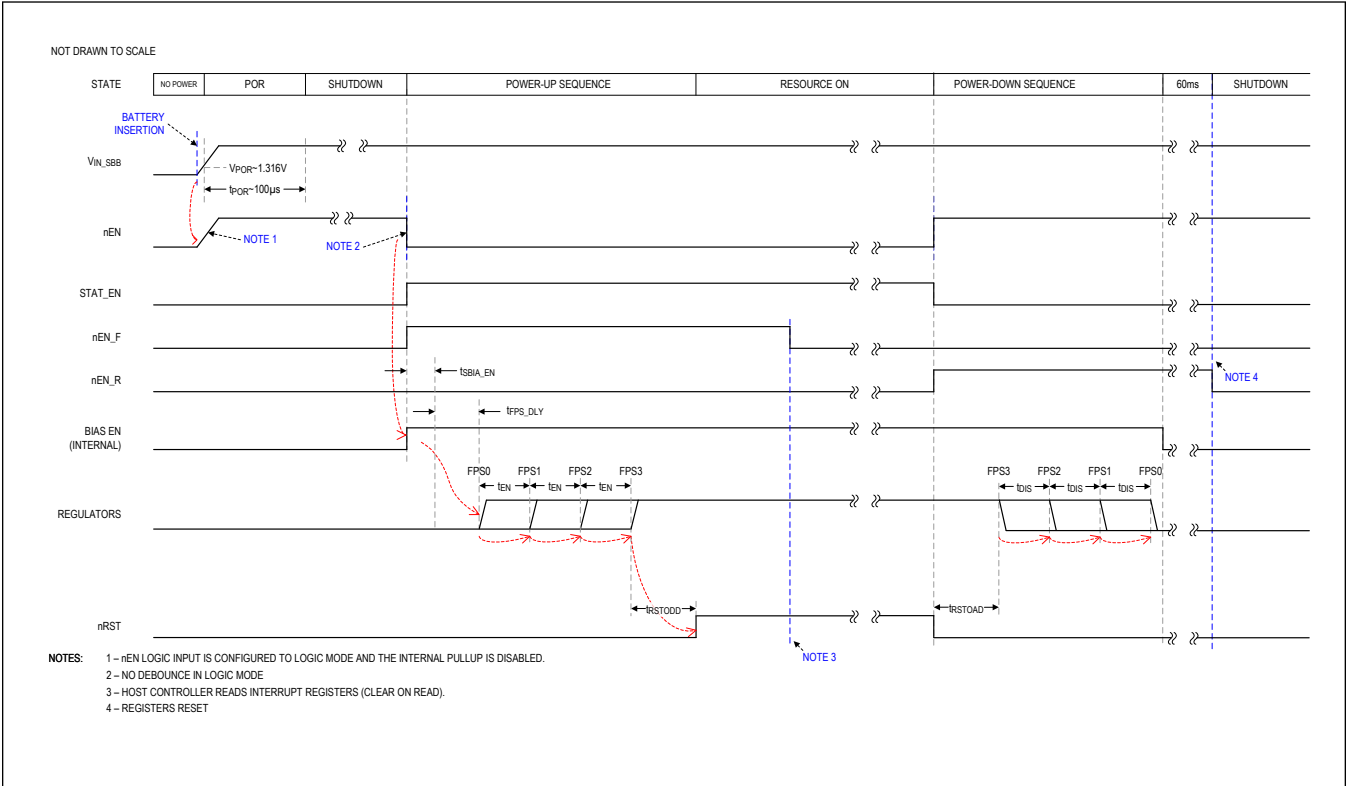


Figure 11. Startup Timing Diagram Due to nEN

**Force Enabled/Disabled Channels (MAX77647)**

Force enable SIMO and LDO output channels by setting CNFG\_SBBx\_B.EN\_SBBx[2:0]. Depending on the OTP, output channels may already be force enabled by default. Output channels configured this way are independent of the flexible power sequence and start up as soon as IN\_SBB (& IN\_LDO) > UVLO rising. The main bias also automatically turns on. Likewise, output channels can be force disabled by setting EN\_SBBx[2:0].

**Debounced Inputs (nEN, GPI) (MAX77647)**

The nEN and GPIO (when operating as an input) have programmable debounce timers on both the rising and falling edges to reject undesired transitions. The input must be at a stable logic level for the entire debounce period for the output to change its logic state. [Figure 12](#) shows an example timing diagram for the nEN debounce.

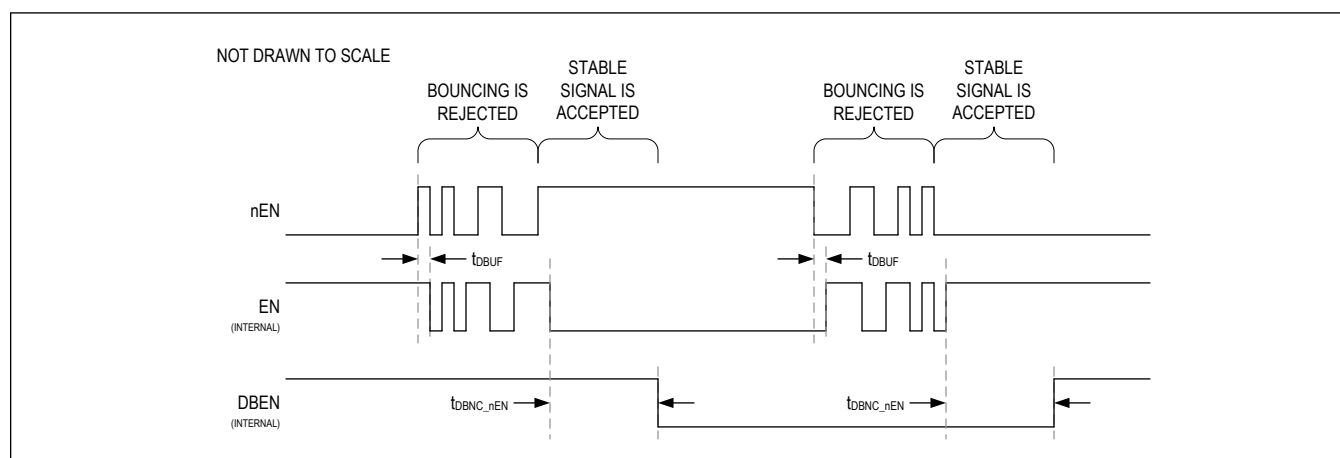


Figure 12. Debounced Inputs

**Watchdog Timer (WDT) (MAX77647)**

The IC features a watchdog timer function for operational safety. If this timer expires without being cleared, the on/off controller causes the IC to enter the shutdown state and resets configuration registers. See the [On/Off Controller](#) and [On/Off Controller Transition Table \(MAX77647\)](#) sections (transitions 0A and 0C) for more details.

Write CNFG\_WDT.WDT\_EN = 1 through the I<sup>2</sup>C interface to enable the timer. The watchdog timer period ( $t_{WDT}$ ) is configurable from 16 seconds to 128 seconds in four steps with CNFG\_WDT.WDT\_PER[1:0]. The default timer period is 128 seconds. While the watchdog timer is enabled, the CNFG\_WDT.WDT\_CLR bit must be set through the I<sup>2</sup>C interface periodically (within  $t_{WDT}$ ) to reset the timer and prevent shutdown. See the [Register Map](#) and [Figure 13](#) for additional details.

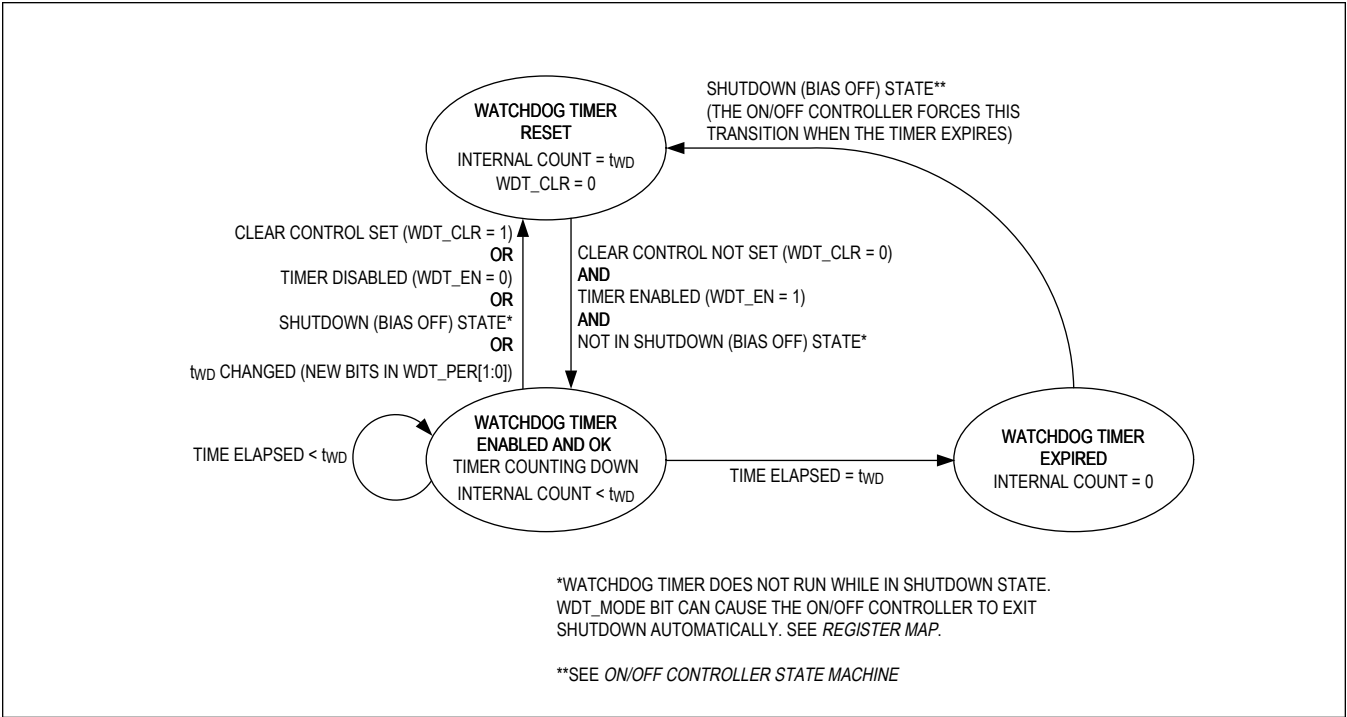


Figure 13. Watchdog Timer State Machine

The timer can be factory-programmed to be enabled by default, disabled by default, or locked from accidental disable. The CNFG\_WDT.WDT\_LOCK bit is read-only and must be configured at the factory. See [Table 7](#) for a full description.

Table 7. Watchdog Timer Factory-Programmed Safety Options

WDT_LOCK	WDT_EN	FUNCTION
0	0	Watchdog timer is disabled by default. Timer can be enabled or disabled by I <sup>2</sup> C writes.
0	1	Watchdog timer is enabled by default. Timer can be enabled or disabled by I <sup>2</sup> C writes.
1	0	Watchdog timer is disabled by default. Timer can be enabled by an I <sup>2</sup> C write, but only a SYSRST can reset the CNFG_WDT.WDT_EN value back to 0. Timer can not be disabled by direct I <sup>2</sup> C writes to CNFG_WDT.WDT_EN (write from 1 → 0 is ignored, write from 0 → 1 is accepted).
1	1	Watchdog timer is enabled by default. Nothing can disable the timer.

### Detailed Description—SIMO Buck-Boost

The device has a micropower single-inductor, multiple-output (SIMO) buck-boost DC-to-DC converter designed for applications that emphasize low supply current and small solution size. A single inductor is used to regulate three separate outputs, saving board space while delivering better total system efficiency than equivalent power solutions using one buck and linear regulators.

The buck-boost configuration utilizes the entire battery voltage range due to its ability to create output voltages that are above, below, or equal to the input voltage. Peak inductor current for each output is programmable to optimize the balance between efficiency, output ripple, EMI, PCB design, and load capability.

To further boost efficiency when the output voltage is always lower than the input, individual channels of the SIMO buck-boost converter can be configured to be in buck-only or boost-only mode, reducing switching losses by toggling less switches compared to buck-boost mode. See the [SIMO Buck Mode \(MAX77647\)](#) and [SIMO Boost Mode \(MAX77647\)](#) sections for more details.

### SIMO Features and Benefits

- 1.8V to 3.3V Input Range
- Three Output Channels
- Ideal for Low-Power Designs
  - Delivers up to 500mA at 1.8V from a 2.4V Input
  - $\pm 2\%$  Accurate Output Voltage
- Small Solution Size
  - Multiple Outputs from a Single 1.5 $\mu$ H Inductor
  - Small 10 $\mu$ F (0402) Output Capacitors
- Flexible and Easy to Use
  - Single Mode of Operation
  - Glitch-free Transitions Between Buck, Boost, and Buck-Boost Modes
  - Programmable Peak Inductor Current
  - Programmable On-Chip Active Discharge
  - Programmable Buck-Only and Boost-Only Mode (MAX77647)
  - Resistor Programmable Output Voltages (MAX77646)
- Long Battery Life
  - High Efficiency, 92% Peak Efficiency
  - Better Total System Efficiency than Buck + LDOs
  - Ultra-Low Quiescent Current, 0.3 $\mu$ A per Channel
  - Low Input Operating Voltage, 1.6V (Falling Threshold)

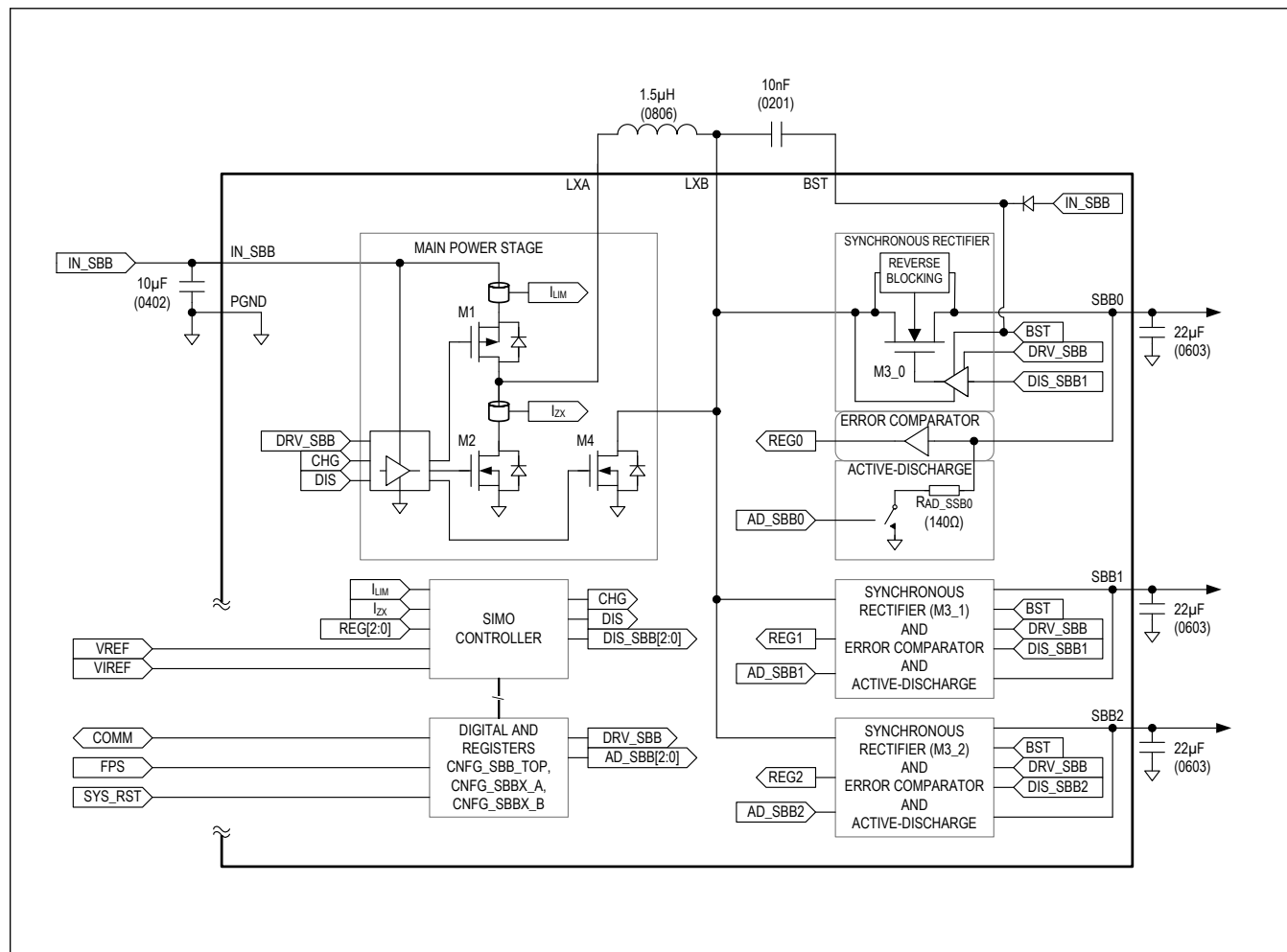
**SIMO Detailed Block Diagram**

Figure 14. SIMO Detailed Block Diagram

**SIMO Control Scheme**

The SIMO buck-boost is designed to service multiple outputs simultaneously. A proprietary controller ensures that all outputs get serviced in a timely manner, even while multiple outputs are contending for the energy stored in the inductor. When no regulator needs service, the state machine rests in a low-power rest state.

When the controller determines that a regulator requires service, it charges the inductor (M1 + M4) until the peak current limit is reached ( $I_{LIM} = IP\_SBB[2:0]$ ). The inductor energy then discharges (M2 + M3\_x) into the output until the current reaches zero ( $I_{ZX}$ ). In the event that multiple output channels need servicing at the same time, the controller ensures that no output utilizes all of the switching cycles. Instead, cycles interleave between all the outputs that are demanding service, while outputs that do not need service are skipped.

**Drive Strength (MAX77647)**

The SIMO regulator's drive strength for its internal power MOSFETs is adjustable using the DRV\_SBB[1:0] bit field. The ideal value is determined experimentally for each application. For a PCB layout comparable to the MAX77646/MAX77647 evaluation kit, 0x1 is the best setting and represents a balance between efficiency and EMI. Faster settings result in higher efficiency but generally require stricter layout rules or shielding to avoid additional EMI. Slower settings limit EMI

in non-ideal settings (e.g., contained layout, antennae adjacent to the device, etc.). Change the drive strength only once during system initialization.

### SIMO Startup Requirements

MAX77646/MAX77647 has special startup requirements for  $V_{OUT} > 3.3V$  with lower input voltage ( $V_{IN} < 2.4V$ ). If operating in these ranges, customers are recommended to change the IPEAK to 1A for normal startup. For more information, see efficiency curves on Typical Operating Characteristics.

### SIMO Fault Indicator

The IC has an SBBx fault shutdown option to protect itself when an SBBx fault is detected. If this option is enabled, then the regulators power down sequentially after an SBBx fault. An SBBx fault occurs when the output voltage falls below 80% of the regulation target. The corresponding SBBx\_F bit remains high until the output voltage rises above 85% of the regulation target.

### SIMO Output Voltage Configuration

Each of the SIMO outputs is independently configurable. To set the output voltages at SBB0/1/2 for the MAX77646, connect the appropriate resistors from RSEL\_SBB0/1/2 to GND as shown in [Table 8](#) and [Table 9](#). The RSEL\_SBB0/1/2 resistors should have 1% or better tolerance. To set the output voltages at SBB0/1/2 for the MAX77647, use the I<sup>2</sup>C interface to load the configuration registers CNFG\_SBBx\_A.TV\_SBBx[7:0]. This 8-bit configuration is a linear transfer function that starts at 0.5V and ends at 5.1V with 25mV increments and sets the output voltage as  $SBBx = 0.5V + 25mV \times TV\_SBBx[7:0]$  (decimal).

**Table 8. SBB0 Output Voltage Settings**

RSEL_SBB0 (kΩ)	OUTPUT VOLTAGE— SBB0 (V)
0.001	0.50
4.99	0.60
6.00	0.70
7.15	0.80
8.45	0.90
10.0	1.00
11.8	1.05
14.0	1.10
16.9	1.20
20.0	1.30
23.7	1.40
28.0	1.60
34.0	1.75
40.2	1.80
47.5	1.85
56.2	2.00
66.5	2.20
80.6	2.40
95.76	2.80
113.0	3.00
133.0	3.20
162.0	3.25
191.0	3.30

**Table 8. SBB0 Output Voltage Settings (continued)**

RSEL_SBB0 (kΩ)	OUTPUT VOLTAGE— SBB0 (V)
226.0	3.35
267.0	3.40
324.0	3.80
383.0	4.00
452.0	4.20
536.0	4.40
634.0	4.60
768.0	5.00

**Table 9. SBB1/2 Output Voltage Settings**

RSEL_SBB1/2 (kΩ)	OUTPUT VOLTAGE— SBB1/2 (V)
0.001	0.50
4.99	0.60
6.00	0.70
7.15	0.80
8.45	0.90
10.0	1.00
11.8	1.10
14.0	1.20
16.9	1.30
20.0	1.40
23.7	1.60
28.0	1.80
34.0	2.00
40.2	2.20
47.5	2.40
56.2	2.60
66.5	2.80
80.6	3.00
95.76	3.20
113.0	3.30
133.0	3.40
162.0	3.60
191.0	3.80
226.0	4.00
267.0	4.20
324.0	4.40
383.0	4.60
452.0	4.80
536.0	5.00



### Peak Current Configuration

The peak inductor current limit corresponding to each SIMO output is independently configurable. To set the inductor peak current for the MAX77646, connect the appropriate resistors from RSEL\_IPK to GND, as shown in [Table 10](#). The RSEL\_IPK resistors should have a 1% or better tolerance. To set the inductor peak current for the MAX77647, use the I<sup>2</sup>C interface to load the [configuration registers](#).

**Table 10. Inductor Peak Current Setting**

RSEL_IPK (kΩ)	SBB_PK2 (A)	SBB_PK1 (A)	SBB_PK0 (A)
20	0.50	0.50	0.50
28	0.50	0.50	0.70
34	0.70	0.50	0.70
47.5	0.50	0.50	0.91
56.2	0.50	0.91	0.91
66.5	0.91	0.91	0.91

### SIMO Soft-Start

The soft-start feature of the SIMO limits inrush current during startup. The soft-start feature is achieved by limiting the slew rate of the output voltage during startup (dV/dt<sub>SS</sub>).

More output capacitance results in higher input current surges during startup. The following equations and example describe the input current surge phenomenon during startup.

In buck-boost mode, the current into the output capacitor (I<sub>CSBB</sub>) during soft-start is:

$$I_{CSBB} = C_{SBB} \times \frac{dV}{dt_{SS}} \quad \left( \text{Equation 1} \right)$$

where:

- C<sub>SBB</sub> is the capacitance on the output of the regulator
- dV/dt<sub>SS</sub> is the voltage change rate of the output

The input current (I<sub>IN</sub>) during soft-start is:

$$I_{IN} = \frac{(I_{CSBB} + I_{LOAD}) \frac{V_{SBBx}}{V_{IN}}}{\xi} \quad \left( \text{Equation 2} \right)$$

where:

- I<sub>CSBB</sub> is calculated using Equation 1
- I<sub>LOAD</sub> is current consumed from the external load
- V<sub>SBBx</sub> is the output voltage
- V<sub>IN</sub> is the input voltage
- ξ is the efficiency of the regulator

For example, given the following conditions, the peak input current ( $I_{IN}$ ) during soft-start is ~79mA:

Given:

- $V_{IN}$  is 2.4V
- $V_{SBB2}$  is 3.3V
- $C_{SBB2} = 10\mu\text{F}$
- $dV/dt_{SS} = 5\text{mV}/\mu\text{s}$
- $R_{LOAD2} = 330\Omega$  ( $I_{LOAD2} = 3.3\text{V}/330\Omega = 10\text{mA}$ )
- $\xi$  is 80%

Calculation:

- $I_{CSBB} = 10\mu\text{F} \times 5\text{mV}/\mu\text{s}$  (from Equation 1)
- $I_{CSBB} = 50\text{mA}$
- $I_{IN} = \frac{(50\text{mA} + 10\text{mA}) \frac{3.3\text{V}}{2.4\text{V}}}{0.80}$  (from Equation 1)
- $I_{IN} \sim 79\text{mA}$

### SIMO Registers (MAX77647)

Each SIMO buck-boost channel has a dedicated register to program its target output voltage (CNFG\_SBBx\_A.TV\_SBBx[7:0]) and its peak current limit (CNFG\_SBBx\_B.IP\_SBBx[2:0]). Additional controls are available for enabling/disabling the active-discharge resistors (CNFG\_SBBx\_C.ADE\_SBBx), buck-only, boost-only, buck-boost only, and automatic mode (CNFG\_SBBx\_B.OP\_MODE), as well as enabling/disabling the SIMO buck-boost channels (CNFG\_SBBx\_B.EN\_SBBx[2:0]) and individual enabling/disabling of the SIMO low-power mode of operation. For a full description of bits, registers, default values, and reset conditions, see the [Register Map](#).

### SIMO Active-Discharge Resistance

Each SIMO buck-boost channel has an active-discharge resistor ( $R_{AD\_SBBx}$ ) that is automatically enabled/disabled based on a CNFG\_SBBx\_C.ADE\_SBBx bit and the status of the SIMO regulator. The active-discharge feature may be enabled (CNFG\_SBBx\_C.ADE\_SBBx = 1) or disabled (CNFG\_SBBx\_C.ADE\_SBBx = 0) independently for each SIMO channel. Enabling the active-discharge feature helps ensure a complete and timely power down of all system peripherals. If the active-discharge resistor is enabled by default, then the active-discharge resistor is on whenever  $V_{IN\_SBB}$  is below  $V_{INSBB\_UVLO}$  and above  $V_{POR}$ .

These resistors discharge the output when CNFG\_SBBx\_C.ADE\_SBBx = 1, and their respective SIMO channel is off. If the regulator is forced on through CNFG\_SBBx\_B.EN\_SBBx[2:0], then the resistors do not discharge the output even if the regulator is disabled by the main bias.

Note that when  $V_{IN\_SBB}$  is less than 1.0V, the NMOS transistors that control the active-discharge resistors lose their gate drive and become open.

### SIMO Buck Mode (MAX77647)

If the input voltage at  $IN\_SBB$  never falls below the output voltage of one or more SIMO converter channels, individual channels can be configured to be in buck mode with the CNFG\_SBBx\_B.OP\_MODE bit. In buck mode, when an output needs service, switch M3\_x remains closed and M4 remains open (see [Figure 14](#)). Only M1 and M2 are toggled as in a traditional buck converter. Efficiency is boosted due to three major factors:

- Reduced switching loss: Buck mode toggles only two switches versus the four in buck-boost mode. Therefore, there are fewer switching events during which power is consumed.
- Lower inductor core losses: Inductor current changes from 0A to peak current. The larger the change in current the inductor experiences, the more energy is lost in the inductor core in the form of heat. In buck mode, the peak current can be reduced since less inductor current is needed to support a load. Less inductor current is needed because of direct energy transfer. Direct energy transfer occurs while the inductor is charged when the input ( $IN\_SBB$ ) is connected directly to the output (SBBx) through the inductor. Therefore, the input not only provides energy to charge the inductor but energy is also supplied to the output capacitor and load devices. Therefore, less current is needed to charge the inductor, which is used to charge the output capacitor in the next switching state.

- Less frequent charging cycles: In buck-boost mode, the output capacitor is charged only while the inductor is being discharged. Again because of direct energy transfer, the output capacitor is charged during both the inductor charge and discharge times. In addition, with the same peak current limit, the inductor charge time is longer with buck mode. Therefore, the output capacitor can support the output voltage longer before needing to be recharged.

Maintain a minimum headroom of 0.7V between IN\_SBB and SBBx in buck mode because inductor charge time ( $dt = L \times I_{P\_SBBx} / (V_{IN\_SBB} - V_{SBBx})$ ) increases as the difference between the IN\_SBB and SBBx voltages shrinks. As the inductor current takes longer to reach its peak, the output voltage might take too long to reach its target voltage, and the MAX77647 might trigger a fault flag.

### SIMO Boost Mode (MAX77647)

If the input voltage at IN\_SBB never rises above the output voltage of one or more SIMO converter channels, individual channels can be configured to be in boost mode with the CNFG\_SBBx\_B.OP\_MODE bit. In boost mode, when an output needs service, switch M1 remains closed and M2 remains open (see [Figure 14](#)). Only M3x and M4 are toggled as in a traditional boost converter. Efficiency is boosted due to three major factors:

- Reduced switching loss: Boost mode toggles only two switches versus the four in buck-boost mode. Therefore, there are less switching events during which power is consumed.
- Lower inductor core losses: Inductor current changes from 0A to peak current. The larger the change in current the inductor experiences, the more energy is lost in the inductor core in the form of heat. In boost mode, the peak current can be reduced since less inductor current is needed to support a load. Less inductor current is needed because of direct energy transfer. Direct energy transfer occurs while the inductor is discharged when the input (IN\_SBB) is connected directly to the output (SBBx) through the inductor. Therefore, the input not only provides energy to charge the inductor but energy is also supplied to the output capacitor and load devices. Therefore, less current is needed to charge the inductor, which is used to charge the output capacitor in the next switching state.
- Less frequent charging cycles: The inductor discharge time is longer with boost mode. Therefore, the output capacitor can support the output voltage longer before needing to be recharged.

## Applications Information

### SIMO Available Output Current

The available output current on a given SIMO channel is a function of the input voltage, output voltage, peak current limit setting, and output current of the other SIMO channels. Analog Devices offers a calculator found in the Design Resources tab of the MAX77646/MAX77647's webpage that outlines the available capacity for specific conditions. [Table 11](#) is an extraction from the calculator.

**Table 11. SIMO Available Output Current for Common Applications**

PARAMETERS	EXAMPLE 1	EXAMPLE 2	EXAMPLE 3
V <sub>IN_SBB</sub>	1.8V	2.4V	3.3V
R <sub>L_DCR</sub>	120mΩ	120mΩ	120mΩ
SBB0	1.0V at 80mA	1.2V at 50mA	1.2V at 20mA
SBB1	1.2V at 50mA	1.8V at 100mA	1.8V at 80mA
SBB2	1.8V at 40mA	3.3V at 30mA	3.3V at 40mA
Operating Mode	Automatic	Automatic	Automatic
IP_SBB0	0.91A	0.91A	0.50A
IP_SBB1	0.91A	0.70A	0.50A
IP_SBB2	0.91A	0.91A	0.50A
Utilized Capacity	75.2%	80.4%	73.5%

\* $ESR_{C\_IN} = ESR_{C\_OUT} = 5m\Omega$ ,  $L = 1.5\mu H$

**Inductor Selection**

Choose an inductance from 1.0μH to 2.2μH; 1.5μH inductors work best for most designs. Larger inductances transfer more energy to the output for each cycle and typically result in a larger output voltage ripple and better efficiency. See the [Output Capacitor Selection](#) section for more information on how to size your output capacitor to control ripple.

Choose the inductor saturation current to be greater than or equal to the maximum peak current limit setting that is used for all of the SIMO buck-boost channels ( $I_{P\_SBBx}$ ). For example, if SBB0 is set for 0.5A, SBB1 is set for 0.7A, and SBB2 is set for 0.91A, then choose the saturation current to be greater than or equal to 1.0A.

Choose the RMS current rating of the inductor (typically the current at which the temperature rises appreciably) based on the expected load currents for the system. For systems where the expected load currents are not well known, be conservative and choose the RMS current to be greater than or equal to half the higher maximum peak current limit setting [ $I_{RMS} \geq \text{MAX}(I_{P\_SBB0}, I_{P\_SBB1}, I_{P\_SBB2}) / 2$ ]. This is a conservative choice because the SIMO buck-boost regulator implements a discontinuous conduction mode (DCM) control scheme, which returns the inductor current to zero for each cycle.

Consider the DC-resistance (DCR), AC-resistance (ACR), and solution size of the inductor. Typically, smaller-sized inductors have larger DC-resistance and larger AC-resistance that reduces efficiency and the available output current. Note that many inductor manufacturers have inductor families which contain different versions of core material to balance trade-offs between DCR, ACR (i.e., core losses), and component cost. For this SIMO regulator, inductors with the lowest ACR in the 1.0MHz to 2.0MHz region tend to provide the best efficiency.

**Input Capacitor Selection**

Choose the input bypass capacitance ( $C_{IN\_SBB}$ ) to be 10μF. Larger values of  $C_{IN\_SBB}$  improve the decoupling for the SIMO regulator.

The  $C_{IN\_SBB}$  reduces the current peaks drawn from the battery or input power source during SIMO regulator operation and reduces switching noise in the system. The ESR/ESL of the input capacitor should be very low (i.e.,  $ESR \leq 5\text{m}\Omega$  and  $ESL \leq 500\text{pH}$ ) for frequencies up to 2MHz. Ceramic capacitors with X5R or X7R dielectric are highly recommended due to their small size, low ESR, and small temperature coefficients.

To fully utilize the available input voltage range of the SIMO (3.3V, max), use a capacitor with a voltage rating of 6.3V at minimum.

**Boost Capacitor Selection**

Choose the boost capacitance ( $C_{BST}$ ) to be 10nF. Smaller values of  $C_{BST}$  (< 1nF) result in insufficient gate drive for M3. Larger values of  $C_{BST}$  (> 10nF) have the potential to degrade the startup performance. Ceramic capacitors with 0201 or 0402 case sizes are recommended.

**Output Capacitor Selection**

Choose each output bypass capacitance ( $C_{SBBx}$ ) based on the target output voltage ripple ( $\Delta V_{SBBx}$ ). Typical values are 22μF. Larger values of  $C_{SBBx}$  improve the output voltage ripple but increase the input surge currents during soft-start and output voltage changes. The output voltage ripple is a function of the inductance (L), the output voltage ( $V_{SBBx}$ ), and the peak current limit setting ( $I_{P\_SBBx}$ ). See Equation 3 to estimate the required, effective capacitance.

$$C_{SBBx} = \frac{I_{P\_SBBx}^2 \times L}{2 \times V_{SBBx} \times \Delta V_{SBBx}} \quad (\text{Equation 3})$$

Analog Devices also offers a calculator to aid in the selection of the output capacitance found in the Design Resources tab of the MAX77646/MAX77647 Product page. Note that most designs concern themselves with having enough capacitance on the output but there is also a maximum capacitance limitation that is calculated within the SIMO calculator; take care not to exceed the maximum capacitance.

The  $C_{SBBx}$  is required to keep the output voltage ripple small. The impedance of the output capacitor (ESR, ESL) should be very low (i.e.,  $ESR \leq 5\text{m}\Omega$  and  $ESL \leq 500\text{pH}$ ) for frequencies up to 2MHz. Ceramic capacitors with X5R or X7R dielectric are highly recommended due to their small size, low ESR, and small temperature coefficients.

A capacitor's effective capacitance decreases with increased DC bias voltage. This effect is more pronounced as capacitor case sizes decrease. Due to this characteristic, an 0603 case-size capacitor can perform well, while an 0402

case-size capacitor of the same value performs poorly. The SIMO regulator is stable with low output capacitance (1 $\mu$ F) but the output voltage ripple would be large; consider the effective output capacitance value after initial tolerance, bias voltage, aging, and temperature derating.

### SIMO Switching Frequency

The SIMO buck-boost regulator uses a pulse frequency modulation (PFM) control scheme. The switching frequency for each output is a function of the operating mode, input voltage, output voltage, load current, and inductance. Output capacitance is a minor factor in SIMO switching frequency. Analog Devices offers a SIMO calculator found in the Design Resources tab of the MAX77646/MAX77647 Product page to estimate expected switching frequency.

[Table 12](#) lists how different factors increase or decrease switching frequency.

**Table 12. Switching Frequency Control**

FACTOR	INCREASING FREQUENCY	DECREASING FREQUENCY
Inductor Current Peak Limit	Lower Peak Limit	Higher Peak Limit
Operating Mode	Buck-Boost Mode	Buck Mode or Boost Mode
Inductor	Decrease Inductance	Increase Inductance
Output Capacitor	Decrease Capacitance	Increase Capacitance
Input Voltage	Higher Voltage	Lower Voltage
Output Voltage	Higher Voltage	Lower Voltage
Load Current	Higher Current	Lower Current

### Unused Outputs

Do not leave unused outputs unconnected. If an output left unconnected is accidentally enabled, the charged inductor experiences an open circuit, and the output voltage soars above the absolute maximum rating, damaging the device. If an output is not used, do one of the following:

1. Disable the output (CNFG\_SBBx\_B.EN\_SBBx[2:0] = 0x4 or 0x5) and connect the output to ground. If an unused output is default enabled or can be accidentally enabled, do one of the other recommendations instead.
2. Bypass the unused output with a 1 $\mu$ F capacitor to ground.
3. Connect the unused output to IN\_SBB or a different output channel if the unused output is programmed to a lower voltage. Since the output voltage is higher than the unused output, the regulator does not service the unused output even if it is unintentionally enabled.
  - Note that some OTP options have the active-discharge resistors enabled by default. Connecting an unused output to IN\_SBB is **not recommended** if the active discharge is enabled by default. If connecting the unused output to a different channel, disable the active-discharge resistor (CNFG\_SBBx\_B.ADE\_SBBx = 0) of the unused channel.

## PCB Layout Guide

### Capacitors

Place decoupling capacitors as close as possible to the IC such that connections from capacitor pads to pin and from capacitor pads to ground pins are short. Keeping the connections short lowers parasitic inductance and resistance, improving performance and shrinking the physical size of hot loops.

If connections to the capacitors are through vias, use multiple vias to minimize parasitics. Also, connect loads to the capacitor pads rather than the device pins.

The most critical are the capacitors for the switching regulator: input capacitors at IN\_SBB and output capacitors at SBBx.

### Input Capacitor at IN\_SBB

Minimize the parasitic inductance from PGND to input capacitor to IN\_SBB to reduce ringing on the LXA voltage.

### Output Capacitors at SBBx

The output capacitors experience large changes in current as the regulator charges (buck mode) and discharges (both modes) the inductor. In buck mode, the capacitor current ramps up at the same rate as mentioned in the [Input Capacitor at IN\\_SBB](#) section. In buck-boost mode, the capacitor current ramps up very quickly. In both modes, the capacitor current ramps down at a rate of  $\frac{dI_{C\_SBBx}}{dt} = \frac{V_{SBBx}}{L}$  from the inductor peak current. Since the ramp down can occur in less than 1μs, and the current increases rapidly for buck-boost mode, minimize parasitic inductance from SBBx to output capacitor to PGND.

### Inductor

Keep the inductor close to the IC to reduce trace resistance; however, prioritize any regulator input/output capacitors over the inductor. Use the appropriate trace width from LXA to the inductor to LXB to support the peak inductor current. Likewise, if there are vias in the path, use an appropriate amount of vias to support the peak current.

### Ground Connections

As the switching regulator charges and discharges the inductor, current flows from PGND to the input capacitor ground, from the output capacitor ground to PGND, or from the output capacitor ground to the input capacitor ground. Therefore, use a wide, continuous copper plane to connect PGND to the capacitor grounds.

When connecting the GND and PGND pins together, ensure noise from the power ground does not enter the analog ground (where GND is connected). For example, assuming the ground pins are connected through a solid ground plane on an internal layer, one via connecting GND to the internal ground plane may be sufficient to protect GND from most of the noise in the power-ground plane. Likewise, if there are other higher currents or noisy circuitry near this device, avoid connecting the GND pin directly to their grounds.

For more guidelines on proper grounding, visit <https://www.analog.com/en/design-center/package-quality-symbols-footprints/package-resources.html>.

**Example PCB Layout**

[Figure 15](#) shows an example layout of the top layer.

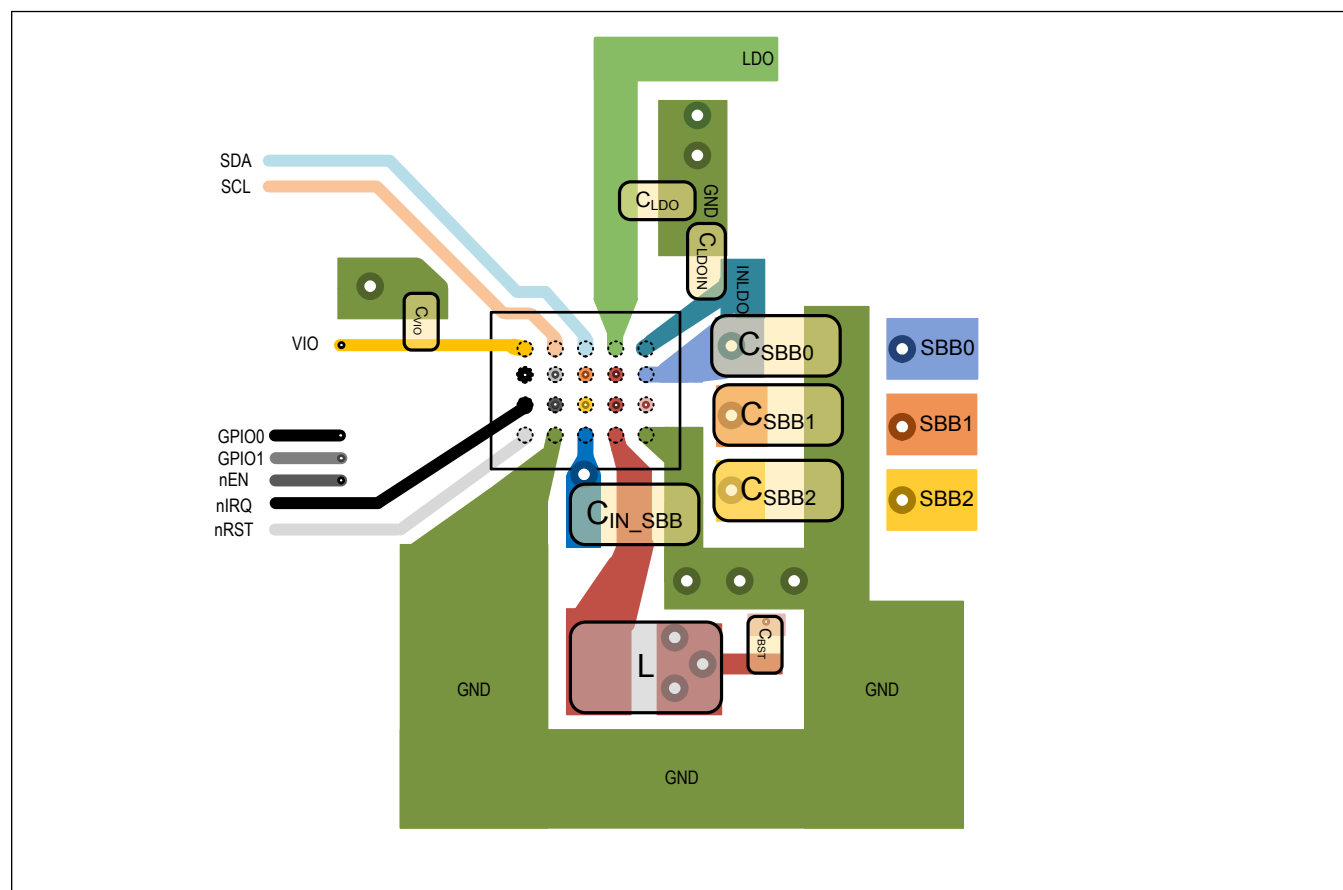


Figure 15. PCB Top-Layer and Component Placement Example

## Detailed Description—Low Dropout Linear Regulator (LDO)/Load Switch (LSW)

The device includes an on-chip low-dropout linear regulator (LDO) that can also be configured as a load switch. The LDO is optimized to have a low-quiescent current. The input voltage range ( $V_{IN\_LDO}$ ) allows it to be powered directly from the main energy source such as a LiMnO<sub>2</sub> (coin cell) or an intermediate regulator. The linear regulator delivers up to 150mA.

### Features and Benefits

- 1x 150mA LDO
- LDO Input Voltage Range: 1.6V to 5.5V
- LSW Input Voltage Range: 1.10V to 5.5V
- Resistor Adjustable Output Voltage (MAX77646)
- I<sup>2</sup>C Adjustable Output Voltage (MAX77647)
- 100mV Maximum Dropout Voltage at ECT Conditions
- Programmable On-Chip Active Discharge

### LDO/LSW Simplified Block Diagram

The LDO/LSW block has one input (IN\_LDO) and one output (LDO) and several ports that exchange information with the rest of the device ( $V_{REF}$ , ADE\_LDO). The  $V_{REF}$  comes from the main bias circuits. The CNFG\_LDO\_B.EN\_LDO and CNFG\_LDO\_B.ADE\_LDO are register bits for controlling the enable and active-discharge features, respectively. See the [Register Map](#) for more information.

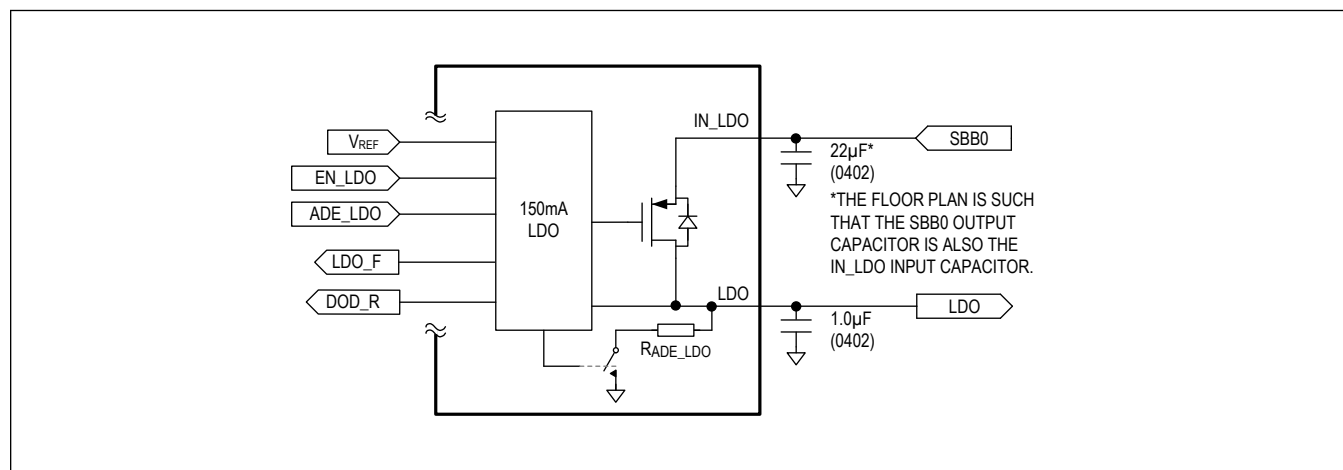


Figure 16. LDO/LSW Simplified Block Diagram

### LDO Output Voltage Configuration

To set the output voltages for the on-chip LDO for the MAX77646, connect the appropriate resistor from RSEL\_LDO to GND, as shown in [Table 13](#). The RSEL\_LDO resistor should have 1% or better tolerance. To set the output voltages for the on-chip LDO for the MAX77647, use the I<sup>2</sup>C interface to load the configuration register TV\_LDO[7:0]. TV\_LDO[7] is used to enable (TV\_LDO[7] = 1) or disabled (TV\_LDO[7] = 0) a 1.325mV offset to the LDO's output voltage. The bits TV\_LDO[6:0] are used to set the output voltage as:

$$\text{LDO} = 0.5\text{V} + \text{TV\_LDO}[7] + (25\text{mV} \times \text{TV\_LDO}[6:0] \text{ (decimal)})$$



**Table 13. LDO Output Voltage Setting**

RSEL_LDO (k $\Omega$ )	LDO OUTPUT VOLTAGE (V)
Short	0.50
4.99	0.60
6.00	0.70
7.15	0.80
8.45	0.90
10.0	0.95
11.8	1.00
14.0	1.05
16.9	1.10
20.0	1.15
23.7	1.20
28.0	1.25
34.0	1.30
40.2	1.40
47.5	1.50
56.2	1.60
66.5	1.70
80.6	1.80
95.76	1.90
113.0	2.00
133.0	2.10
162.0	2.20
191.0	2.40
226.0	2.50
267.0	2.60
324.0	2.80
383.0	3.00
452.0	3.20
536.0	3.40
634.0	3.60
768.0	3.80
909.0	4.00

**LDO/LSW Active-Discharge Resistor**

The LDO/LSW block has an active-discharge resistor ( $R_{AD\_LDO}$ ) that is enabled if  $CNFG\_LDO\_B.ADE\_LDO = 1$  and LDO is disabled. Enabling the active-discharge feature helps ensure a complete and timely power down of the resource. During power up, if  $V_{INLDO} > V_{POR}$  and  $CNFG\_LDO\_B.ADE\_LDO = 1$ , the active-discharge resistor is enabled.

**LDO/LSW Soft-Start**

The soft-start feature limits inrush current during startup and is achieved by limiting the slew rate of the output voltage during startup ( $dV_{OUT\_LDO}/dt_{SS}$ ).

More output capacitance results in higher input current surges during startup. The following equation and example describe the input current surge phenomenon during startup.

The input current ( $I_{IN\_LDO}$ ) during soft-start is calculated as:

$$I_{IN\_LDO} = C_{LDO} \times \frac{dV_{OUT\_LDO}}{dt_{SS}} + I_{OUT\_LDO}$$

where:

- $C_{LDO}$  is the capacitance on the output of the regulator
- $dV_{OUT\_LDO}/dt_{SS}$  is the voltage change rate of the output

For example, given the following conditions, the input current ( $I_{IN\_LDO}$ ) during soft-start is 13.08mA:

Given:

- $C_{LDO} = 2.2\mu F$
- $dV_{OUT\_LDO}/dt_{SS} = 2.2mV/\mu s$
- LDO programmed to 1.85V
- $R_{LDO} = 185\Omega$  ( $I_{OUT\_LDO} = 1.85V/185\Omega = 10mA$ )

Calculation:

- $I_{IN} = 2.2\mu F \times 2.2mV/\mu s + 10mA$
- $I_{IN} = 14.84mA$

**Load Switch Configuration**

The LDO can be configured as a load switch with the `CNFG_LDO_B.LDO_MD` bit. As shown in [Figure 17](#), the transition from LDO to LSW mode is controlled by a defined slew rate until dropout is detected. Once dropout is detected, the load switch is fully closed and the dropout interrupt flag (`INT_GLBL.DOD_R`) is set.

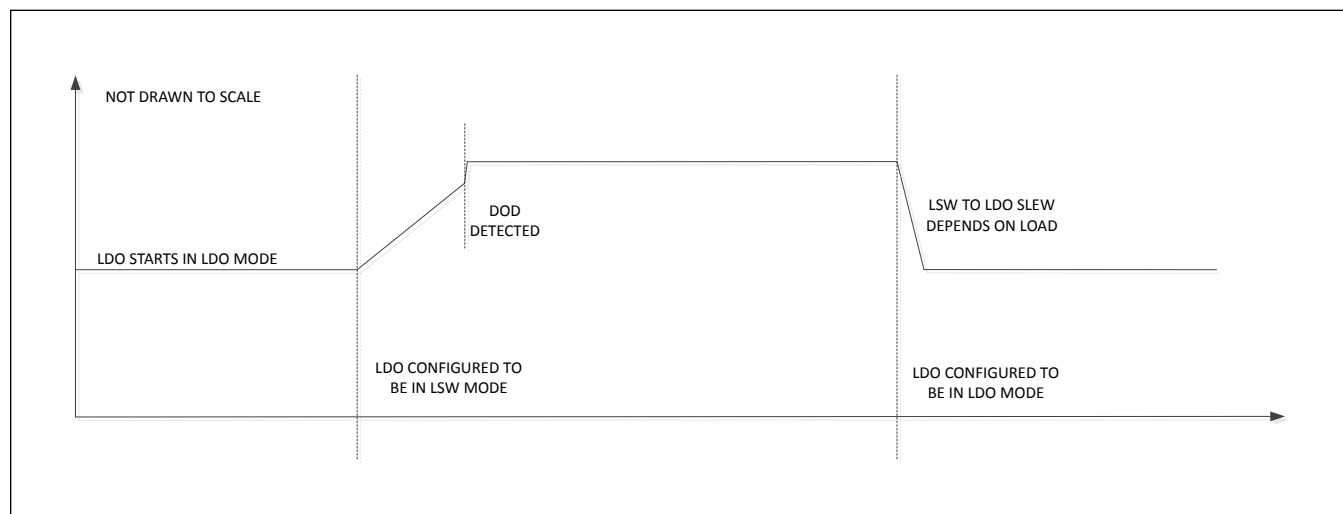


Figure 17. LDO to LSW Transition Waveform

## Applications Information

### Input Capacitor Selection

Make sure the input bypass capacitance ( $C_{IN\_LDO}$ ) is at least 2.2 $\mu$ F. Larger values of  $C_{IN\_LDO}$  improve the decoupling for LDO. The floor plan of the device is such that SBB0 is adjacent to IN\_LDO and if the SIMO channel 0 output powers the input of LDO, then its output capacitor ( $C_{SBB0}$ ) can also serve as  $C_{IN\_LDO}$  such that only one capacitor is required.

The  $C_{IN\_LDO}$  reduces the current peaks drawn from the battery or input power source during operation. The impedance of the input capacitor (ESR, ESL) should be very low (i.e.,  $ESR \leq 50m\Omega$  and  $ESL \leq 5nH$ ) for frequencies up to 0.5MHz. Ceramic capacitors with X5R or X7R dielectric are highly recommended due to their small size, low ESR, and small temperature coefficients.

### Output Capacitor Selection

For both LDO and LSW modes, choose the output bypass capacitance ( $C_{LDO}$ ) to be 1 $\mu$ F.

In LDO mode, larger values of  $C_{LDO}$  improve output PSRR but increase input surge currents during soft-start and output voltage changes. The effective output capacitance should not exceed 2.8 $\mu$ F to maintain stability.

While in LDO mode,  $C_{LDO}$  is required to keep stability. The series inductance of the output capacitor and its series resistance should be low (i.e.,  $ESR \leq 10m\Omega$  and  $ESL \leq 1nH$ ) for frequencies up to 0.5MHz. Ceramic capacitors with X5R or X7R dielectric are highly recommended due to their small size, low ESR, and small temperature coefficients.

A capacitor's effective capacitance decreases with increased DC bias voltage. This effect is more pronounced with smaller capacitor case sizes. Due to this characteristic, 0603 case-size capacitors tend to perform well while 0402 case-size capacitors of the same value perform poorly.

## Detailed Description—I<sup>2</sup>C Serial Communication

### General Description

The IC features a revision 3.0 I<sup>2</sup>C-compatible, 2-wire serial interface consisting of a bidirectional serial data line (SDA) and a serial clock line (SCL). This device acts as a slave-only device, relying on the master to generate a clock signal. SCL clock rates from 0Hz to 1MHz are supported.

The I<sup>2</sup>C is an open-drain bus, and therefore, SDA and SCL require pullups. Optional resistors (24 $\Omega$ ) in series with SDA and SCL protect the device inputs from high-voltage spikes on the bus lines. Series resistors also minimize crosstalk and undershoot on bus signals.

[Figure 18](#) shows the functional diagram for the I<sup>2</sup>C-based communications controller. For additional information on I<sup>2</sup>C, refer to the *I<sup>2</sup>C Bus Specification and User Manual* which is available for free on the internet.

### Features

- I<sup>2</sup>C Revision 3.0 Compatible Serial Communications Channel
- Compatible with any Bus Timing up to 1000kHz
- Does not utilize I<sup>2</sup>C Clock Stretching

## I<sup>2</sup>C Simplified Block Diagram

There are three pins (aside from GND) for the I<sup>2</sup>C-compatible interface. The  $V_{IO}$  determines the logic level, SCL is the clock line, and SDA is the data line. Note that the interface does **not** have the ability to drive the SCL line.

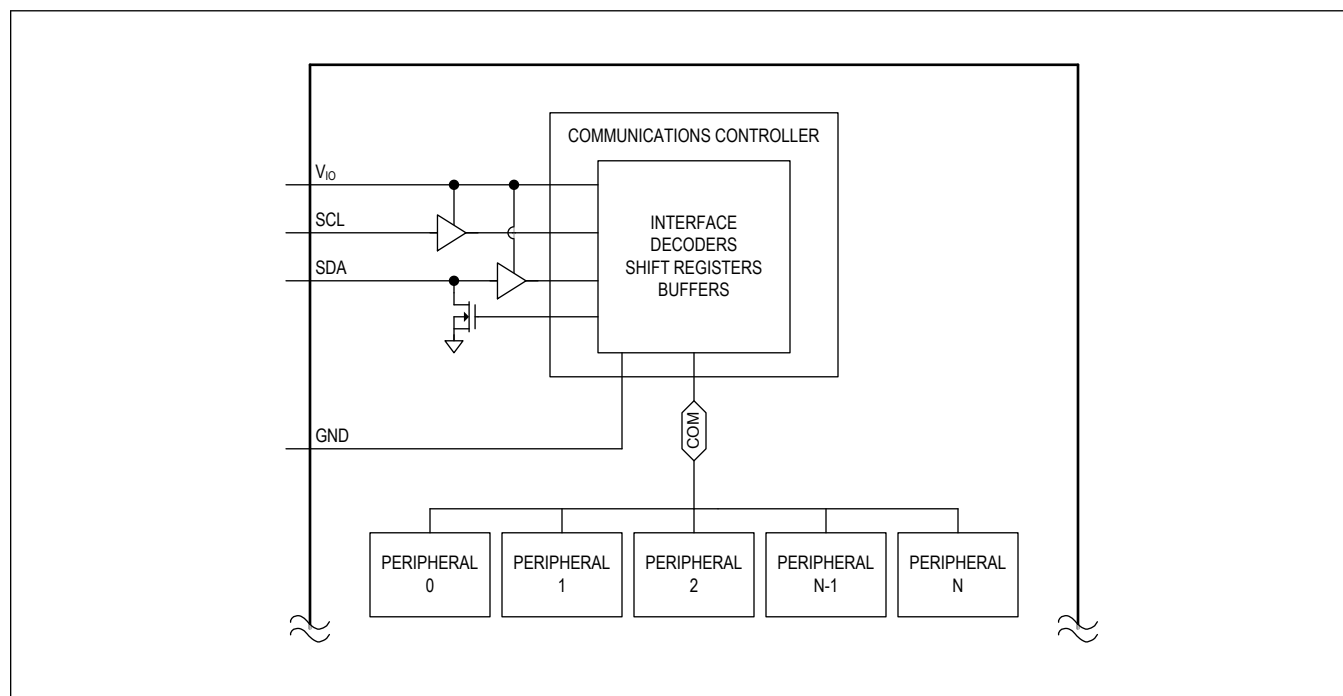


Figure 18. I<sup>2</sup>C Simplified Block Diagram

## I<sup>2</sup>C System Configuration

The I<sup>2</sup>C-compatible interface is a multimaster bus. The maximum number of devices that can attach to the bus is only limited by bus capacitance.

A device on the I<sup>2</sup>C bus that sends data to the bus is called a transmitter. A device that receives data from the bus is called a receiver. The device that initiates a data transfer and generates the SCL clock signals to control the data transfer is a master. Any device that is being addressed by the master is considered a slave. The I<sup>2</sup>C-compatible interface operates as a slave on the I<sup>2</sup>C bus with transmit and receive capabilities.

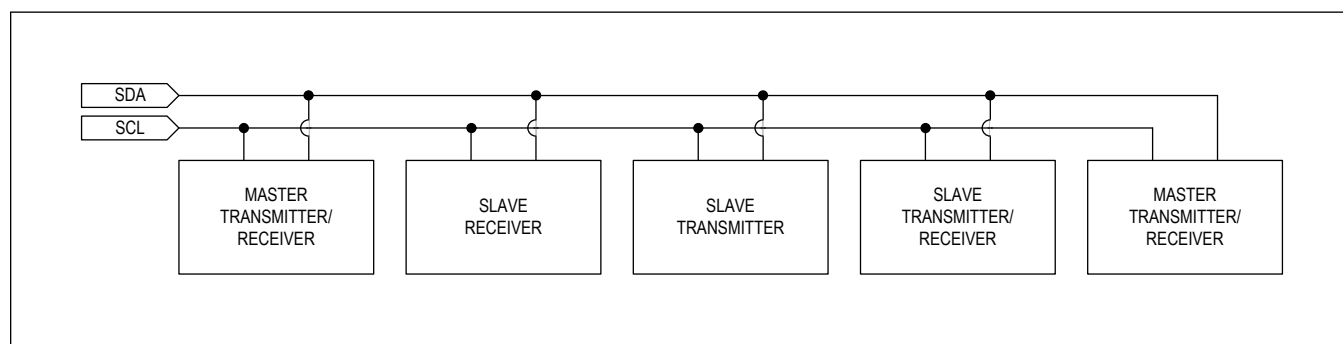


Figure 19. I<sup>2</sup>C System Configuration

## I<sup>2</sup>C Interface Power

The I<sup>2</sup>C interface derives its power from V<sub>IO</sub>. Typically a power input such as V<sub>IO</sub> would require a local 0.1μF ceramic bypass capacitor to ground. However, in highly-integrated power distribution systems, a dedicated capacitor might not be necessary. If the impedance between V<sub>IO</sub> and the next closest capacitor ( $\geq 0.1\mu\text{F}$ ) is less than 100mΩ in series with 10nH, then a local capacitor is not needed. Otherwise, bypass V<sub>IO</sub> to GND with a 0.1μF ceramic capacitor.

The V<sub>IO</sub> accepts voltages from 1.7V to 3.6V (V<sub>IO</sub>). Cycling V<sub>IO</sub> does not reset the I<sup>2</sup>C registers. When V<sub>IO</sub> is less than V<sub>IOUVLO</sub> and V<sub>SYSA</sub> is less than V<sub>SYSAUVLO</sub>, SDA and SCL are high impedance.

## I<sup>2</sup>C Data Transfer

One data bit is transferred during each SCL clock cycle. The data on SDA must remain stable during the high period of the SCL clock pulse. Changes in SDA, while SCL is high, are control signals. See the [I<sup>2</sup>C Start and Stop Conditions](#) section. Each transmit sequence is framed by a START (S) condition and a STOP (P) condition. Each data packet is 9 bits long: 8 bits of data followed by the acknowledge bit. Data is transferred with the MSB first.

## I<sup>2</sup>C Start and Stop Conditions

When the serial interface is inactive, SDA and SCL idle high. A master device initiates communication by issuing a START condition. A START condition is a high-to-low transition on SDA with SCL high. A STOP condition is a low-to-high transition on SDA, while SCL is high. See [Figure 20](#).

A START condition from the master signals the beginning of a transmission to the device. The master terminates transmission by issuing a not-acknowledge followed by a STOP condition (see the [I<sup>2</sup>C Acknowledge Bit](#) section for information on not-acknowledge). The STOP condition frees the bus. To issue a series of commands to the slave, the master can issue repeated start (Sr) commands instead of a STOP command to maintain control of the bus. In general, a repeated start command is functionally equivalent to a regular start command.

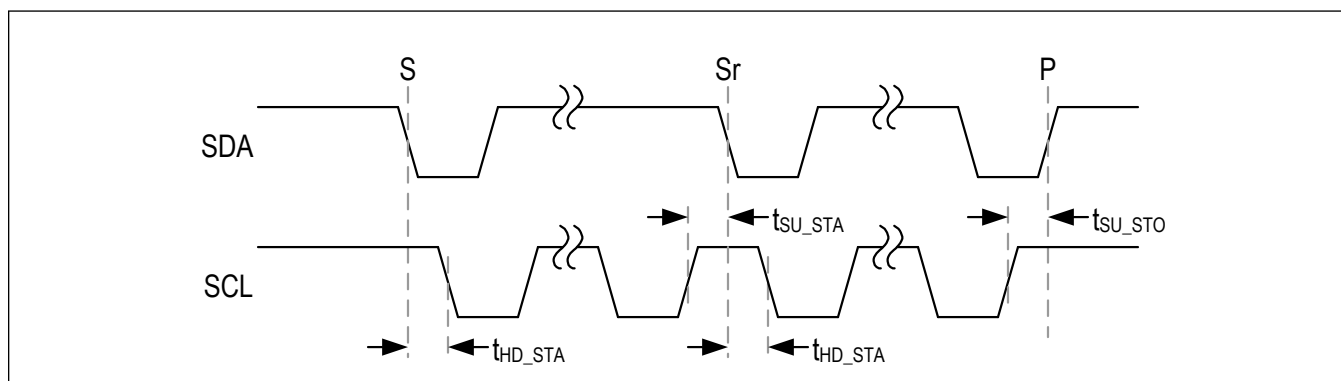


Figure 20. I<sup>2</sup>C Start and Stop Conditions

## I<sup>2</sup>C Acknowledge Bit

Both the I<sup>2</sup>C bus master and slave devices generate acknowledge bits when receiving data. The acknowledge bit is the last bit of each 9-bit data packet. To generate an acknowledge (A), the receiving device must pull SDA low before the rising edge of the acknowledge-related clock pulse (ninth pulse) and keep it low during the high period of the clock pulse. See [Figure 21](#). To generate a not-acknowledge (nA), the receiving device allows SDA to be pulled high before the rising edge of the acknowledge-related clock pulse and leaves it high during the high period of the clock pulse.

Monitoring the acknowledge bits allows for the detection of unsuccessful data transfers. An unsuccessful data transfer occurs if a receiving device is busy or if a system fault has occurred. In the event of an unsuccessful data transfer, the bus master should reattempt communication at a later time.

This device issues an ACK for all register addresses in the possible address space even if the particular register does not exist.

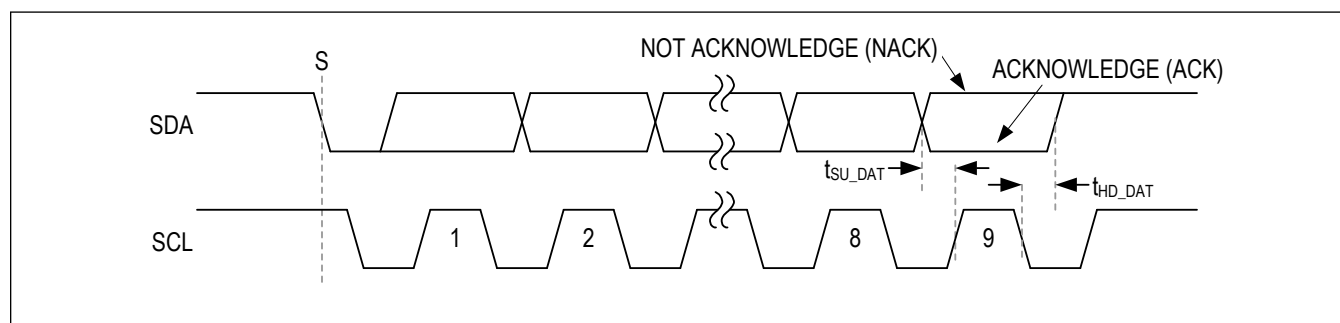


Figure 21. Acknowledge Bit

## I<sup>2</sup>C Slave Address

The I<sup>2</sup>C controller implements 7-bit slave addressing. An I<sup>2</sup>C bus master initiates communication with the slave by issuing a START condition followed by the slave address. See [Figure 22](#). The OTP address is factory-programmable for one of two options. See [Table 14](#). All slave addresses not mentioned in [Table 14](#) are not acknowledged.

**Table 14. I<sup>2</sup>C Slave Address Options**

ADDRESS	7-BIT SLAVE ADDRESS	8-BIT WRITE ADDRESS	8-BIT READ ADDRESS
Main Address (ADDR = 0)*	0x40, 0b 100 0000	0x80, 0b 1000 0000	0x81, 0b 1000 0001
Main Address (ADDR = 1)*	0x48, 0b 100 1000	0x90, 0b 1001 0000	0x91, 0b 1001 0001
Main Address (ADDR = 2)*	0x44, 0b 100 0100	0x88, 0b 1000 1000	0x89, 0b 1000 1001
Main Address (ADDR = 3)*	0x4C, 0b 100 1100	0x98, 0b 1001 1000	0x99, 0b 1001 1001
Test Mode**	0x49, 0b 100 1001	0x92, 0b 1001 0010	0x93, 0b 1001 0011

\*Perform all reads and writes on the main address. The ADDR is a factory one-time programmable (OTP) option, allowing for address changes in the event of a bus conflict.

\*\*When test mode is unlocked, the additional address is acknowledged. Test mode details are confidential. If possible, leave the test mode address unallocated to allow for the rare event that debugging needs to be performed in cooperation with Analog Devices.

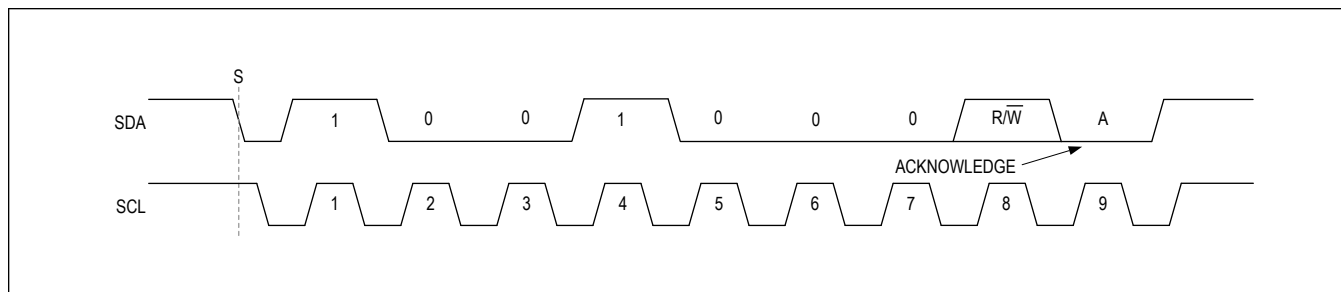


Figure 22. Slave Address Example

### I<sup>2</sup>C Clock Stretching

In general, the clock signal generation for the I<sup>2</sup>C bus is the responsibility of the master device. The I<sup>2</sup>C specification allows slow slave devices to alter the clock signal by holding down the clock line. The process in which a slave device holds down the clock line is typically called clock stretching. The IC does not use any form of clock stretching to hold down the clock line.

### I<sup>2</sup>C General Call Address

This device does not implement the I<sup>2</sup>C specifications general call address and does not acknowledge the general call address (0b0000\_0000).

### I<sup>2</sup>C Device ID

This device does not support the I<sup>2</sup>C Device ID feature.

### I<sup>2</sup>C Communication Speed

This device is compatible with all four communication speed ranges as defined by the revision 3.0 I<sup>2</sup>C specification:

- 0Hz to 100kHz (Standard Mode)
- 0Hz to 400kHz (Fast Mode)
- 0Hz to 1MHz (Fast-Mode Plus)

Operating in standard mode, fast mode, and fast-mode plus does not require any special protocols. The main consideration when changing bus speed through this range is the combination of the bus capacitance and pullup resistors. Larger values of bus capacitance and pullup resistance increase the time constant ( $C \times R$ ), slowing bus operation. Therefore, when increasing bus speeds, the pullup resistance must be decreased to maintain a reasonable time constant. Refer to the *Pullup Resistor Sizing* section of the I<sup>2</sup>C bus specification and user manual (available for free on the internet) for detailed guidance on the pullup resistor selection. In general, for bus capacitances of 200pF, a 100kHz bus needs 5.6k $\Omega$  pullup resistors, a 400kHz bus needs about 1.5k $\Omega$  pullup resistors, and a 1MHz bus needs 680 $\Omega$  pullup resistors. Remember that, while the open-drain bus is low, the pullup resistor is dissipating power, and lower-value pullup resistors dissipate more power ( $V^2/R$ ).

Operating in high-speed mode requires some special considerations. For a full list of considerations, refer to the publicly available I<sup>2</sup>C bus specification and user manual. Major considerations with respect to this part are:

- The I<sup>2</sup>C bus master uses current source pullups to shorten the signal rise.
- The I<sup>2</sup>C slave must use a different set of input filters on its SDA and SCL lines to accommodate for the higher bus.
- The communication protocols need to utilize the high-speed master code.

At power-up and after each stop condition, the bus input filters are set for standard mode, fast mode, and fast-mode plus (i.e., 0Hz to 1MHz). To switch the input filters for high-speed mode, use the high-speed master code protocols that are described in the [I<sup>2</sup>C Communication Protocols](#) section.

## I<sup>2</sup>C Communication Protocols

Both writing to and reading from registers are supported as described in the following subsections.

### Writing to a Single Register

[Figure 23](#) shows the protocol for the I<sup>2</sup>C master device to write one byte of data to this device. This protocol is the same as the SMBus specification's write-byte protocol.

The write-byte protocol is as follows:

1. The master sends a start command (S).
2. The master sends the 7-bit slave address followed by a write bit ( $R/\overline{W} = 0$ ).
3. The addressed slave asserts an acknowledge (A) by pulling SDA low.
4. The master sends an 8-bit register pointer.
5. The slave acknowledges the register pointer.
6. The master sends a data byte.
7. The slave updates with the new data.
8. The slave asserts an acknowledge or not acknowledge for the data byte. The next rising edge on SDA loads the data byte into its target register and the data becomes active.
9. The master sends a stop condition (P) or a repeated start condition (Sr). Issuing a P ensures that the bus input filters are set for 1MHz or slower operation. Issuing an Sr leaves the bus input filters in their current state.

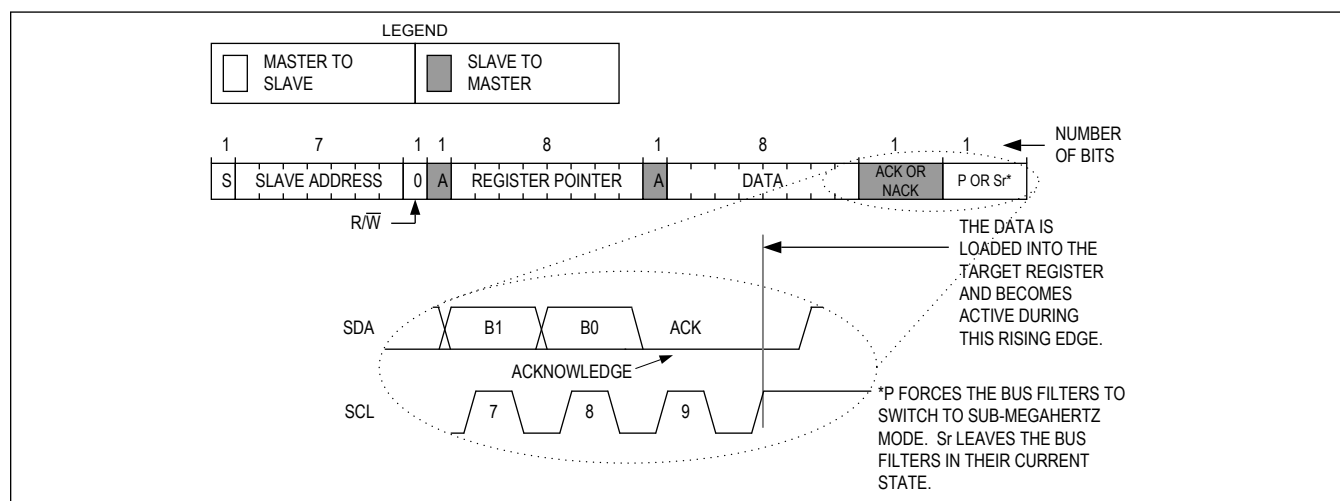


Figure 23. Writing to a Single Register with the Write-Byte Protocol

### Writing Multiple Bytes to Sequential Registers

[Figure 24](#) shows the protocol for writing to sequential registers. This protocol is similar to the write-byte protocol in the [Writing to a Single Register](#) section, except the master continues to write after it receives the first byte of data. When the master is done writing, it issues a stop or repeated start.

The write to sequential registers protocol is as follows:

1. The master sends a start command (S).
2. The master sends the 7-bit slave address followed by a write bit ( $R/\overline{W} = 0$ ).
3. The addressed slave asserts an acknowledge (A) by pulling SDA low.
4. The master sends an 8-bit register pointer.
5. The slave acknowledges the register pointer.
6. The master sends a data byte.
7. The slave acknowledges the data byte. The next rising edge on SDA loads the data byte into its target register and the data becomes active.



8. Steps 6 to 7 are repeated as many times as the master requires.
9. During the last acknowledge related clock pulse, the master can issue an acknowledge or a not acknowledge.
10. The master sends a stop condition (P) or a repeated start condition (Sr). Issuing a P ensures that the bus input filters are set for 1MHz or slower operation. Issuing an Sr leaves the bus input filters in their current state.

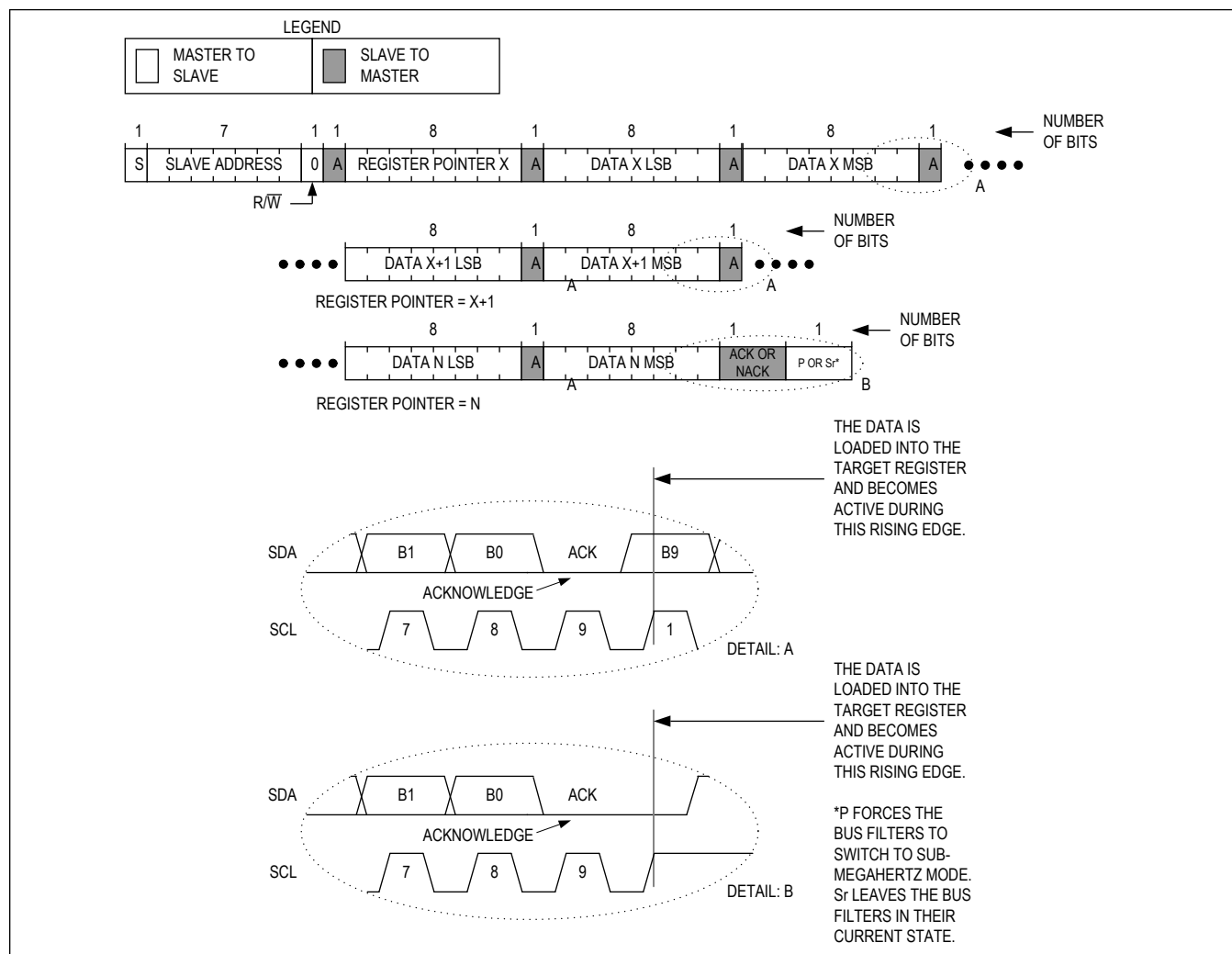


Figure 24. Writing to Sequential Registers X to N

**Reading from a Single Register**

Figure 25 shows the protocol for the I<sup>2</sup>C master device to read one byte of data. This protocol is the same as the SMBus specification's read byte protocol.

The read byte protocol is as follows:

1. The master sends a start command (S).
2. The master sends the 7-bit slave address followed by a write bit ( $R/\overline{W} = 0$ ).
3. The addressed slave asserts an acknowledge (A) by pulling SDA low.
4. The master sends an 8-bit register pointer.
5. The slave acknowledges the register pointer.
6. The master sends a repeated start command (Sr).
7. The master sends the 7-bit slave address followed by a read bit ( $R/\overline{W} = 1$ ).
8. The addressed slave asserts an acknowledge by pulling SDA low.
9. The addressed slave places 8-bits of data on the bus from the location specified by the register pointer.
10. The master issues a not acknowledge (nA).
11. The master sends a stop condition (P) or a repeated start condition (Sr). Issuing a P ensures that the bus input filters are set for 1MHz or slower operation. Issuing an Sr leaves the bus input filters in their current state.

Note that when this device receives a stop, the register pointer is not modified. Therefore, if the master re-reads the same register, it can immediately send another read command, omitting the command to send a register pointer.

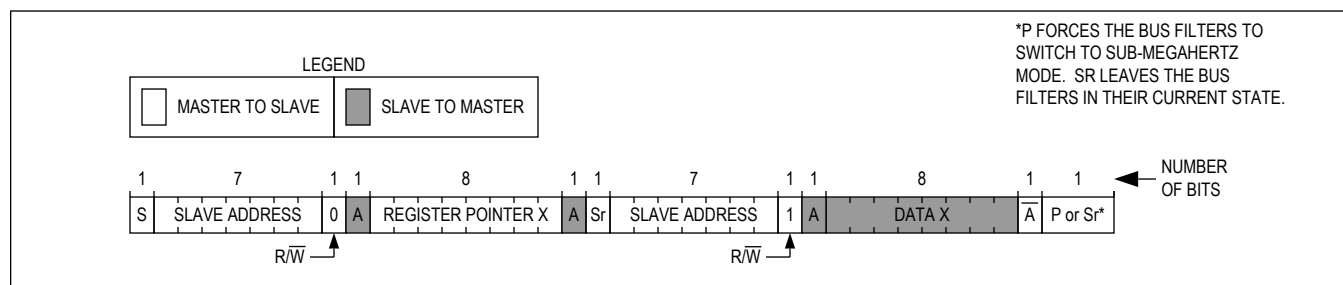


Figure 25. Reading from a Single Register with the Read Byte Protocol

### Reading from Sequential Registers

Figure 26 shows the protocol for reading from sequential registers. This protocol is similar to the read-byte protocol except the master issues an acknowledge to signal the slave that it wants more data. When the master has all the data it requires, it issues a not acknowledge (nA) and a stop (P) to end the transmission. The continuous read from sequential registers protocol is as follows:

1. The master sends a start command (S).
2. The master sends the 7-bit slave address followed by a write bit ( $R/\overline{W} = 0$ ).
3. The addressed slave asserts an acknowledge (A) by pulling SDA low.
4. The master sends an 8-bit register pointer.
5. The slave acknowledges the register pointer.
6. The master sends a repeated start command (Sr).
7. The master sends the 7-bit slave address followed by a read bit ( $R/\overline{W} = 1$ ).
8. The addressed slave asserts an acknowledge by pulling SDA low.
9. The addressed slave places 8-bits of data on the bus from the location specified by the register pointer.
10. The master issues an acknowledge (A) signaling the slave that it wishes to receive more data.
11. Steps 9 to 10 are repeated as many times as the master requires. Following the last byte of data, the master must issue a not acknowledge (nA) to signal that it wishes to stop receiving data.
12. The master sends a stop condition (P) or a repeated start condition (Sr). Issuing a stop (P) ensures that the bus input filters are set for 1MHz or slower operation. Issuing an Sr leaves the bus input filters in their current state.

Note that when this device receives a stop, it does not modify its register pointer. Therefore, if the master re-reads the same register, it can immediately send another read command, omitting the command to send a register pointer.

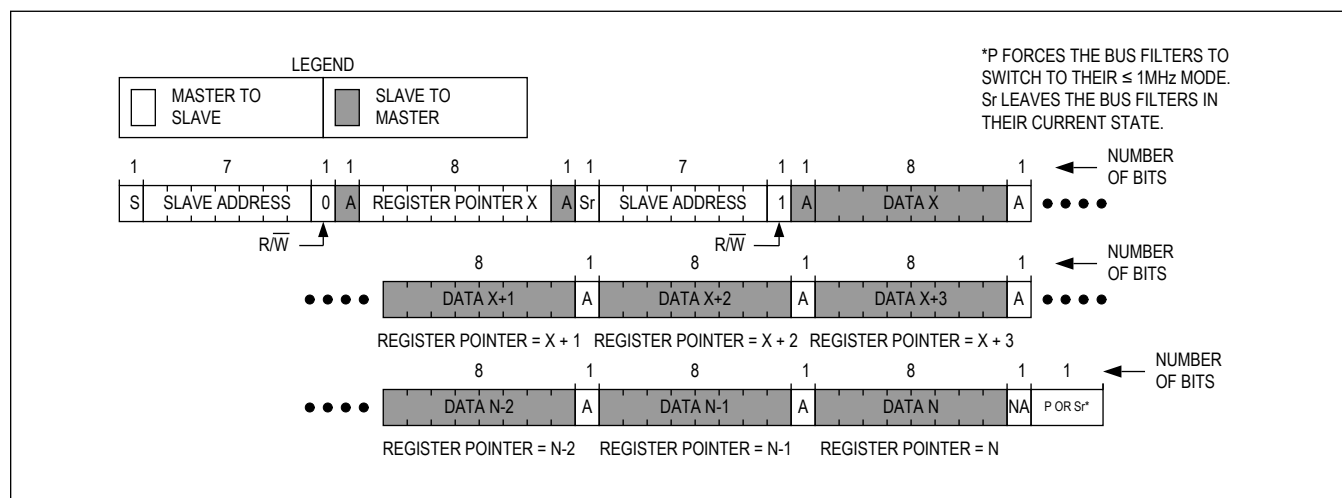


Figure 26. Reading Continuously from Sequential Registers X to N

## Register Map

## MAX77647

ADDRESS	NAME	MSB							LSB
Global									
0x00	<a href="#">INT_GLBL0[7:0]</a>	RSVD	DOD_R	TJAL2_R	TJAL1_R	nEN_R	nEN_F	GPI0_R	GPI0_F
0x01	<a href="#">INT_GLBL1[7:0]</a>	RSVD	RSVD	LDO_F	SBB2_F	SBB1_F	SBB0_F	GPI1_R	GPI1_F
0x02	<a href="#">ERCFLAG[7:0]</a>	SBB_FA ULT_F	WDT_EX P_F	SFT_CR ST_F	SFT_OF F_F	MRST_F	INUVLO	INOVLO	TOVLD
0x03	<a href="#">STAT_GLBL[7:0]</a>	DIDM	BOK	RSVD	DOD_S	TJAL2_S	TJAL1_S	STAT_E N	STAT_IR Q
0x04	<a href="#">INTM_GLBL0[7:0]</a>	RSVD	DOD_R M	TJAL2_R M	TJAL1_R M	nEN_RM	nEN_FM	GPI0_R M	GPI0_F M
0x05	<a href="#">INTM_GLBL1[7:0]</a>	RSVD	RSVD	LDO_M	SBB2_F M	SBB1_F M	SBB0_F M	GPI1_R M	GPI1_F M
0x06	<a href="#">CNFG_GLBL0[7:0]</a>	PU_DIS	MRST	SBIA_LP M	nEN_MODE[1:0]		DBEN_n EN	SFT_CTRL[1:0]	
0x07	<a href="#">CNFG_GLBL1[7:0]</a>	RSVD[4:0]					SBB_F SHUTDN	AUTO_WKT[1:0]	
0x08	<a href="#">CNFG_GPIO0[7:0]</a>	RSVD[1:0]		ALT_GPI O0	DBEN_G PI	DO	DRV	DI	DIR
0x09	<a href="#">CNFG_GPIO1[7:0]</a>	RSVD[1:0]		ALT_GPI O1	DBEN_G PI	DO	DRV	DI	DIR
0x10	<a href="#">CID[7:0]</a>	–	–	–	CID[4:0]				
0x17	<a href="#">CNFG_WDT[7:0]</a>	RSVD[1:0]		WDT_PER[1:0]		WDT_M ODE	WDT_CL R	WDT_E N	WDT_LO CK
SBB									
0x28	<a href="#">CNFG_SBB_TOP[7:0]</a>	GPI0x_L PM	–	–	–	–	–	DRV_SBB[1:0]	
0x29	<a href="#">CNFG_SBB0_A[7:0]</a>	TV_SBB0[7:0]							
0x2A	<a href="#">CNFG_SBB0_B[7:0]</a>	OP_MODE[1:0]		IP_SBB0[2:0]			EN_SBB0[2:0]		
0x2B	<a href="#">CNFG_SBB1_A[7:0]</a>	TV_SBB1[7:0]							
0x2C	<a href="#">CNFG_SBB1_B[7:0]</a>	OP_MODE[1:0]		IP_SBB1[2:0]			EN_SBB1[2:0]		
0x2D	<a href="#">CNFG_SBB2_A[7:0]</a>	TV_SBB2[7:0]							
0x2E	<a href="#">CNFG_SBB2_B[7:0]</a>	OP_MODE[1:0]		IP_SBB2[2:0]			EN_SBB2[2:0]		
0x2F	<a href="#">CNFG_DVS_SBB0_A[7:0]</a>	TV_SBB0_DVS[7:0]							
0x30	<a href="#">CNFG_SBB_BTM[7:0]</a>	–	–	DIS_LP M_SBB2	DIS_LP M_SBB1	DIS_LP M_SBB0	ADE_SB B2	ADE_SB B1	ADE_SB B0
LDO									
0x38	<a href="#">CNFG_LDO_A[7:0]</a>	TV_OFS _LDO	TV_LDO[6:0]						
0x39	<a href="#">CNFG_LDO_B[7:0]</a>	–	–	–	LDO_MD	ADE_LD O	EN_LDO[2:0]		

## Register Details

[INT\\_GLBL0 \(0x00\)](#)

BIT	7	6	5	4	3	2	1	0
Field	RSVD	DOD_R	TJAL2_R	TJAL1_R	nEN_R	nEN_F	GPI0_R	GPI0_F
Reset	0b0	0b0	0b0	0b0	0b0	0b0	0b0	0b0
Access Type	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	7	RSVD	
DOD_R	6	LDO Dropout Detector Rising Interrupt	0b0 = The LDO has not detected dropout since the last time this bit was read. 0b1 = The LDO has detected dropout since the last time this bit was read.
TJAL2_R	5	Thermal Alarm 2 Rising Interrupt	0b0 = The junction temperature has not risen above TJAL2 since the last time this bit was read. 0b1 = The junction temperature has risen above TJAL2 since the last time this bit was read.
TJAL1_R	4	Thermal Alarm 1 Rising Interrupt	0b0 = The junction temperature has not risen above TJAL1 since the last time this bit was read. 0b1 = The junction temperature has risen above TJAL1 since the last time this bit was read.
nEN_R	3	nEN Rising Interrupt	0b0 = No nEN rising edges have occurred since the last time this bit was read. 0b1 = A nEN rising edge has occurred since the last time this bit was read.
nEN_F	2	nEN Falling Interrupt	0b0 = No nEN falling edges have occurred since the last time this bit was read. 0b1 = A nEN falling edge occurred since the last time this bit was read.
GPI0_R	1	GPI Rising Interrupt Note that "GPI" refers to the GPIO programmed to be an input.	0b0 = No GPI rising edges have occurred since the last time this bit was read. 0b1 = A GPI rising edge has occurred since the last time this bit was read.
GPI0_F	0	GPI Falling Interrupt Note that the GPI is the GPIO programmed to be an input.	0b0 = No GPI falling edges have occurred since the last time this bit was read. 0b1 = A GPI falling edge has occurred since the last time this bit was read.

[INT\\_GLBL1 \(0x01\)](#)

BIT	7	6	5	4	3	2	1	0
Field	RSVD	RSVD	LDO_F	SBB2_F	SBB1_F	SBB0_F	GPI1_R	GPI1_F
Reset	0b000	0b000	0b0	0b0	0b0	0b0	0b0	0b0
Access Type	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All
BITFIELD	BITS	DESCRIPTION			DECODE			
RSVD	7	Reserved. Unutilized bit. Write to 0. Reads are don't care.						

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	6	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
LDO_F	5	LDO Fault Interrupt	0b0 = No fault has occurred on LDO since the last time this bit was read. 0b1 = LDO has fallen out of regulation since the last time this bit was read.
SBB2_F	4	SBB2 Fault Indicator	0b0 = No fault has occurred on SBB2 since the last time this bit was read. 0b1 = SBB2 has fallen out of regulation since the last time this bit was read.
SBB1_F	3	SBB1 Fault Indicator	0b0 = No fault has occurred on SBB2 since the last time this bit was read. 0b1 = SBB1 has fallen out of regulation since the last time this bit was read.
SBB0_F	2	SBB0 Fault Indicator	0b0 = No fault has occurred on SBB2 since the last time this bit was read. 0b1 = SBB0 has fallen out of regulation since the last time this bit was read.
GPI1_R	1	GPI1 Rising Interrupt Note that "GPI" refers to the GPIO programmed to be an input.	0b0 = No GPI1 rising edges have occurred since the last time this bit was read. 0b1 = A GPI1 rising edge has occurred since the last time this bit was read.
GPI1_F	0	GPI1 Falling Interrupt Note that the GPI is the GPIO programmed to be an input.	0b0 = No GPI1 falling edges have occurred since the last time this bit was read. 0b1 = A GPI1 falling edge has occurred since the last time this bit was read.

**ERCFLAG (0x02)**

BIT	7	6	5	4	3	2	1	0
Field	SBB_FAULT_F	WDT_EXP_F	SFT_CRST_F	SFT_OFF_F	MRST_F	INUVLO	INOVLO	TOVLD
Reset	0b0	0b0	0b0	0b0	0b0	0b0	0b0	0b0
Access Type	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All	Read Clears All

BITFIELD	BITS	DESCRIPTION	DECODE
SBB_FAULT_F	7	SBBx Fault and Shutdown Flag. This bit sets when a SBBx fault and consequent SBBx shutdown occurs.	0b0 = No SBB shutdown occurred since the last time this bit was read. 0b1 = SBBx fault and SBB shutdown occurred since the last time this bit was read.
WDT_EXP_F	6	Watchdog Timer OFF or RESET Flag. This bit sets when the watchdog timer expires and causes a power-off or reset based on WDT_MODE.	0b0 = Watchdog timer has not caused a power-off or reset since the last time this bit was read. 0b1 = Watchdog timer has expired and caused a power-off or reset since the last time this bit was read.
SFT_CRST_F	5	Software Cold Reset Flag	0b0 = The software cold reset has not occurred since the last read of this register. 0b1 = The software cold reset has occurred since the last read of this register. This indicates that software has set SFT_CTRL[1:0] = 0b01.

BITFIELD	BITS	DESCRIPTION	DECODE
SFT_OFF_F	4	Software OFF Flag	0b0 = The SFT_OFF function has not occurred since the last read of this register. 0b1 = The SFT_OFF function has occurred since the last read of this register. This indicates that software has set SFT_CTRL[1:0] = 0b10.
MRST_F	3	Manual Reset Timer	0b0 = A manual reset has not occurred since the last read of this register. 0b1 = A manual reset has occurred since the last read of this register.
INUVLO	2	IN Domain Undervoltage Lockout	0b0 = The IN domain undervoltage lockout has not occurred since the last read of this register. 0b1 = The IN domain undervoltage lockout has occurred since the last read of this register. This indicates that the IN domain voltage fell below VINUVLO (~2.6V).
INOVLO	1	IN Domain Overvoltage Lockout	0b0 = The IN domain overvoltage lockout has not occurred since the last read of this register. 0b1 = The IN domain overvoltage lockout has occurred since the last read of this register. This indicates that the IN domain voltage rose above VINOVL0 (~5.85V).
TOVLD	0	Thermal Overload	0b0 = Thermal overload has not occurred since the last read of this register. 0b1 = Thermal overload has occurred since the last read of this register. This indicates that the junction temperature has exceeded +165°C.

**STAT\_GLBL (0x03)**

BIT	7	6	5	4	3	2	1	0
Field	DIDM	BOK	RSVD	DOD_S	TJAL2_S	TJAL1_S	STAT_EN	STAT_IRQ
Reset	OTP	0b1	0b0	0b0	0b0	0b0	0b0	0b0
Access Type	Read Only	Read Only	Read Clears All	Read Only	Read Only	Read Only	Read Only	Read Only

BITFIELD	BITS	DESCRIPTION	DECODE
DIDM	7	Device Identification Bits for Metal Options	0b0 = MAX77647 0b1 = Reserved
BOK	6	BOK Interrupt Status	0b0 = Main bias is not ready. 0b1 = Main bias enabled and ready.
RSVD	5	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
DOD_S	4	LDO Dropout Detector Rising Status	0b0 = LDO is not in dropout. 0b1 = LDO is in dropout.
TJAL2_S	3	Thermal Alarm 2 Status	0b0 = The junction temperature is less than TJA2. 0b1 = The junction temperature is greater than TJAL2.
TJAL1_S	2	Thermal Alarm 1 Status	0b0 = The junction temperature is less than TJAL1. 0b1 = The junction temperature is greater than TJAL1.
STAT_EN	1	Debounced Status for the nEN Input	0b0 = nEN is not active (logic-high). 0b1 = nEN is active (logic-low).

BITFIELD	BITS	DESCRIPTION	DECODE
STAT_IRQ	0	Software Version of the nIRQ MOSFET Gate Drive	0b0 = Unmasked gate drive is logic-low. 0b1 = Unmasked gate drive is logic-high.

**INTM\_GLBL0 (0x04)**

BIT	7	6	5	4	3	2	1	0
Field	RSVD	DOD_RM	TJAL2_RM	TJAL1_RM	nEN_RM	nEN_FM	GPI0_RM	GPI0_FM
Reset	0b0	0b1	0b1	0b1	0b1	0b1	0b1	0b1
Access Type	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	7	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
DOD_RM	6	LDO Dropout Detector Rising Interrupt Mask	0b0 = Unmasked. If DOD_R goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to DOD_R.
TJAL2_RM	5	Thermal Alarm 2 Rising Interrupt Mask	0b0 = Unmasked. If TJAL2_R goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to TJAL2_R.
TJAL1_RM	4	Thermal Alarm 1 Rising Interrupt Mask	0b0 = Unmasked. If TJAL1_R goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to TJAL1_R.
nEN_RM	3	nEN Rising Interrupt Mask	0b0 = Unmasked. If nEN_R goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to nEN_R.
nEN_FM	2	nEN Falling Interrupt Mask	0b0 = Unmasked. If nEN_F goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to nEN_F.
GPI0_RM	1	GPI0 Rising Interrupt Mask	0b0 = Unmasked. If GPI0_R goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to GPI0_R.
GPI0_FM	0	GPI0 Falling Interrupt Mask	0b0 = Unmasked. If GPI0_F goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to GPI0_F.



[INTM\\_GLBL1 \(0x05\)](#)

BIT	7	6	5	4	3	2	1	0
Field	RSVD	RSVD	LDO_M	SBB2_FM	SBB1_FM	SBB0_FM	GPI1_RM	GPI1_FM
Reset	0b0	0b0	0b1	0b0	0b0	0b0	0b1	0b1
Access Type	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	7	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
RSVD	6	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
LDO_M	5	LDO Fault Interrupt Mask	0b0 = Unmasked. If LDO_F goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to LDO_F.
SBB2_FM	4	SBB2 Fault Interrupt Mask	0b0 = Unmasked. If SBB2_F goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to SBB2_F.
SBB1_FM	3	SBB2 Fault Interrupt Mask	0b0 = Unmasked. If SBB1_F goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to SBB1_F.
SBB0_FM	2	SBB2 Fault Interrupt Mask	0b0 = Unmasked. If SBB0_F goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to SBB0_F.
GPI1_RM	1	GPI1 Rising Interrupt Mask	0b0 = Unmasked. If GPI1_R goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to GPI1_R.
GPI1_FM	0	GPI1 Falling Interrupt Mask	0b0 = Unmasked. If GPI1_F goes from 0 to 1, then nIRQ goes low. nIRQ goes high when all interrupt bits are cleared. 0b1 = Masked. nIRQ does not go low due to GPI1_F.

[CNFG\\_GLBL0 \(0x06\)](#)

BIT	7	6	5	4	3	2	1	0
Field	PU_DIS	MRST	SBIA_LPM	nEN_MODE[1:0]		DBEN_nEN	SFT_CTRL[1:0]	
Reset	0b0	OTP	OTP	OTP		OTP	0b00	
Access Type	Write, Read	Write, Read	Write, Read	Write, Read		Write, Read	Write, Read	

BITFIELD	BITS	DESCRIPTION	DECODE
PU_DIS	7	nEN Internal Pullup Resistor	0b0 = Strong internal nEN pullup (200kΩ). 0b1 = Weak internal nEN pullup (10MΩ).
MRST	6	Sets the Manual Reset Time (t <sub>MRST</sub> )	0b0 = 8s 0b1 = 4s
SBIA_LPM	5	Main Bias Power Mode	0b0 = Main bias is in normal-power mode by software. 0b1 = Main bias is in auto-power mode by software.
nEN_MODE	4:3	nEN Input (ON-KEY) Default Configuration Mode	0b00 = Push-button mode 0b01 = Slide-switch mode 0b10 = Logic mode 0b11 = Reserved
DBEN_nEN	2	Debounce Timer Enable for the nEN Pin	0b0 = 500μs Debounce 0b1 = 30ms Debounce  Applies only to push-button and slide-switch mode.
SFT_CTRL	1:0	Software Reset Functions  Note that the SFT_CRST and SFT_OFF commands initiate the power-down sequence flow as described in the data sheet. This power-down sequence flow has delay elements that add up to 205.24ms (60ms delay + 10.24ms nRST assert delay + 4x2.56ms power-down slot delays + 125ms output discharge delay). If issuing the SFT_CRST and/or SFT_OFF functions in software, wait for more than 300ms before trying to issue any additional commands through I <sup>2</sup> C.	0b00 = No action. 0b01 = Software cold reset (SFT_CRST). The device powers down, resets, and then powers up again. 0b10 = Software off (SFT_OFF). The device powers down, resets, and then remains off and waiting for a wake-up event. 0b11 = Auto wake up (SFT_AUTO). The device powers down, and then automatically starts power up sequence after time determined in AUTO_WKT. Note that the registers do not get reset with auto wake up.

**CNFG\_GLBL1 (0x07)**

BIT	7	6	5	4	3	2	1	0
Field	RSVD[4:0]					SBB_F_SH UTDN	AUTO_WKT[1:0]	
Reset	0b0000					OTP	0b00	
Access Type	Write, Read					Read Only	Write, Read	

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	7:3	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
SBB_F_SHU TDN	2	SBB Shutdown from SBB Fault	0b0 = The SBB regulators do not shut off when SBB fault occurs. 0b1 = The SBB regulators powers down sequentially when a SBB fault occurs.
AUTO_WKT	1:0	Auto Wake-Up Timer	0b00 = 100ms auto wake-up time 0b01 = 200ms auto wake-up time 0b10 = 500ms auto wake-up time 0b11 = 1000ms auto wake-up time

[CNFG\\_GPIO0 \(0x08\)](#)

BIT	7	6	5	4	3	2	1	0
Field	RSVD[1:0]		ALT_GPIO0	DBEN_GPI	DO	DRV	DI	DIR
Reset	0b00		OTP	0b0	0b0	0b0	0b0	0b1
Access Type	Write, Read		Write, Read	Write, Read	Write, Read	Write, Read	Read Only	Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	7:6	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
ALT_GPIO0	5	Alternate Mode Enable for GPIO0	0b0 = Standard GPIO 0b1 = Active-high GPI controls whether SBB0 is set by TV_SBB0 or TV_SBB0_DVS when GPIOx_LPM = 1.
DBEN_GPI	4	General Purpose Input Debounce Timer Enable	0b0 = No debounce 0b1 = 30ms debounce
DO	3	General Purpose Output Data Output	This bit is a don't care when DIR = 1 (configured as input).  When set for GPO (DIR = 0): 0b0 = GPIO is output logic-low. 0b1 = GPIO is output logic-high when set as push-pull output (DRV = 1). GPIO is high-impedance when set as an open-drain output (DRV = 0).
DRV	2	General Purpose Output Driver Type	This bit is a don't care when DIR = 1 (configured as input).  When set for GPO (DIR = 0): 0b0 = Open-drain 0b1 = Push-pull
DI	1	GPIO Digital Input Value. Irrespective of whether the GPIO is set for GPI (DIR = 1) or GPO (DIR = 0), DI reflects the state of the GPIO.	0b0 = Input logic-low 0b1 = Input logic-high
DIR	0	GPIO Direction	0b0 = General purpose output (GPO) 0b1 = General purpose input (GPI)

[CNFG\\_GPIO1 \(0x09\)](#)

BIT	7	6	5	4	3	2	1	0
Field	RSVD[1:0]		ALT_GPIO1	DBEN_GPI	DO	DRV	DI	DIR
Reset	0b00		OTP	0b0	0b0	0b0	0b0	0b1
Access Type	Write, Read		Write, Read	Write, Read	Write, Read	Write, Read	Read Only	Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	7:6	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
ALT_GPIO1	5	Alternate Mode Enable for GPIO1	0b0 = Standard GPIO 0b1 = Active-high GPO of flexible power sequencer active-high output for SBB2 when GPIOx_LPM = 1.

BITFIELD	BITS	DESCRIPTION	DECODE
DBEN_GPI	4	General Purpose Input Debounce Timer Enable	0b0 = No debounce 0b1 = 30ms debounce
DO	3	General Purpose Output Data Output	This bit is a don't care when DIR = 1 (configured as input).  When set for GPO (DIR = 0): 0b0 = GPIO is output logic-low. 0b1 = GPIO is output logic-high when set as push-pull output (DRV = 1). GPIO is high-impedance when set as an open-drain output (DRV = 0).
DRV	2	General Purpose Output Driver Type	This bit is a don't care when DIR = 1 (configured as input).  When set for GPO (DIR = 0): 0b0 = Open-drain 0b1 = Push-pull
DI	1	GPIO Digital Input Value. Irrespective of whether the GPIO is set for GPI (DIR = 1) or GPO (DIR = 0), DI reflects the state of the GPIO.	0b0 = Input logic-low 0b1 = Input logic-high
DIR	0	GPIO Direction	0b0 = General purpose output (GPO) 0b1 = General purpose input (GPI)

**CID (0x10)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	–	CID[4:0]				
Reset	–	–	–	OTP				
Access Type	–	–	–	Read Only				

BITFIELD	BITS	DESCRIPTION
CID	4:0	Chip Identification Code The chip identification code refers to a set of reset values in the register map or the "OTP configuration."

**CNFG\_WDT (0x17)**

BIT	7	6	5	4	3	2	1	0
Field	RSVD[1:0]		WDT_PER[1:0]		WDT_MOD E	WDT_CLR	WDT_EN	WDT_LOC K
Reset	0b00		0b11		0b0	0b0	OTP	OTP
Access Type	Write, Read		Write, Read		Write, Read	Write, Read	Write, Read	Read Only

BITFIELD	BITS	DESCRIPTION	DECODE
RSVD	7:6	Reserved. Unutilized bit. Write to 0. Reads are don't care.	
WDT_PER	5:4	Watchdog Timer Period. Sets $t_{WD}$ . Watchdog timer is reset to the programmed value as soon as this bitfield is changed.	0b00 = 16 seconds 0b01 = 32 seconds 0b10 = 64 seconds 0b11 = 128 seconds

BITFIELD	BITS	DESCRIPTION	DECODE
WDT_MODE	3	Watchdog Timer Expired Action. Determines what the IC does after the watchdog timer expires.	0b0 = Watchdog timer expire causes power-off. 0b1 = Watchdog timer expire causes power reset.
WDT_CLR	2	Watchdog Timer Clear Control. Set this bit to feed (reset) the watchdog timer.	0b0 = Watchdog timer period is not reset. 0b1 = Watchdog timer is reset back to $t_{WD}$ .
WDT_EN	1	Watchdog Timer Enable. Write protected depending on WDT_LOCK.	0b0 = Watchdog timer is not enabled. 0b1 = Watchdog timer is enabled. The timer expires if not reset by setting WDT_CLR.
WDT_LOCK	0	Factory-Set Safety Bit for the Watchdog Timer. Determines if the timer can be disabled through WDT_EN or not.	0b0 = Watchdog timer can be enabled and disabled with WDT_EN. 0b1 = Watchdog timer can not be disabled with WDT_EN. However, WDT_EN can still be used to enable the watchdog timer.

**CNFG\_SBB\_TOP (0x28)**

BIT	7	6	5	4	3	2	1	0
Field	GPIOx_LPM	–	–	–	–	–	DRV_SBB[1:0]	
Reset	OTP	–	–	–	–	–	OTP	
Access Type	Write, Read	–	–	–	–	–	Write, Read	

BITFIELD	BITS	DESCRIPTION	DECODE
GPIOx_LPM	7	Controls power mode feature for all SIMO channels. Enables alternate mode for GPIOx.	0b0 = Disable low-power mode feature for all SIMO channels. 0b1 = Automatic low-power mode feature for all SIMO channels when ALT_GPIOx = 0. 0b1 = Active-high GPI, controls whether SBB0 is set by TV_SBB0 or TV_SBB0_DVS when ALT_GPIO0 = 1. 0b1 = Active-high GPO of flexible power sequencer active-high output for SBB2 when ALT_GPIO1 = 1.
DRV_SBB	1:0	SIMO Buck-Boost (All Channels) Drive Strength Trim. See the <i>Drive Strength</i> section for more details.	0b00 = Fastest transition time. 0b01 = A little slower than 0b00. 0b10 = A little slower than 0b01. 0b11 = A little slower than 0b10.

**CNFG\_SBB0\_A (0x29)**

BIT	7	6	5	4	3	2	1	0
Field	TV_SBB0[7:0]							
Reset	OTP							
Access Type	Write, Read							

BITFIELD	BITS	DESCRIPTION	DECODE
TV_SBB0	7:0	SIMO Buck-Boost Channel 0 Target Output Voltage This 8-bit configuration is a linear transfer function that starts at 0.5V and ends at 5.1V with 25mV increments.	0x00 = 0.500V 0x01 = 0.525V 0x02 = 0.550V 0x03 = 0.575V 0x04 = 0.600V 0x05 = 0.625V 0x06 = 0.650V 0x07 = 0.675V 0x08 = 0.700V ... 0xB6 = 5.050V 0xB7 = 5.075V 0xB8 = 5.100V 0xB9 to 0xFF = RSVD

**CNFG\_SBB0\_B (0x2A)**

BIT	7	6	5	4	3	2	1	0
Field	OP_MODE[1:0]		IP_SBB0[2:0]			EN_SBB0[2:0]		
Reset	OTP		OTP			OTP		
Access Type	Write, Read		Write, Read			Write, Read		

BITFIELD	BITS	DESCRIPTION	DECODE
OP_MODE	7:6	Operation Mode of SBB0	0b00 = Automatic 0b01 = Buck mode 0b10 = Boost mode 0b11 = Buck-boost mode
IP_SBB0	5:3	SIMO Buck-Boost Channel 0 Peak Current Limit	0b000 = 0.910A 0b001 = 0.700A 0b010 = 0.500A
EN_SBB0	2:0	Enable Control for SIMO Buck-Boost Channel 0. Select the FPS slot that the channel powers up and powers down in or whether the channel is forced on or off.	0b000 = FPS slot 0 0b001 = FPS slot 1 0b010 = FPS slot 2 0b011 = FPS slot 3 0b100 = Off irrespective of FPS 0b101 = Same as 0b100 0b110 = On irrespective of FPS 0b111 = Same as 0b110

**CNFG\_SBB1\_A (0x2B)**

BIT	7	6	5	4	3	2	1	0
Field	TV_SBB1[7:0]							
Reset	OTP							
Access Type	Write, Read							

BITFIELD	BITS	DESCRIPTION	DECODE
TV_SBB1	7:0	SIMO Buck-Boost Channel 1 Target Output Voltage This 8-bit configuration is a linear transfer function that starts at 0.5V and ends at 5.1V with 25mV increments.	0x00 = 0.500V 0x01 = 0.525V 0x02 = 0.550V 0x03 = 0.575V 0x04 = 0.600V 0x05 = 0.625V 0x06 = 0.650V 0x07 = 0.675V 0x08 = 0.700V ... 0xB6 = 5.050V 0xB7 = 5.075V 0xB8 = 5.100V 0xB9 to 0xFF = RSVD

**CNFG\_SBB1\_B (0x2C)**

BIT	7	6	5	4	3	2	1	0
Field	OP_MODE[1:0]		IP_SBB1[2:0]			EN_SBB1[2:0]		
Reset	OTP		OTP			OTP		
Access Type	Write, Read		Write, Read			Write, Read		

BITFIELD	BITS	DESCRIPTION	DECODE
OP_MODE	7:6	Operation Mode of SBB1	0b00 = Automatic 0b01 = Buck mode 0b10 = Boost mode 0b11 = Buck-boost mode
IP_SBB1	5:3	SIMO Buck-Boost Channel 1 Peak Current Limit	0b000 = 0.910A 0b001 = 0.700A 0b010 = 0.500A
EN_SBB1	2:0	Enable Control for SIMO Buck-Boost Channel 1. Select the FPS slot that the channel powers up and powers down in or whether the channel is forced on or off.	0b000 = FPS slot 0 0b001 = FPS slot 1 0b010 = FPS slot 2 0b011 = FPS slot 3 0b100 = Off irrespective of FPS 0b101 = Same as 0b100 0b110 = On irrespective of FPS 0b111 = Same as 0b110

**CNFG\_SBB2\_A (0x2D)**

BIT	7	6	5	4	3	2	1	0
Field	TV_SBB2[7:0]							
Reset	OTP							
Access Type	Write, Read							

BITFIELD	BITS	DESCRIPTION	DECODE
TV_SBB2	7:0	SIMO Buck-Boost Channel 2 Target Output Voltage This 8-bit configuration is a linear transfer function that starts at 0.5V and ends at 5.1V with 25mV increments.	0x00 = 0.500V 0x01 = 0.525V 0x02 = 0.550V 0x03 = 0.575V 0x04 = 0.600V 0x05 = 0.625V 0x06 = 0.650V 0x07 = 0.675V 0x08 = 0.700V ... 0xB6 = 5.050V 0xB7 = 5.075V 0xB8 = 5.100V 0xB9 to 0xFF = RSVD

**CNFG\_SBB2\_B (0x2E)**

BIT	7	6	5	4	3	2	1	0
Field	OP_MODE[1:0]		IP_SBB2[2:0]			EN_SBB2[2:0]		
Reset	OTP		OTP			OTP		
Access Type	Write, Read		Write, Read			Write, Read		

BITFIELD	BITS	DESCRIPTION	DECODE
OP_MODE	7:6	Operation Mode of SBB2	0b00 = Automatic 0b01 = Buck mode 0b10 = Boost mode 0b11 = Buck-boost mode
IP_SBB2	5:3	SIMO Buck-Boost Channel 2 Peak Current Limit	0b000 = 0.910A 0b001 = 0.700A 0b010 = 0.500A
EN_SBB2	2:0	Enable Control for SIMO Buck-Boost Channel 2. Select the FPS slot that the channel powers up and powers down in or whether the channel is forced on or off.	0b000 = FPS slot 0 0b001 = FPS slot 1 0b010 = FPS slot 2 0b011 = FPS slot 3 0b100 = Off irrespective of FPS 0b101 = Same as 0b100 0b110 = On irrespective of FPS 0b111 = Same as 0b110

**CNFG\_DVS\_SBB0\_A (0x2F)**

BIT	7	6	5	4	3	2	1	0
Field	TV_SBB0_DVS[7:0]							
Reset	OTP							
Access Type	Write, Read							



BITFIELD	BITS	DESCRIPTION	DECODE
TV_SBB0_D VS	7:0	SIMO Buck-Boost Channel 0 Target Output Voltage This 8-bit configuration is a linear transfer function that starts at 0.5V and ends at 5.1V with 25mV increments.	0x00 = 0.500V 0x01 = 0.525V 0x02 = 0.550V 0x03 = 0.575V 0x04 = 0.600V 0x05 = 0.625V 0x06 = 0.650V 0x07 = 0.675V 0x08 = 0.700V ... 0xB6 = 5.050V 0xB7 = 5.075V 0xB8 = 5.100V 0xB9 to 0xFF = RSVD

**CNFG\_SBB\_BTMM(0x30)**

BIT	7	6	5	4	3	2	1	0
Field	–	–	DIS_LPM_S BB2	DIS_LPM_S BB1	DIS_LPM_S BB0	ADE_SBB2	ADE_SBB1	ADE_SBB0
Reset	–	–	0b0	0b0	0b0	OTP	OTP	OTP
Access Type	–	–	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read	Write, Read

BITFIELD	BITS	DESCRIPTION	DECODE
DIS_LPM_S BB2	5	Disables the Automatic Low-Power Mode for SBB2	0b0 = Automatic low-power mode feature for SBB2 0b1 = Disable low-power mode feature for SBB2
DIS_LPM_S BB1	4	Disables the Automatic Low-Power Mode for SBB1	0b0 = Automatic low-power mode feature for SBB1 0b1 = Disable low-power mode feature for SBB1
DIS_LPM_S BB0	3	Disables the Automatic Low-Power Mode for SBB0	0b0 = Automatic low-power mode feature for SBB0 0b1 = Disable low-power mode feature for SBB0
ADE_SBB2	2	SIMO Buck-Boost Channel 2 Active-Discharge Enable	0b0 = The active-discharge function is disabled. When SBB2 is disabled, its discharge rate is a function of the output capacitance and the external load. 0b1 = The active-discharge function is enabled. When SBB2 is disabled, an internal resistor (RAD_SBB2) is activated from SBB2 to PGND to help the output voltage discharge. The output voltage discharge rate is a function of the output capacitance, the external loading, and the internal RAD_SBB2 load.
ADE_SBB1	1	SIMO Buck-Boost Channel 1 Active-Discharge Enable	0b0 = The active-discharge function is disabled. When SBB1 is disabled, its discharge rate is a function of the output capacitance and the external load. 0b1 = The active-discharge function is enabled. When SBB1 is disabled, an internal resistor (RAD_SBB1) is activated from SBB1 to PGND to help the output voltage discharge. The output voltage discharge rate is a function of the output capacitance, the external loading, and the internal RAD_SBB1 load.

BITFIELD	BITS	DESCRIPTION	DECODE
ADE_SBB0	0	SIMO Buck-Boost Channel 0 Active-Discharge Enable	0b0 = The active-discharge function is disabled. When SBB0 is disabled, its discharge rate is a function of the output capacitance and the external load. 0b1 = The active-discharge function is enabled. When SBB0 is disabled, an internal resistor (RAD_SBB0) is activated from SBB0 to PGND to help the output voltage discharge. The output voltage discharge rate is a function of the output capacitance, the external loading, and the internal RAD_SBB0 load.

**CNFG\_LDO\_A (0x38)**

BIT	7	6	5	4	3	2	1	0
Field	TV_OFS_LDO	TV_LDO[6:0]						
Reset	OTP	OTP						
Access Type	Write, Read	Write, Read						

BITFIELD	BITS	DESCRIPTION	DECODE
TV_OFS_LDO	7	LDO Output Voltage. This bit applies a 1.325V offset to the output voltage of the LDO.	0b0 = No offset 0b1 = 1.325V Offset
TV_LDO	6:0	LDO Target Output Voltage This 7-bit configuration is a linear transfer function that starts at 0.5V and ends at 3.675V with 25mV increments.	0x00 = 0.500V 0x01 = 0.525V 0x02 = 0.550V 0x03 = 0.575V 0x04 = 0.600V 0x05 = 0.625V 0x06 = 0.650V 0x07 = 0.675V 0x08 = 0.700V ... 0x7E = 3.650V 0x7F = 3.675V  When TV_LDO[7] = 0, TV_LDO[6:0] sets the LDO's output voltage range from 0.5V to 3.675V.  When TV_LDO[7] = 1, TV_LDO[6:0] sets the LDO's output voltage from 1.825V to 5V.

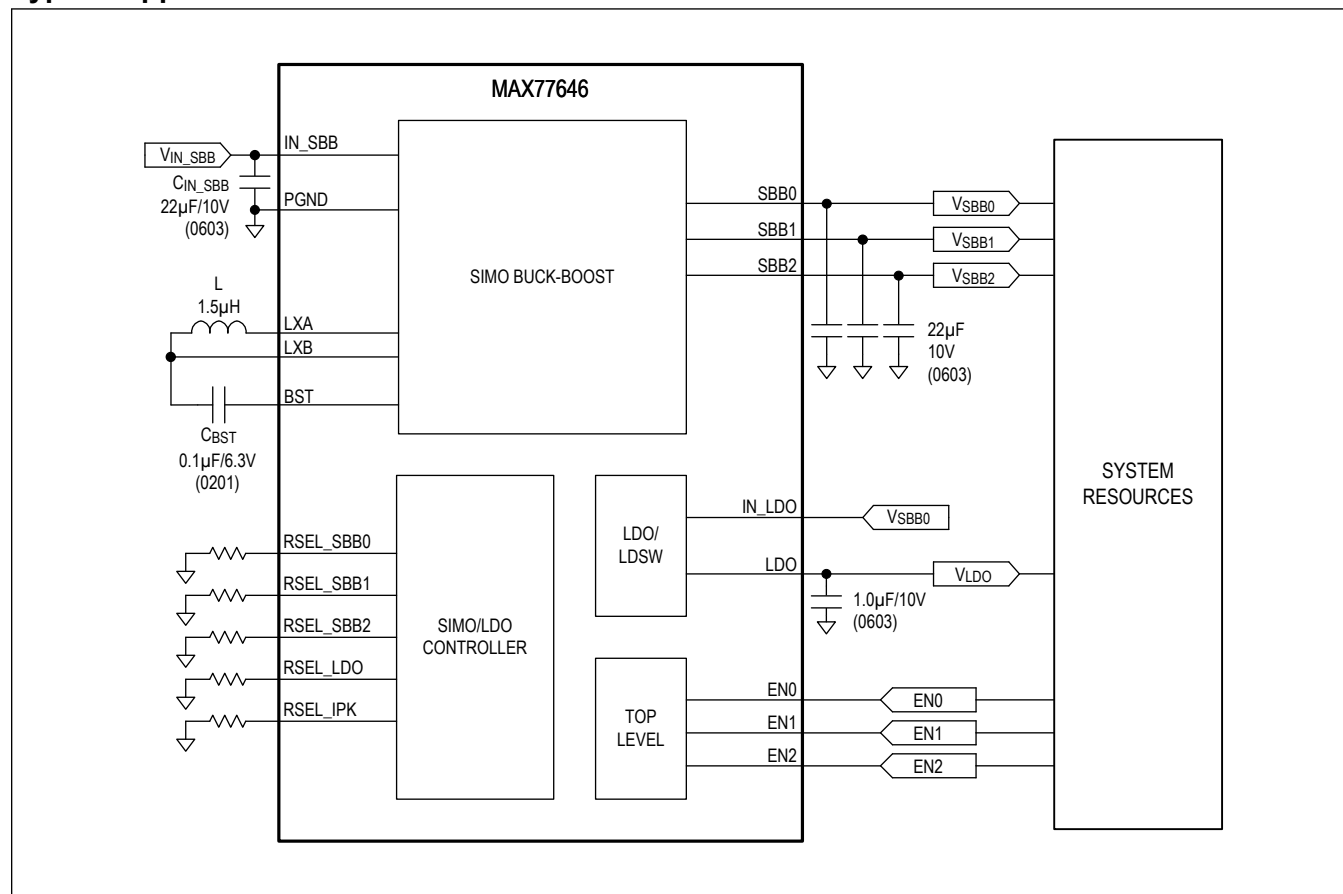
**CNFG\_LDO\_B (0x39)**

BIT	7	6	5	4	3	2	1	0
Field	—	—	—	LDO_MD	ADE_LDO	EN_LDO[2:0]		
Reset	—	—	—	OTP	OTP	OTP		
Access Type	—	—	—	Write, Read	Write, Read	Write, Read		

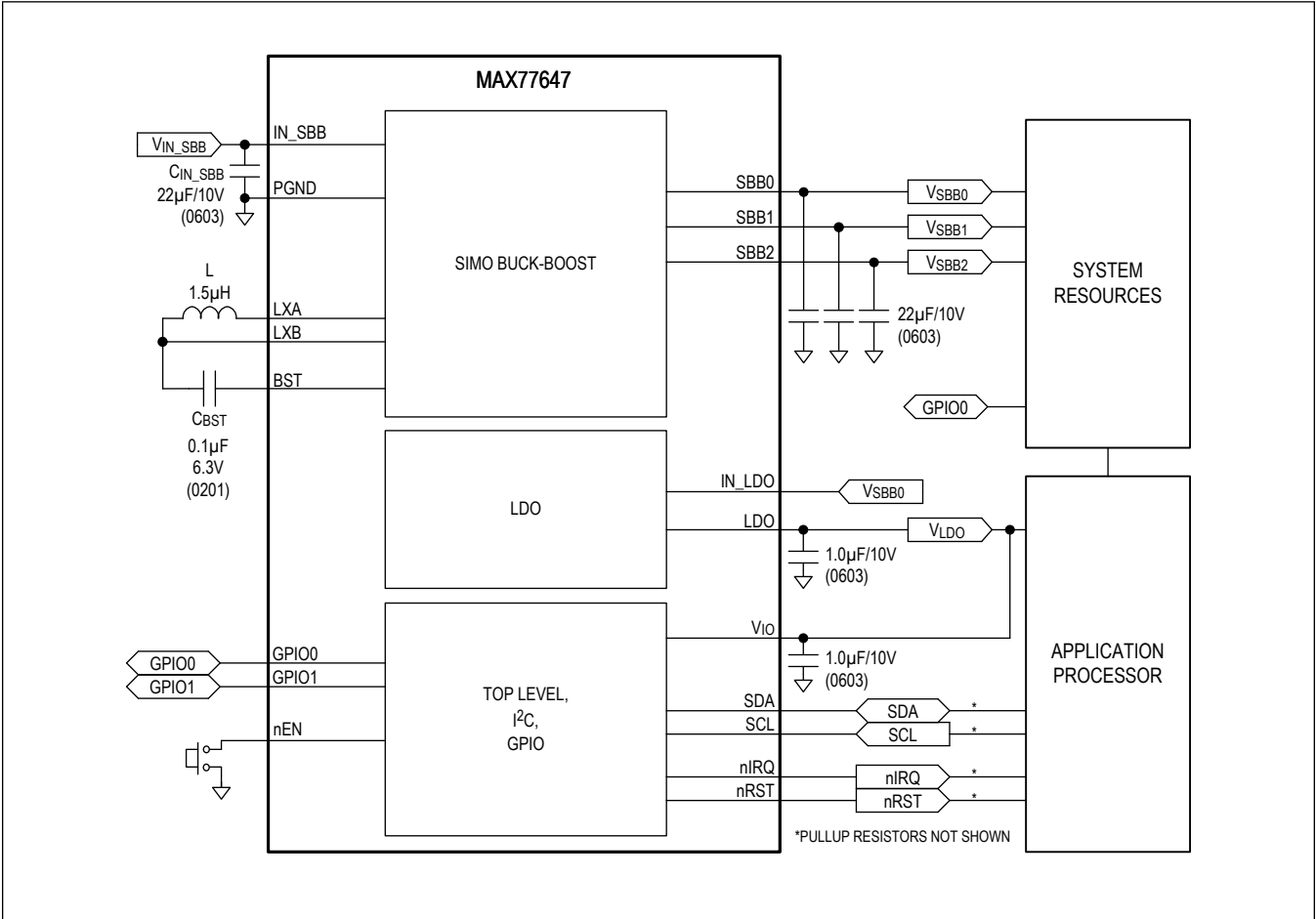
BITFIELD	BITS	DESCRIPTION	DECODE
LDO_MD	4	Operation Mode of LDO	0b0 = Low dropout linear regulator (LDO) mode 0b1 = Load switch (LSW) mode
ADE_LDO	3	LDO Active-Discharge Enable	0b0 = The active discharge function is disabled. When LDO is disabled, its discharge rate is a function of the output capacitance and the external load. 0b1 = The active discharge function is enabled. When LDO is disabled, an internal resistor (RAD_LDO) is activated from LDO to GND to help the output voltage discharge. The output voltage discharge rate is a function of the output capacitance, the external loading, and the internal RAD_LDO load.
EN_LDO	2:0	Enable Control for LDO. Select the FPS slot that the channel powers up and powers down in or whether the channel is forced on or off.	0b000 = FPS slot 0 0b001 = FPS slot 1 0b010 = FPS slot 2 0b011 = FPS slot 3 0b100 = Off irrespective of FPS 0b101 = Same as 0b100 0b110 = On irrespective of FPS 0b111 = Same as 0b110

## Typical Application Circuits

## Typical Applications Circuits



Typical Application Circuits (continued)



Ordering Information

PART	TEMP RANGE	PIN-PACKAGE	OPTIONS
MAX77647AANP+T	-20°C to +85°C	20 WLP	<a href="#">Table 3</a>
MAX77646ANP+T	-20°C to +85°C	20 WLP	—
MAX77647ANP+*	-20°C to +85°C	20 WLP	—

+Denotes a lead(Pb)-free/RoHS-compliant package.

T = Tape and reel.

\*Custom samples only. Not for production or stock. Contact factory for more information.

MAX77646/MAX77647

# Ultra-Configurable SIMO PMIC Featuring 3-Output Buck-Boost, 1-LDO for Long Battery Life Primary Cell Applications

## Revision History

REVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	6/23	Release for Market Intro	—