

# ADXL356/ADXL357/ADXL357B

## Low Noise, Low Drift, Low Power, 3-Axis MEMS Accelerometers

### FEATURES

- ▶ 0 g offset vs. temperature (all axes): 0.75 mg/°C maximum
- ▶ Ultralow noise spectral density (all axes): 75  $\mu\text{g}/\sqrt{\text{Hz}}$
- ▶ Low power,  $V_{\text{SUPPLY}}$  (LDO regulator enabled)
  - ▶ ADXL356 in measurement mode: 150  $\mu\text{A}$
  - ▶ ADXL357/ADXL357B in measurement mode: 200  $\mu\text{A}$
  - ▶ ADXL356/ADXL357/ADXL357B in standby mode: 21  $\mu\text{A}$
- ▶ ADXL356 has user adjustable analog output bandwidth
- ▶ ADXL357/ADXL357B digital output features
  - ▶ Digital SPI and limited I<sup>2</sup>C interfaces supported
  - ▶ 20-bit ADC
  - ▶ Data interpolation routine for synchronous sampling
  - ▶ Programmable high- and low-pass digital filters
- ▶ Integrated temperature sensor
- ▶ Voltage range options
  - ▶  $V_{\text{SUPPLY}}$  with internal regulators: 2.25 V to 3.6 V
  - ▶  $V_{1P8\text{ANA}}$ ,  $V_{1P8\text{DIG}}$  with internal LDO regulator bypassed: 1.8 V typical  $\pm 10\%$
- ▶ Operating temperature range: -40°C to +125°C
- ▶ 14-terminal, 6 mm × 5.6 mm × 2.2 mm, LCC package

### APPLICATIONS

- ▶ Inertial measurement units (IMUs)/attitude and heading reference systems (AHRs)
- ▶ Platform stabilization systems
- ▶ Structural health monitoring
- ▶ Seismic imaging
- ▶ Tilt sensing
- ▶ Robotics
- ▶ Condition monitoring

### FUNCTIONAL BLOCK DIAGRAMS

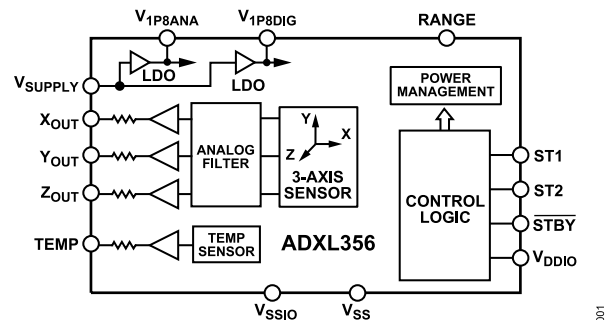


Figure 1. ADXL356

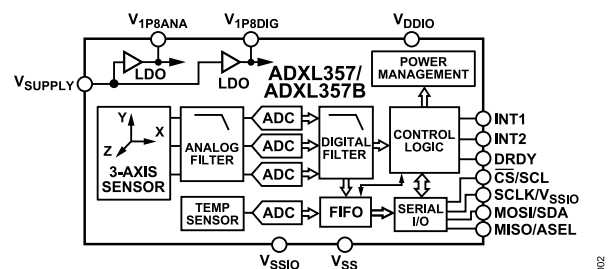


Figure 2. ADXL357 and ADXL357B

### GENERAL DESCRIPTION

The analog output ADXL356 and the digital outputs ADXL357 and ADXL357B are low noise density, low 0 g offset drift, low power, 3-axis accelerometers with selectable measurement ranges. The ADXL356B supports the  $\pm 10\text{ g}$  and  $\pm 20\text{ g}$  ranges, the ADXL356C supports the  $\pm 10\text{ g}$  and  $\pm 40\text{ g}$  ranges, and the ADXL357 and ADXL357B support the  $\pm 10\text{ g}$ ,  $\pm 20\text{ g}$ , and  $\pm 40\text{ g}$  ranges.

The ADXL356/ADXL357/ADXL357B<sup>1</sup> offer industry leading noise, minimal offset drift over temperature, and long-term stability, enabling precision applications with minimal calibration.

The low drift, low noise, and low power ADXL357 and ADXL357B enable accurate tilt measurement. The ADXL357B is more robust/suitable for tilt sensing in high vibration environments. The low noise of the ADXL356 over higher frequencies is ideal for condition-based monitoring and other vibration sensing applications.

The ADXL357 and ADXL357B multifunction pin names may be referenced only by their relevant function for either the serial peripheral interface (SPI) or limited I<sup>2</sup>C interface.

Analog Devices is in the process of updating documentation to provide terminology and language that is culturally appropriate. This is a process with a wide scope and will be phased in as quickly as possible. Thank you for your patience.

<sup>1</sup> Protected by U.S. Patents 8,472,270; 9,041,462; 8,665,627; 8,917,099; 6,892,576; 9,297,825; and 7,956,621.

### Rev. B

#### DOCUMENT FEEDBACK

#### TECHNICAL SUPPORT

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**REVISION HISTORY****3/2024—Rev. A to Rev. B**

Added ADXL357B (Universal).....	1
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## SPECIFICATIONS

## ADXL356 (ANALOG OUTPUT)

$T_A = 25^\circ\text{C}$ ,  $V_{\text{SUPPLY}} = 3.3\text{ V}$ , x-axis acceleration and y-axis acceleration = 0 g, z-axis acceleration = 1 g, and full-scale range =  $\pm 10\text{ g}$ , unless otherwise noted.

Table 1.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
<b>SENSOR INPUT</b>					
Output Full-Scale Range (FSR)	Each axis				
	ADXL356B supports two ranges		$\pm 10, \pm 20$		g
Resonant Frequency <sup>1</sup>	ADXL356C supports two ranges		$\pm 10, \pm 40$		g
	X-axis and y-axis		5.5		kHz
Quality Factor	Z-axis		5.1		kHz
	X-axis		17		
	Y-axis		14		
Nonlinearity	Z-axis		4		
	$\pm 10\text{ g}$		0.1		% FSR
Cross Axis Sensitivity	$\pm 40\text{ g}$		1.3		% FSR
			1		%
<b>SENSITIVITY</b>					
Sensitivity at $X_{\text{OUT}}$ , $Y_{\text{OUT}}$ , $Z_{\text{OUT}}$	Ratiometric to $V_{1P8ANA}$				
	$\pm 10\text{ g}$	73.6	80	86.4	mV/g
	$\pm 20\text{ g}$	36.8	40	43.2	mV/g
	$\pm 40\text{ g}$	18.4	20	21.6	mV/g
Sensitivity Change Due to Temperature	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		$\pm 0.01$		%/ $^\circ\text{C}$
Repeatability <sup>2</sup>	X-axis and y-axis		0.1		%
	Z-axis		0.2		%
<b>0 g OFFSET</b>					
0 g Output for $X_{\text{OUT}}$ , $Y_{\text{OUT}}$ , $Z_{\text{OUT}}$	Each axis, $\pm 10\text{ g}$				
	Referred to $V_{1P8ANA}/2$	-375	$\pm 125$	+375	mg
0 g Offset vs. Temperature (X-Axis, Y-Axis, and Z-Axis) <sup>3</sup>	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	-0.75	$\pm 0.2$	+0.75	mg/ $^\circ\text{C}$
Repeatability <sup>2</sup>	X-axis and y-axis		$\pm 4.25$		mg
	Z-axis		$\pm 5$		mg
Vibration Rectification Error (VRE) <sup>4</sup>	Offset due to 7.5 g rms vibration, $\pm 10\text{ g}$ range, in a 1 g orientation		<0.1		g
<b>NOISE</b>					
Spectral Density <sup>5</sup>					
	X-Axis, Y-Axis, and Z-Axis	$\pm 10\text{ g}$	75		$\mu\text{g}/\sqrt{\text{Hz}}$
Velocity Random Walk		$\pm 40\text{ g}$	110		$\mu\text{g}/\sqrt{\text{Hz}}$
	X-Axis and Y-Axis	$\pm 10\text{ g}$	38.2		mm/sec/ $\sqrt{\text{hr}}$
Z-Axis			26.5		mm/sec/ $\sqrt{\text{hr}}$
BANDWIDTH	-3 dB, overall transfer function <sup>6</sup>		2.4		kHz
<b>SELF TEST</b>					
Output Change					
	Z-Axis	$\pm 10\text{ g}$ range <sup>7</sup>	0.5	1.25	3.0
POWER SUPPLY					
	Voltage Range				
$V_{\text{SUPPLY}}$ <sup>8</sup>		2.25	2.5	3.6	V
	$V_{\text{DDIO}}$	$V_{1P8DIG}$	2.5	3.6	V
$V_{1P8ANA}$ , $V_{1P8DIG}$	Internal low dropout (LDO) regulator bypassed, $V_{\text{SUPPLY}} = 0\text{ V}$	1.62	1.8	1.98	V

## SPECIFICATIONS

Table 1. (Continued)

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
Current					
Measurement Mode					
V <sub>SUPPLY</sub>	LDO regulator enabled		150		μA
V <sub>1P8ANA</sub>	LDO regulator disabled		138		μA
V <sub>1P8DIG</sub>	LDO regulator disabled		12		μA
Standby Mode					
V <sub>SUPPLY</sub>	LDO regulator enabled		21		μA
V <sub>1P8ANA</sub>	LDO regulator disabled		7		μA
V <sub>1P8DIG</sub>	LDO regulator disabled		10		μA
Turn On Time <sup>9</sup>	10 g range		10		ms
	Power-off to standby		<10		ms
OUTPUT AMPLIFIER	X <sub>OUT</sub> , Y <sub>OUT</sub> , Z <sub>OUT</sub> , and TEMP pins				
Swing	No load	0.03		V <sub>1P8ANA</sub> - 0.03	V
Output Series Resistance			32		kΩ
TEMPERATURE SENSOR					
Output at 25°C			967		mV
Standard Deviation			6		mV
Scale Factor			3.0		mV/°C
TEMPERATURE					
Operating Temperature Range		-40		+125	°C

<sup>1</sup> The resonant frequency is a sensor characteristic.

<sup>2</sup> Repeatability is predicted for a 10 year life and includes shifts due to the high temperature operating life (HTOL) (T<sub>A</sub> = 150°C, V<sub>SUPPLY</sub> = 3.6 V, and 1000 hours), temperature cycling (-55°C to +125°C and 1000 cycles), velocity random walk, broadband noise, and temperature hysteresis. Repeatability in relation to time follows the square root law. For example, to obtain offset repeatability of the x-axis for 2.5 years, use the following equation:  $\pm 4.25 \text{ mg} \times \sqrt{(2.5 \text{ years}/10 \text{ years})} = \pm 2.125 \text{ mg}$ .

<sup>3</sup> The temperature change is -40°C to +25°C, or +25°C to +125°C.

<sup>4</sup> The VRE measurement is the shift in dc offset while the device is subject to 7.5 g rms of random vibration from 50 Hz to 2 kHz. The device under test (DUT) is configured for the ±10 g range and an output data rate of 4 kHz. The VRE scales with the range setting.

<sup>5</sup> Based on characterization.

<sup>6</sup> Overall transfer function includes the sensor mechanical response and all other filters on the signal chain.

<sup>7</sup> ±10 g indicates a test condition. The self test result converted to the acceleration value is independent of the selected range.

<sup>8</sup> When V<sub>1P8ANA</sub> and V<sub>1P8DIG</sub> are generated internally, V<sub>SUPPLY</sub> is valid. To disable the LDO regulator and drive V<sub>1P8ANA</sub> and V<sub>1P8DIG</sub> externally, connect V<sub>SUPPLY</sub> to V<sub>SS</sub>.

<sup>9</sup> Standby to measurement mode. This specification is valid when the output is within 5 mg of the final value.

## ADXL357 (DIGITAL OUTPUT)

T<sub>A</sub> = 25°C, V<sub>SUPPLY</sub> = 3.3 V, x-axis acceleration and y-axis acceleration = 0 g, z-axis acceleration = 1 g, full-scale range = ±10 g, and output data rate (ODR) = 500 Hz, unless otherwise noted. Note that multifunction pin names may be referenced only by their relevant function.

Table 2.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
SENSOR INPUT					
Output Full-Scale Range (FSR)	Each axis				
	User selectable, supports three ranges		±10		g
			±20		g
			±40		g
Resonant Frequency <sup>1</sup>	X-axis and y-axis		5.5		kHz
	Z-axis		5.1		kHz
Quality Factor	X-axis		18		
	Y-axis		15		

## SPECIFICATIONS

Table 2. (Continued)

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
Nonlinearity	Z-axis		4		
	$\pm 10\text{ g}$		0.1		% FSR
	$\pm 40\text{ g}$		1.3		% FSR
Cross Axis Sensitivity			1		%
SENSITIVITY <sup>2</sup>	Each axis				
X-Axis, Y-Axis, and Z-Axis Sensitivity	$\pm 10\text{ g}$	47,104	51,200	55,296	LSB/g
	$\pm 20\text{ g}$	23,552	25,600	27,648	LSB/g
	$\pm 40\text{ g}$	11,776	12,800	13,824	LSB/g
X-Axis, Y-Axis, and Z-Axis Scale Factor	$\pm 10\text{ g}$		19.5		$\mu\text{g/LSB}$
	$\pm 20\text{ g}$		39		$\mu\text{g/LSB}$
	$\pm 40\text{ g}$		78		$\mu\text{g/LSB}$
Sensitivity Change due to Temperature	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$		$\pm 0.01$		%/ $^\circ\text{C}$
Repeatability <sup>3</sup>	X-axis and y-axis		0.1		%
	Z-axis		0.2		%
0 g OFFSET	Each axis, $\pm 10\text{ g}$				
X-Axis, Y-Axis, and Z-Axis 0 g Output		-375	$\pm 125$	+375	mg
0 g Offset vs. Temperature (X-Axis, Y-Axis, and Z-Axis) <sup>4</sup>	$T_A = -40^\circ\text{C}$ to $+125^\circ\text{C}$	-0.75	$\pm 0.20$	+0.75	mg/ $^\circ\text{C}$
Repeatability <sup>3</sup>	X-axis and y-axis		$\pm 4.25$		mg
	Z-axis		$\pm 5$		mg
VRE <sup>5</sup>	Offset due to 7.5 g rms vibration, $\pm 10\text{ g}$ range, in a 1 g orientation		<0.1		g
NOISE					
Spectral Density <sup>6</sup>					
X-Axis, Y-Axis, and Z-Axis	$\pm 10\text{ g}$		75		$\mu\text{g}/\sqrt{\text{Hz}}$
	$\pm 40\text{ g}$		90		$\mu\text{g}/\sqrt{\text{Hz}}$
Velocity Random Walk	$\pm 10\text{ g}$				
X-Axis and Y-Axis			38.2		mm/sec/ $\sqrt{\text{hr}}$
Z-Axis			26.5		mm/sec/ $\sqrt{\text{hr}}$
BANDWIDTH AND OUTPUT DATA RATE					
Analog-to-Digital Converter (ADC) Resolution			20		Bits
Low-Pass Filter Passband Frequency	User programmable, Register 0x28	0.977		1000	Hz
High-Pass Filter Passband Frequency When Enabled (Disabled by Default)	User programmable, Register 0x28 for 4 kHz ODR	0.0095		10	Hz
SELF TEST					
Output Change					
Z-Axis	$\pm 10\text{ g}$ range <sup>7</sup>	0.5	1.25	3.0	g
POWER SUPPLY					
Voltage Range					
$V_{\text{SUPPLY}}$ Operating <sup>8</sup>		2.25	2.5	3.6	V
$V_{\text{DDIO}}$		$V_{\text{IP8DIG}}$	2.5	3.6	V
$V_{\text{IP8ANA}}$ and $V_{\text{IP8DIG}}$	Internal LDO regulator bypassed, $V_{\text{SUPPLY}} = 0\text{ V}$	1.62	1.8	1.98	V
Current					
Measurement Mode					
$V_{\text{SUPPLY}}$	LDO regulator enabled		200		$\mu\text{A}$
$V_{\text{IP8ANA}}$	LDO regulator disabled		160		$\mu\text{A}$
$V_{\text{IP8DIG}}$	LDO regulator disabled		35.5		$\mu\text{A}$
Standby Mode					
$V_{\text{SUPPLY}}$	LDO regulator enabled		21		$\mu\text{A}$
$V_{\text{IP8ANA}}$	LDO regulator disabled		7		$\mu\text{A}$

## SPECIFICATIONS

Table 2. (Continued)

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
$V_{1P8DIG}$	LDO regulator disabled		10		$\mu A$
Turn On Time <sup>9</sup>	$\pm 10 g$ range		<10		ms
	Power-off to standby		<10		ms
TEMPERATURE SENSOR					
Output at 25°C			1885		LSB
Standard Deviation			28.5		LSB
Scale Factor			-9.05		LSB/°C
TEMPERATURE					
Operating Temperature Range		-40		+125	°C

<sup>1</sup> The resonant frequency is a sensor characteristic.

<sup>2</sup> Characterized but not 100% tested.

<sup>3</sup> Repeatability is predicted for a 10 year life and includes shifts due to the HTOL ( $T_A = 150^\circ C$ ,  $V_{SUPPLY} = 3.6 V$ , and 1000 hours), temperature cycling ( $-55^\circ C$  to  $+125^\circ C$  and 1000 cycles), velocity random walk, broadband noise, and temperature hysteresis. Repeatability in relation to time follows the square root law. For example, to obtain offset repeatability of the x-axis for 2.5 years, use the following equation:  $\pm 4.25 mg \times \sqrt{(2.5 \text{ years}/10 \text{ years})} = \pm 2.125 mg$ .

<sup>4</sup> The temperature change is  $-40^\circ C$  to  $+25^\circ C$  or  $+25^\circ C$  to  $+125^\circ C$ .

<sup>5</sup> The VRE measurement is the shift in dc offset while the device is subject to 7.5 g rms random vibration from 50 Hz to 2 kHz. The DUT is configured for the  $\pm 10 g$  range and an output data rate of 4 kHz. The VRE scales with the range setting.

<sup>6</sup> Based on characterization.

<sup>7</sup>  $\pm 10 g$  indicates a test condition. The self test result converted to the acceleration value is independent of the selected range.

<sup>8</sup> When  $V_{1P8ANA}$  and  $V_{1P8DIG}$  are generated internally,  $V_{SUPPLY}$  is valid. To disable the LDO regulator and drive  $V_{1P8ANA}$  and  $V_{1P8DIG}$  externally, connect  $V_{SUPPLY}$  to  $V_{SS}$ .

<sup>9</sup> Standby to measurement mode. This specification is valid when the output is within 1 mg of final value.

## ADXL357B (DIGITAL OUTPUT)

$T_A = 25^\circ C$ ,  $V_{SUPPLY} = 3.3 V$ , x-axis acceleration and y-axis acceleration = 0 g, z-axis acceleration = 1 g, full-scale range =  $\pm 10 g$ , and ODR = 500 Hz, unless otherwise noted. Note that multifunction pin names may be referenced only by their relevant function.

Table 3.

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
SENSOR INPUT	Each axis				
Output Full-Scale Range (FSR)	User selectable, supports three ranges		$\pm 10$		g
			$\pm 20$		g
			$\pm 40$		g
Resonant Frequency <sup>1</sup>	X-axis and y-axis		5.3		kHz
	Z-axis		4.8		KHz
Quality Factor (Q)	X-axis and y-axis		1.9		
	Z-axis		0.7		
Nonlinearity	$\pm 10 g$		0.1		%FSR
	$\pm 40 g$		1.6		%FSR
Cross Axis Sensitivity			1		%
SENSITIVITY <sup>2</sup>	Each axis				
X-Axis, Y-Axis, and Z-Axis Sensitivity	$\pm 10 g$	47,104	51,200	55,296	LSB/g
	$\pm 20 g$	23,552	25,600	27,648	LSB/g
	$\pm 40 g$	11,520	12,800	14,080	LSB/g
X-Axis, Y-Axis, and Z-Axis Scale Factor	$\pm 10 g$		19.5		$\mu g/LSB$
	$\pm 20 g$		39		$\mu g/LSB$
	$\pm 40 g$		78		$\mu g/LSB$
Sensitivity Change due to Temperature	$T_A = -40^\circ C$ to $+125^\circ C$		$\pm 0.01$		%/°C

## SPECIFICATIONS

Table 3. (Continued)

Parameter	Test Conditions/Comments	Min	Typ	Max	Unit
Repeatability <sup>3</sup>	X-axis and y-axis		0.1		%
	Z-axis		0.2		%
0 g OFFSET	Each axis, $\pm 10$ g				
X-Axis, Y-Axis, and Z-Axis 0 g Output		-375	$\pm 125$	+375	mg
0 g Offset vs. Temperature (X-Axis, Y-Axis, and Z-Axis) <sup>4</sup>	$T_A = -40^{\circ}\text{C}$ to $+125^{\circ}\text{C}$	-0.75	$\pm 0.20$	+0.75	mg/ $^{\circ}\text{C}$
Repeatability <sup>3</sup>	X-axis and y-axis		$\pm 4.25$		mg
	Z-axis		$\pm 5$		mg
VRE <sup>5</sup>	Offset due to 12.5 g rms vibration, $\pm 40$ g range, in a 1 g orientation		<0.1		g
NOISE					
Spectral Density <sup>6</sup>					
X-Axis, Y-Axis, and Z-Axis	$\pm 10$ g		75		$\mu\text{g}/\sqrt{\text{Hz}}$
	$\pm 40$ g		85		$\mu\text{g}/\sqrt{\text{Hz}}$
Velocity Random Walk	$\pm 40$ g				
X-Axis and Y-Axis			65		mm/sec/ $\sqrt{\text{hr}}$
Z-Axis			12		mm/sec/ $\sqrt{\text{hr}}$
BANDWIDTH AND OUTPUT DATA RATE					
Analog-to-Digital Converter (ADC) Resolution			20		Bits
Low-Pass Filter Passband Frequency	User programmable, Register 0x28	1.1		942	Hz
High-Pass Filter Passband Frequency when Enabled (Disabled by Default)	User programmable, Register 0x28 for 4 kHz ODR	0.0095		10	Hz
SELF TEST					
Output Change					
X-axis and Y-axis	$\pm 40$ g range <sup>7</sup>	0.18	0.33	0.48	g
Z-Axis		0.875	1.25	1.825	g
POWER SUPPLY					
Voltage Range					
$V_{\text{SUPPLY}}$ Operating <sup>8</sup>		2.25	3.3	3.6	V
$V_{\text{DDIO}}$		$V_{1\text{P8DIG}}$	3.3	3.6	V
$V_{1\text{P8ANA}}$ and $V_{1\text{P8DIG}}$	Internal LDO regulator bypassed, $V_{\text{SUPPLY}} = 0$ V	1.62	1.8	1.98	V
Current					
Measurement Mode					
$V_{\text{SUPPLY}}$	LDO regulator enabled		200		$\mu\text{A}$
$V_{1\text{P8ANA}}$	LDO regulator disabled		160		$\mu\text{A}$
$V_{1\text{P8DIG}}$	LDO regulator disabled		35.5		$\mu\text{A}$
Standby Mode					
$V_{\text{SUPPLY}}$	LDO regulator enabled		21		$\mu\text{A}$
$V_{1\text{P8ANA}}$	LDO regulator disabled		7		$\mu\text{A}$
$V_{1\text{P8DIG}}$	LDO regulator disabled		10		$\mu\text{A}$
Turn On Time <sup>9</sup>	$\pm 10$ g range		<10		ms
	Power-off to standby		<10		ms
TEMPERATURE SENSOR					
Output at $25^{\circ}\text{C}$			1829		LSB
Standard Deviation			24.3		LSB
Scale Factor			-9.05		LSB/ $^{\circ}\text{C}$
TEMPERATURE					
Operating Temperature Range		-40		+125	$^{\circ}\text{C}$



## SPECIFICATIONS

- <sup>1</sup> The resonant frequency is a sensor characteristic.
- <sup>2</sup> Characterized but not 100% tested.
- <sup>3</sup> Repeatability is predicted for a 10 year life and includes shifts due to the HTOL ( $T_A = 150^{\circ}\text{C}$ ,  $V_{\text{SUPPLY}} = 3.6\text{ V}$ , and 1000 hours), temperature cycling ( $-55^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$  and 1000 cycles), velocity random walk, broadband noise, and temperature hysteresis. Repeatability in relation to time follows the square root law. For example, to obtain offset repeatability of the x-axis for 2.5 years, use the following equation:  $\pm 4.25\text{ mg} \times \sqrt{(2.5\text{ years}/10\text{ years})} = \pm 2.125\text{ mg}$ .
- <sup>4</sup> The temperature change is  $-40^{\circ}\text{C}$  to  $+25^{\circ}\text{C}$  or  $+25^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ .
- <sup>5</sup> The VRE measurement is the shift in dc offset while the device is subject to 12.5 g rms random vibration from 50 Hz to 2 kHz. The DUT is configured for the  $\pm 40\text{ g}$  range and an output data rate of 4 kHz. The VRE scales with the range setting.
- <sup>6</sup> Based on characterization.
- <sup>7</sup>  $\pm 40\text{ g}$  indicates a test condition. The self test result converted to the acceleration value is independent of the selected range.
- <sup>8</sup> When  $V_{1\text{P8ANA}}$  and  $V_{1\text{P8DIG}}$  are generated internally,  $V_{\text{SUPPLY}}$  is valid. To disable the LDO regulator and drive  $V_{1\text{P8ANA}}$  and  $V_{1\text{P8DIG}}$  externally, connect  $V_{\text{SUPPLY}}$  to  $V_{\text{SS}}$ .
- <sup>9</sup> Standby to measurement mode. This specification is valid when the output is within 1 mg of final value.

## SPI DIGITAL INTERFACE CHARACTERISTICS FOR THE ADXL357 AND ADXL357B

Note that multifunction pin names may be referenced by their relevant function only.

Table 4.

Parameter	Symbol	Test Conditions/Comments	Min	Typ	Max	Unit
DC INPUT LEVELS						
Input Voltage						
Low Level	$V_{\text{IL}}$				$0.3 \times V_{\text{DDIO}}$	V
High Level	$V_{\text{IH}}$		$0.7 \times V_{\text{DDIO}}$			V
Input Current						
Low Level	$I_{\text{IL}}$	Input voltage ( $V_{\text{IN}} = 0\text{ V}$ )	-0.2			$\mu\text{A}$
High Level	$I_{\text{IH}}$	$V_{\text{IN}} = V_{\text{DDIO}}$			0.2	$\mu\text{A}$
DC OUTPUT LEVELS						
Output Voltage						
Low Level	$V_{\text{OL}}$	$I_{\text{OL}} = I_{\text{OL, MIN}}$			$0.2 \times V_{\text{DDIO}}$	V
High Level	$V_{\text{OH}}$	$I_{\text{OH}} = I_{\text{OH, MAX}}$	$0.8 \times V_{\text{DDIO}}$			V
Output Current						
Low Level	$I_{\text{OL}}$	$V_{\text{OL}} = V_{\text{OL, MAX}}$	-10			mA
High Level	$I_{\text{OH}}$	$V_{\text{OH}} = V_{\text{OH, MIN}}$			4	mA
AC INPUT LEVELS						
SCLK Frequency			0.1		10	MHz
SCLK High Time	$t_{\text{HIGH}}$		40			ns
SCLK Low Time	$t_{\text{LOW}}$		40			ns
$\overline{\text{CS}}$ Setup Time	$t_{\text{CSS}}$		20			ns
$\overline{\text{CS}}$ Hold Time	$t_{\text{CSH}}$		20			ns
$\overline{\text{CS}}$ Disable Time	$t_{\text{CSD}}$		40			ns
Rising SCLK Setup Time	$t_{\text{SCLKS}}$		20			ns
MOSI Setup Time	$t_{\text{SU}}$		20			ns
MOSI Hold Time	$t_{\text{HD}}$		20			ns
AC OUTPUT LEVELS						
Propagation Delay	$t_{\text{P}}$	Load capacitance ( $C_{\text{LOAD}} = 30\text{ pF}$ )			30	ns
Enable MISO Time	$t_{\text{EN}}$		30			ns
Disable MISO Time	$t_{\text{DIS}}$				20	ns

## SPECIFICATIONS

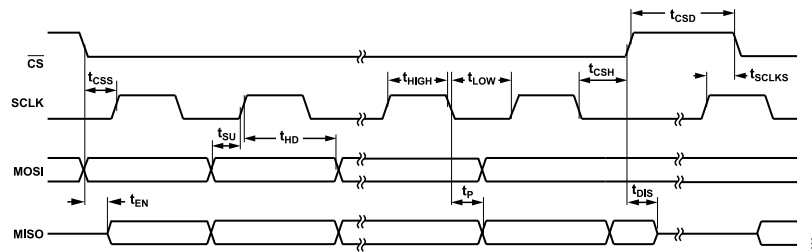


Figure 3. SPI Interface Timing Diagram

I<sup>2</sup>C DIGITAL INTERFACE CHARACTERISTICS FOR THE ADXL357 AND ADXL357B

Note that multifunction pin names may be referenced only by their relevant function.

Table 5.

Parameter	Symbol	Test Conditions/ Comments	I2C_HS = 0 (Fast Mode)			I2C_HS = 1 (High Speed Mode)			Unit
			Min	Typ	Max	Min	Typ	Max	
DC INPUT LEVELS									
Input Voltage									
Low Level	V <sub>IL</sub>				0.3 × V <sub>DDIO</sub>			0.3 × V <sub>DDIO</sub>	V
High Level	V <sub>IH</sub>		0.7 × V <sub>DDIO</sub>			0.7 × V <sub>DDIO</sub>			V
Hysteresis of Schmitt Triggered Inputs	V <sub>HYS</sub>		0.05 × V <sub>DDIO</sub>			0.1 × V <sub>DDIO</sub>			V
Input Current	I <sub>IL</sub>	0.1 × V <sub>DDIO</sub> < V <sub>IN</sub> < 0.9 × V <sub>DDIO</sub>	−10		+10				μA
DC OUTPUT LEVELS									
Output Voltage		I <sub>OL</sub> = 3 mA							
Low Level	V <sub>OL1</sub> V <sub>OL2</sub>	V <sub>DDIO</sub> > 2 V V <sub>DDIO</sub> ≤ 2 V			0.4 0.2 × V <sub>DDIO</sub>			0.4 0.2 × V <sub>DDIO</sub>	V V
Output Current									
Low Level	I <sub>OL</sub>	V <sub>OL</sub> = 0.4 V V <sub>OL</sub> = 0.6 V	20 6			20 6			mA mA
AC INPUT LEVELS									
SCL Frequency			0		1	0		3.4	MHz
SCL High Time	t <sub>HIGH</sub>		260			60			ns
SCL Low Time	t <sub>LOW</sub>		500			160			ns
Start Setup Time	t <sub>SUSTA</sub>		260			160			ns
Start Hold Time	t <sub>HDSTA</sub>		260			160			ns
SDA Setup Time	t <sub>SUDAT</sub>		50			10			ns
SDA Hold Time	t <sub>HDDAT</sub>		0			0			ns
Stop Setup Time	t <sub>SUSTO</sub>		260			160			ns
Bus Free Time	t <sub>BUF</sub>		500						ns
SCL Input Rise Time	t <sub>RCL</sub>				120			80	ns
SCL Input Fall Time	t <sub>FCL</sub>				120			80	ns
SDA Input Rise Time	t <sub>RDA</sub>				120			160	ns
SDA Input Fall Time	t <sub>FDA</sub>				120			160	ns
Width of Spikes to Suppress	t <sub>SP</sub>	Not shown in <a href="#">Figure 4</a>			50			10	ns
AC OUTPUT LEVELS									
Propagation Delay		C <sub>LOAD</sub> = 500 pF							
Data	t <sub>VDDAT</sub>		97		450	27		135	ns
Acknowledge	t <sub>VDAK</sub>				450				ns

SPECIFICATIONS

Table 5. (Continued)

Parameter	Symbol	Test Conditions/ Comments	I2C_HS = 0 (Fast Mode)			I2C_HS = 1 (High Speed Mode)			Unit
			Min	Typ	Max	Min	Typ	Max	
Output Fall Time	t <sub>F</sub>	Not shown in Figure 4	20 × (V <sub>DDIO</sub> /5.5)		120				ns

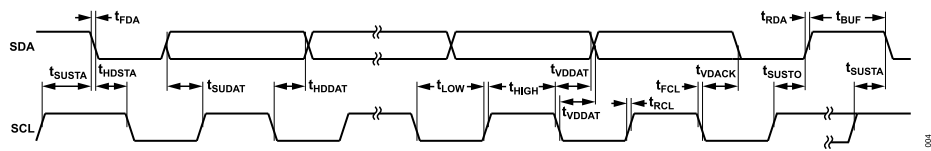


Figure 4. I<sup>2</sup>C Interface Timing Diagram

## ABSOLUTE MAXIMUM RATINGS

Table 6.

Parameter	Rating
Half Sine-Wave Shock (Any Axis)	
Unpowered	10,000 g, 0.1 ms shock duration
Powered	10,000 g, 0.1 ms shock duration
Vibration (Any Axis)	Per MIL-STD-883 Method 2007, Test Condition C <sup>1</sup>
V <sub>SUPPLY</sub> , V <sub>DDIO</sub>	5.4 V
V <sub>1P8ANA</sub> , V <sub>1P8DIG</sub> Configured as Inputs	1.98 V
ADXL356	
Digital Inputs (RANGE, ST1, ST2, $\overline{\text{STBY}}$ )	-0.3 V to V <sub>DDIO</sub> + 0.3 V
Analog Outputs (X <sub>OUT</sub> , Y <sub>OUT</sub> , Z <sub>OUT</sub> , TEMP)	-0.3 V to V <sub>1P8ANA</sub> + 0.3 V
ADXL357/ADXL357B	
Digital Pins ( $\overline{\text{CS}}$ , SCL, SCLK/V <sub>SSIO</sub> , MOSI/SDA, MISO/ASEL, INT1, INT2, DRDY)	-0.3 V to V <sub>DDIO</sub> + 0.3 V
Operating Temperature Range	-40°C to +125°C
Storage Temperature Range	-55°C to +150°C

<sup>1</sup> Sine-wave excitation swept logarithmically from 20 Hz to 2 kHz, 70 g peak amplitude. Sweep duration of 4 minutes. Repeat sweep four times per axis.

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

## THERMAL RESISTANCE

Thermal performance is directly linked to printed circuit board (PCB) design and operating environment. Careful attention to PCB thermal design is required.

$\theta_{JA}$  is the natural convection junction to ambient thermal resistance measured in a one cubic foot sealed enclosure.  $\psi_{JB}$  is the junction to board thermal resistance.

Table 7. Thermal Resistance

Package Type	$\theta_{JA}$	$\psi_{JB}$	Unit
E-14-1 <sup>1</sup>	42	17.6	°C/W

<sup>1</sup> Thermal impedance simulated values are based on a JEDEC 2S2P thermal test board with four thermal vias. See JEDEC JESD51.

## RECOMMENDED SOLDERING PROFILE

Figure 5 and Table 8 provide details about the recommended soldering profile.

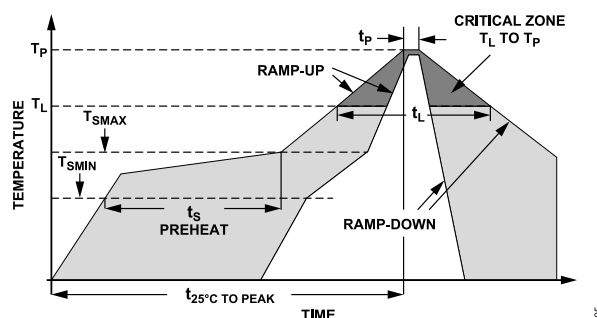


Figure 5. Recommended Soldering Profile

Table 8. Recommended Soldering Profile

Profile Feature	Condition	
	Sn63/Pb37	Pb-Free
Average Ramp Rate from Liquid Temperature (T <sub>L</sub> ) to Peak Temperature (T <sub>P</sub> )	3°C/sec maximum	3°C/sec maximum
Preheat		
Minimum Temperature (T <sub>SMIN</sub> )	100°C	150°C
Maximum Temperature (T <sub>SMAX</sub> )	150°C	200°C
Time from T <sub>SMIN</sub> to T <sub>SMAX</sub> (t <sub>s</sub> )	60 sec to 120 sec	60 sec to 180 sec
T <sub>SMAX</sub> to T <sub>L</sub> Ramp-Up Rate	3°C/sec maximum	3°C/sec maximum
Liquid Temperature (T <sub>L</sub> )	183°C	217°C
Time Maintained Above T <sub>L</sub> (t <sub>L</sub> )	60 sec to 150 sec	60 sec to 150 sec
Peak Temperature (T <sub>P</sub> )	240°C + 0°C/-5°C	260°C + 0°C/-5°C
Time of Actual T <sub>P</sub> - 5°C (t <sub>p</sub> )	10 sec to 30 sec	20 sec to 40 sec
Ramp-Down Rate	6°C/sec maximum	6°C/sec maximum
Time from 25°C to Peak Temperature (t <sub>25°C TO PEAK</sub> )	6 minutes maximum	8 minutes maximum

## ESD CAUTION



**ESD (electrostatic discharge) sensitive device.** Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

## PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

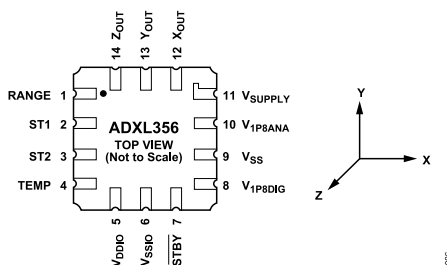


Figure 6. ADXL356 Pin Configuration

Table 9. ADXL356 Pin Function Descriptions

Pin No.	Mnemonic	Description
1	RANGE	Range Selection Pin. Set this pin to ground to select the $\pm 10\text{ g}$ range, or set this pin to $V_{DDIO}$ to select the $\pm 20\text{ g}$ or $\pm 40\text{ g}$ range. This pin is model dependent (see the <a href="#">Ordering Guide</a> section).
2	ST1	Self Test Pin 1. This pin enables self test mode. This pin must be forced low when not in self test mode.
3	ST2	Self Test Pin 2. This pin activates electromechanical self test actuation. This pin must be forced low when not in self test mode.
4	TEMP	Temperature Sensor Output.
5	$V_{DDIO}$	Digital Interface Supply Voltage.
6	$V_{SSIO}$	Digital Ground.
7	$\overline{\text{STBY}}$	Standby or Measurement Mode Selection Pin. Set this pin to ground to enter standby mode, or set this pin to $V_{DDIO}$ to enter measurement mode.
8	$V_{1P8DIG}$	Digital Supply. This pin requires a decoupling capacitor. If $V_{SUPPLY}$ connects to $V_{SS}$ , supply the voltage to this pin externally.
9	$V_{SS}$	Analog Ground.
10	$V_{1P8ANA}$	Analog Supply. This pin requires a decoupling capacitor. If $V_{SUPPLY}$ connects to $V_{SS}$ , supply the voltage to this pin externally.
11	$V_{SUPPLY}$	Supply Voltage. When $V_{SUPPLY}$ equals 2.25 V to 3.6 V, $V_{SUPPLY}$ enables the internal LDO regulators to generate $V_{1P8DIG}$ and $V_{1P8ANA}$ . For $V_{SUPPLY} = V_{SS}$ , $V_{1P8DIG}$ and $V_{1P8ANA}$ are externally supplied.
12	$X_{OUT}$	X-Axis Output.
13	$Y_{OUT}$	Y-Axis Output.
14	$Z_{OUT}$	Z-Axis Output.

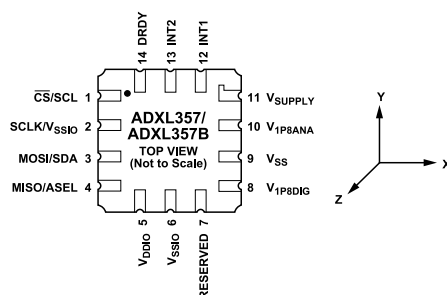
Figure 7. ADXL357 and ADXL357B Pin Configuration (SPI/I<sup>2</sup>C)

Table 10. ADXL357 and ADXL357B Pin Function Descriptions

Pin No.	Mnemonic	Description
1	$\overline{\text{CS}}/\text{SCL}$	Chip Select for SPI ( $\overline{\text{CS}}$ ). Serial Communications Clock for I <sup>2</sup> C (SCL).
2	$\text{SCLK}/V_{SSIO}$	Serial Communications Clock for SPI (SCLK). I <sup>2</sup> C Mode Enable ( $V_{SSIO}$ ). Connect this pin to Pin 6 ( $V_{SSIO}$ ) to enable I <sup>2</sup> C mode.
3	$\text{MOSI}/\text{SDA}$	Master Output, Slave Input for SPI (MOSI). Serial Data for I <sup>2</sup> C (SDA).
4	$\text{MISO}/\text{ASEL}$	Master Input, Slave Output for SPI (MISO). Alternate I <sup>2</sup> C Address Select for I <sup>2</sup> C (ASEL).

## PIN CONFIGURATIONS AND FUNCTION DESCRIPTIONS

Table 10. ADXL357 and ADXL357B Pin Function Descriptions (Continued)

Pin No.	Mnemonic	Description
5	V <sub>DDIO</sub>	Digital Interface Supply Voltage.
6	V <sub>SSIO</sub>	Digital Ground.
7	RESERVED	Reserved. This pin can be connected to ground or left open.
8	V <sub>1P8DIG</sub>	Digital Supply. This pin requires a decoupling capacitor. If V <sub>SUPPLY</sub> connects to V <sub>SS</sub> , supply the voltage to this pin externally.
9	V <sub>SS</sub>	Analog Ground.
10	V <sub>1P8ANA</sub>	Analog Supply. This pin requires a decoupling capacitor. If V <sub>SUPPLY</sub> connects to V <sub>SS</sub> , supply the voltage to this pin externally.
11	V <sub>SUPPLY</sub>	Supply Voltage. When V <sub>SUPPLY</sub> equals 2.25 V to 3.6 V, V <sub>SUPPLY</sub> enables the internal LDO regulators to generate V <sub>1P8DIG</sub> and V <sub>1P8ANA</sub> . For V <sub>SUPPLY</sub> = V <sub>SS</sub> , V <sub>1P8DIG</sub> and V <sub>1P8ANA</sub> are externally supplied.
12	INT1	Interrupt Pin 1.
13	INT2	Interrupt Pin 2.
14	DRDY	Data Ready Pin.

## TYPICAL PERFORMANCE CHARACTERISTICS

All figures include data for multiple devices and multiple lots, and they were taken in the  $\pm 10\text{ g}$  range and  $T_A = 25^\circ\text{C}$ , unless otherwise noted. For Figure 52 and Figure 76, the ODR is derived from a master clock, with a frequency of 1.024 MHz and  $\pm 1.4\%$  device to device variation (similar to ODR device to device variation). For a given device, however, clock frequency variation over the temperature range ( $-40^\circ\text{C}$  to  $+125^\circ\text{C}$ ) is no more than  $\pm 1.2\%$ , guaranteed by design.

## ADXL356

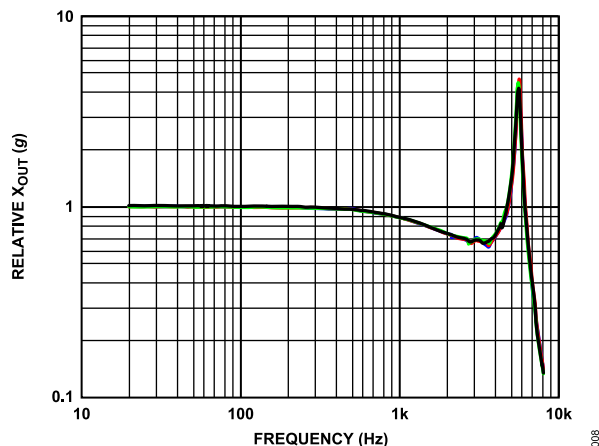


Figure 8. Frequency Response for X-Axis

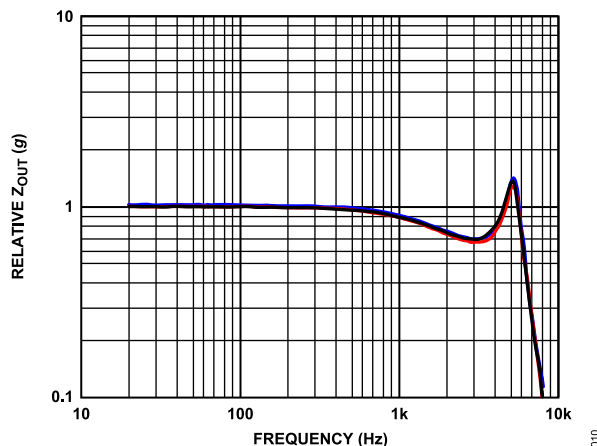


Figure 10. Frequency Response for Z-Axis

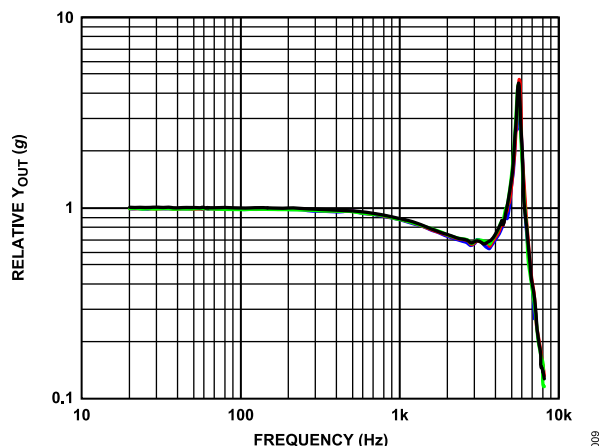


Figure 9. Frequency Response for Y-Axis

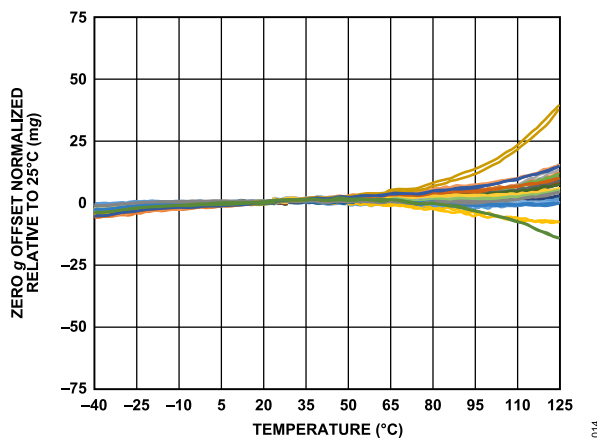


Figure 11. Zero g Offset Normalized Relative to 25°C vs. Temperature, X-Axis

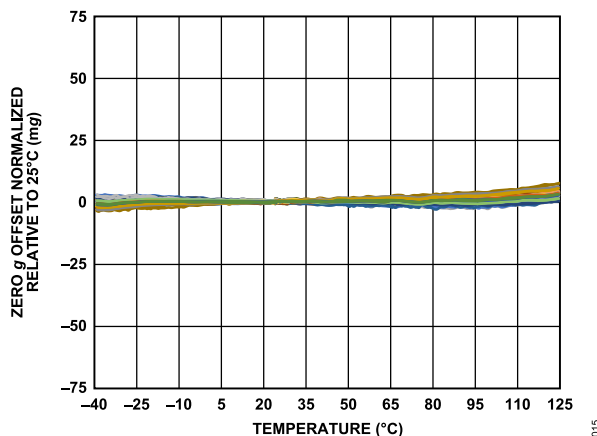


Figure 12. Zero g Offset Normalized Relative to 25°C vs. Temperature, Y-Axis

## TYPICAL PERFORMANCE CHARACTERISTICS

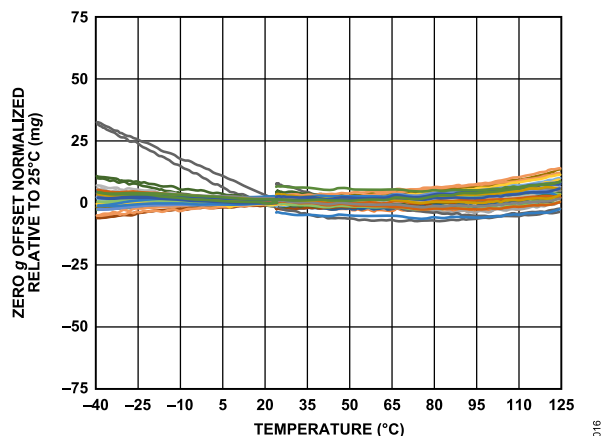


Figure 13. Zero g Offset Normalized Relative to 25°C vs. Temperature, Z-Axis

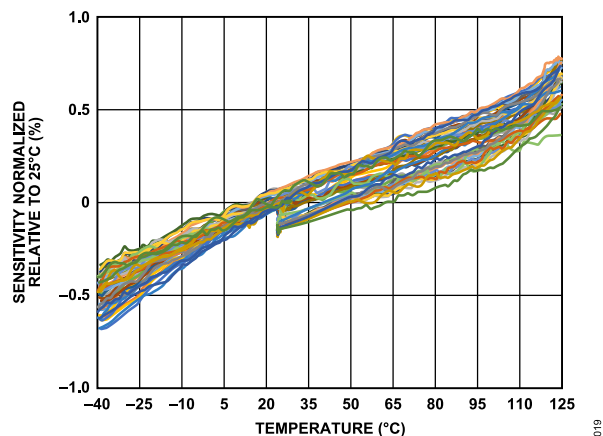


Figure 16. Sensitivity Normalized Relative to 25°C vs. Temperature, Z-Axis

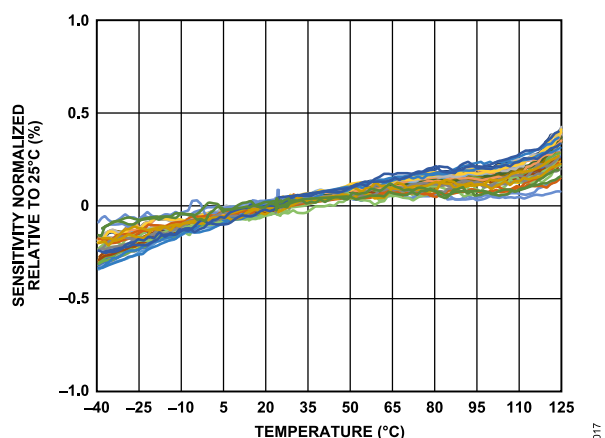


Figure 14. Sensitivity Normalized Relative to 25°C vs. Temperature X-Axis

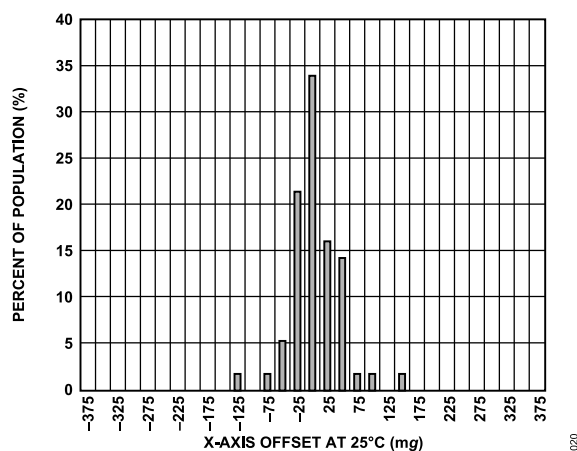


Figure 17. Zero g Offset Histogram at 25°C, X-Axis

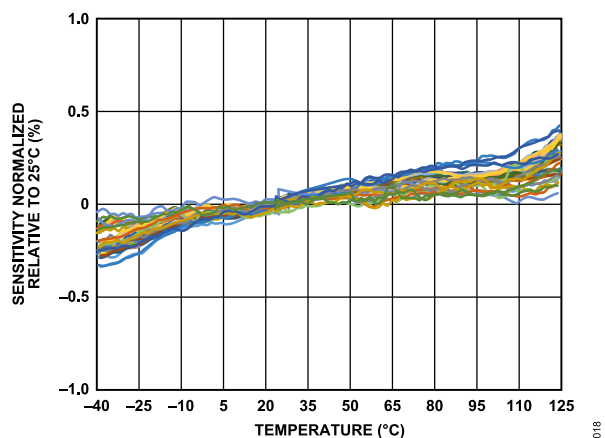


Figure 15. Sensitivity Normalized Relative to 25°C vs. Temperature, Y-Axis

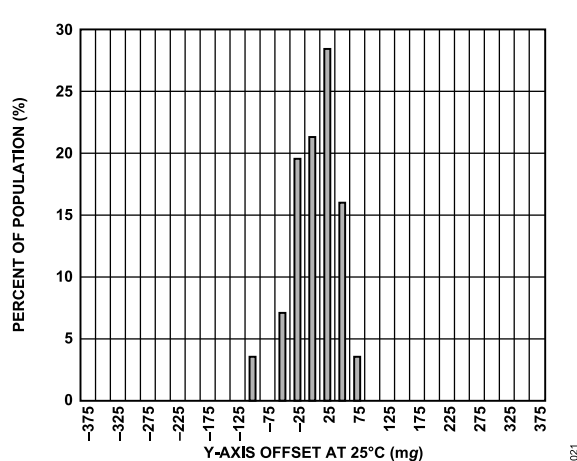


Figure 18. Zero g Offset Histogram at 25°C, Y-Axis



## TYPICAL PERFORMANCE CHARACTERISTICS

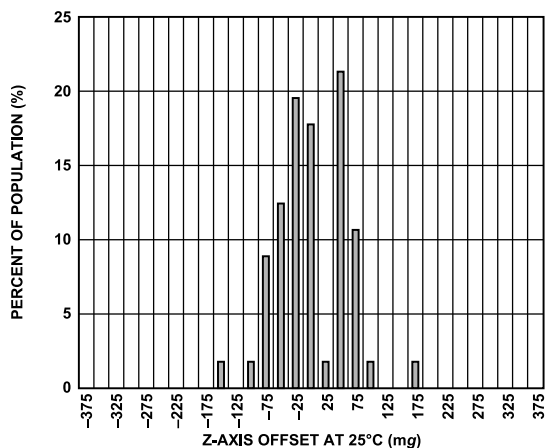


Figure 19. Zero g Offset Histogram at 25°C, Z-Axis

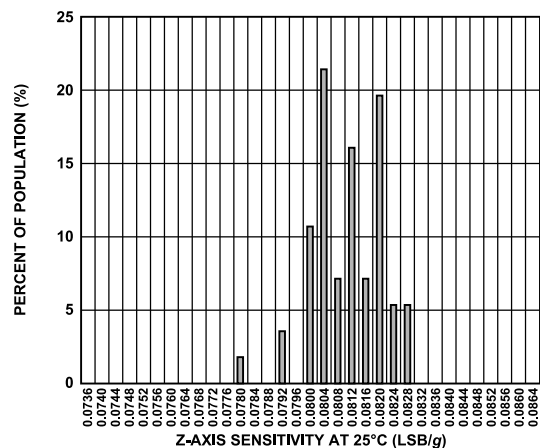


Figure 22. Sensitivity Histogram at 25°C, Z-Axis

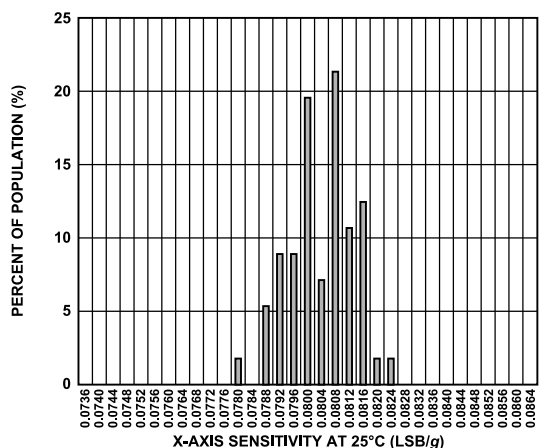


Figure 20. Sensitivity Histogram at 25°C, X-Axis

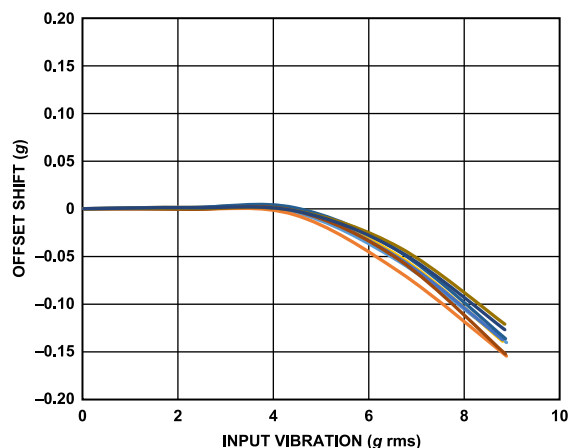
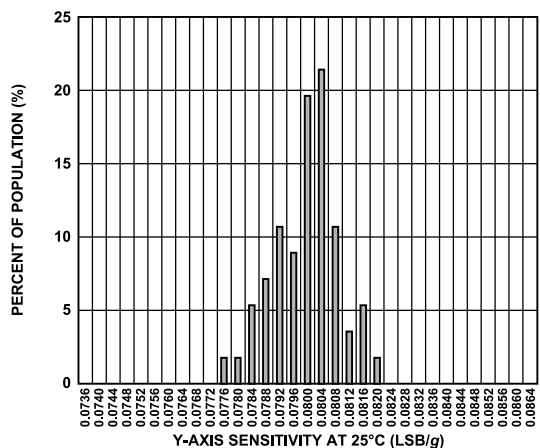
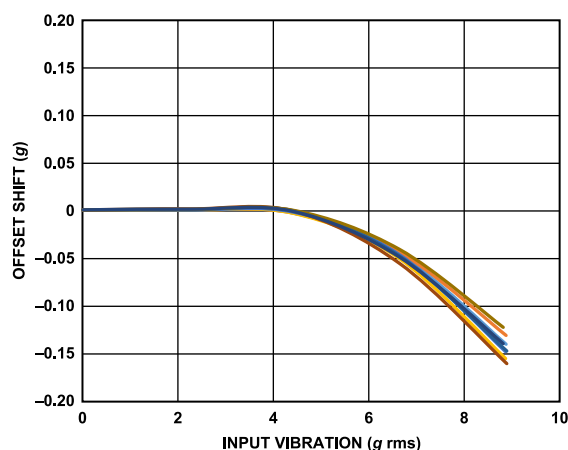
Figure 23. VRE, X-Axis Offset from +1 g,  $\pm 10$  g Range, X-Axis Orientation = +1 g

Figure 21. Sensitivity Histogram at 25°C, Y-Axis

Figure 24. VRE, Y-Axis Offset from +1 g,  $\pm 10$  g Range, Y-Axis Orientation = +1 g

## TYPICAL PERFORMANCE CHARACTERISTICS

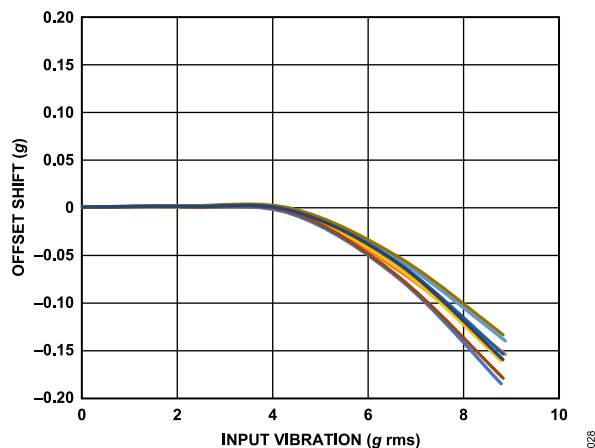


Figure 25. VRE, Z-Axis Offset from +1 g,  $\pm 10$  g Range, Z-Axis Orientation = +1 g

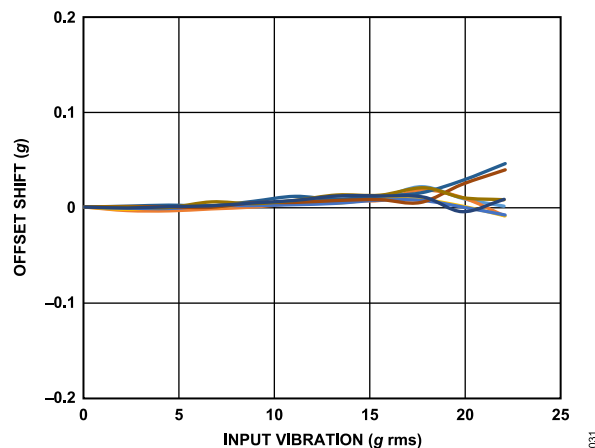


Figure 28. VRE, Z-Axis Offset from -1 g,  $\pm 10$  g Range, Z-Axis Orientation = -1 g

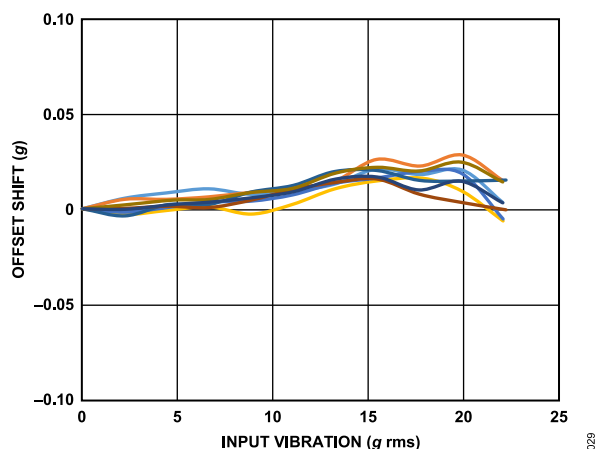


Figure 26. VRE, X-Axis Offset from -1 g,  $\pm 10$  g Range, X-Axis Orientation = -1 g

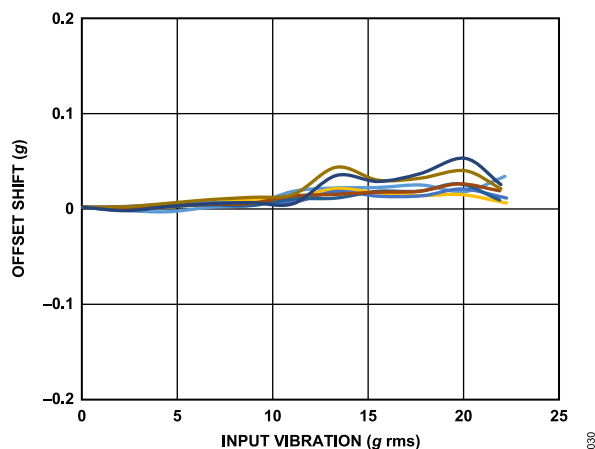


Figure 27. VRE, Y-Axis Offset from -1 g,  $\pm 10$  g Range, Y-Axis Orientation = -1 g

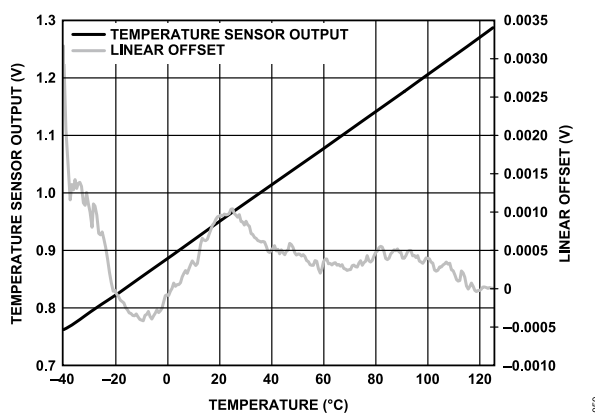


Figure 29. Temperature Sensor Output and Linear Offset vs. Temperature

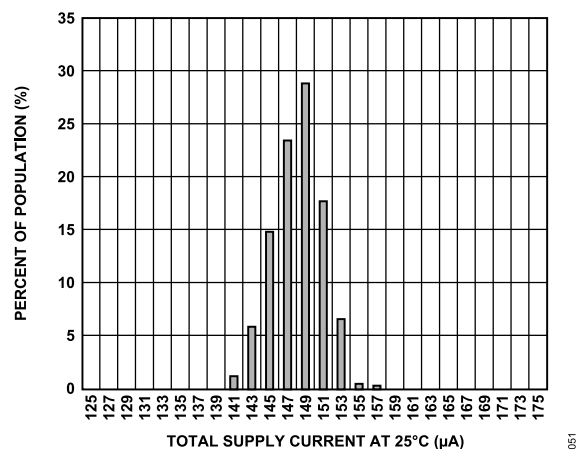


Figure 30. Total Supply Current, 3.3 V

## TYPICAL PERFORMANCE CHARACTERISTICS

## ADXL357

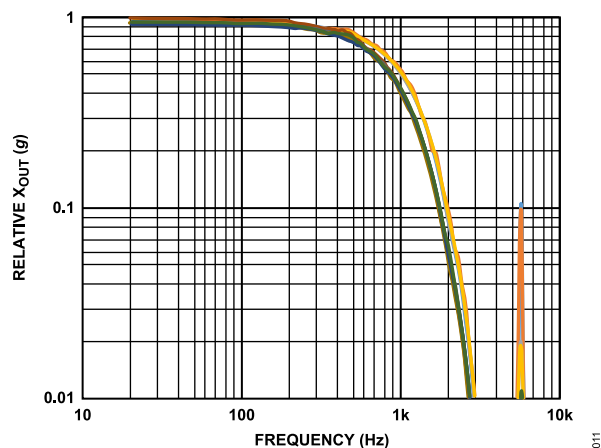


Figure 31. Frequency Response for X-Axis at 4 kHz ODR

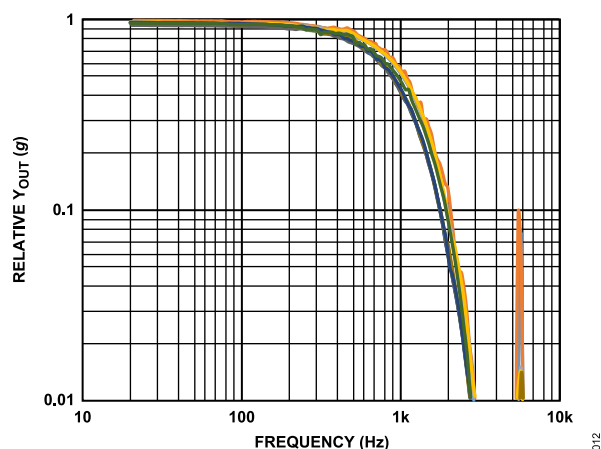


Figure 32. Frequency Response for Y-Axis at 4 kHz ODR

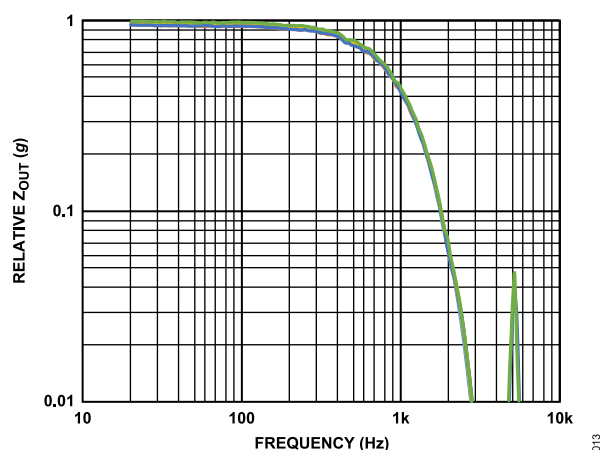


Figure 33. Frequency Response for Z-Axis at 4 kHz ODR

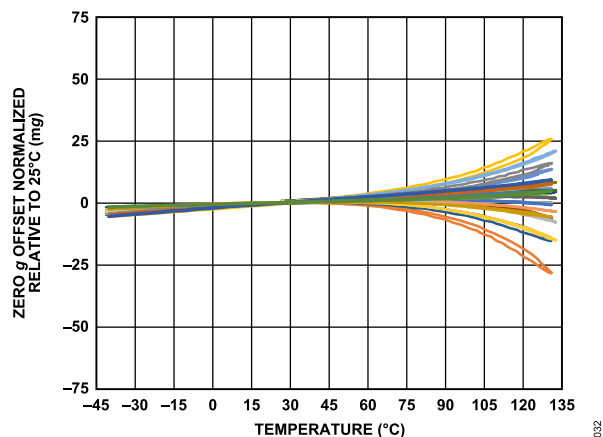


Figure 34. Zero g Offset Normalized Relative to 25°C vs. Temperature, X-Axis

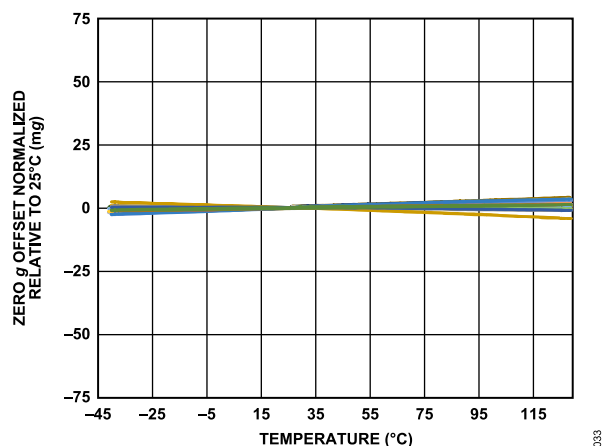


Figure 35. Zero g Offset Normalized Relative to 25°C vs. Temperature, Y-Axis

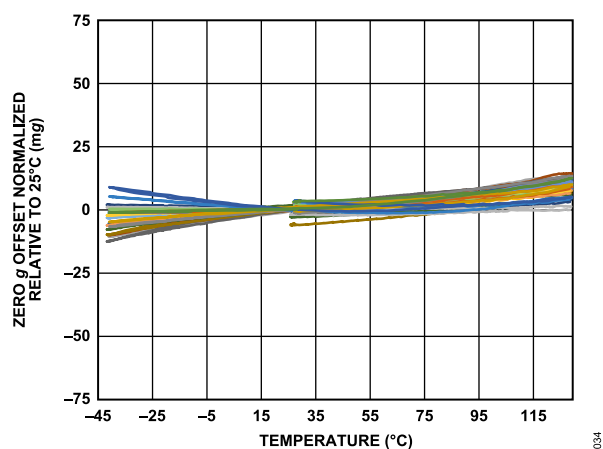


Figure 36. Zero g Offset Normalized Relative to 25°C vs. Temperature, Z-Axis

## TYPICAL PERFORMANCE CHARACTERISTICS

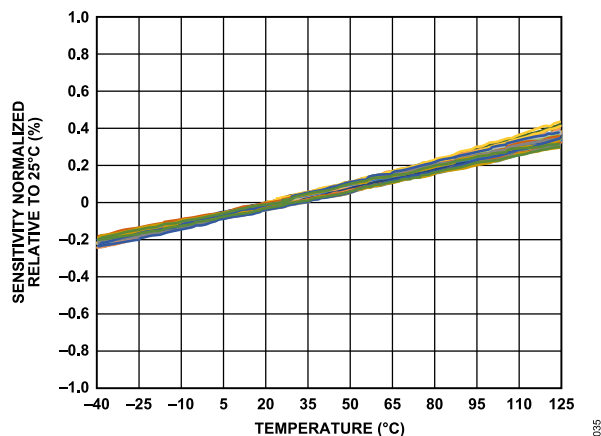


Figure 37. Sensitivity Normalized Relative to 25°C vs. Temperature X-Axis

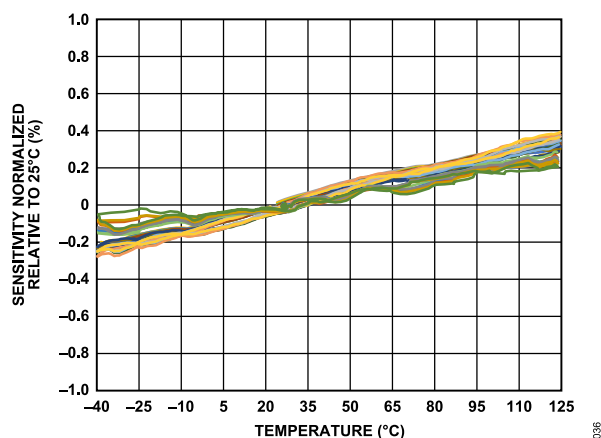


Figure 38. Sensitivity Normalized Relative to 25°C vs. Temperature Y-Axis

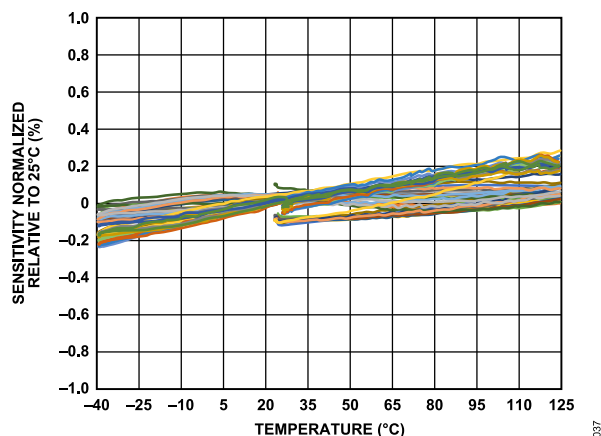


Figure 39. Sensitivity Normalized Relative to 25°C vs. Temperature Z-Axis

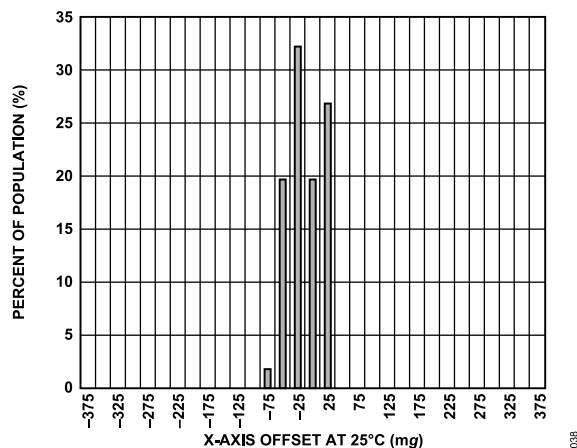


Figure 40. Zero g Offset Histogram at 25°C, X-Axis

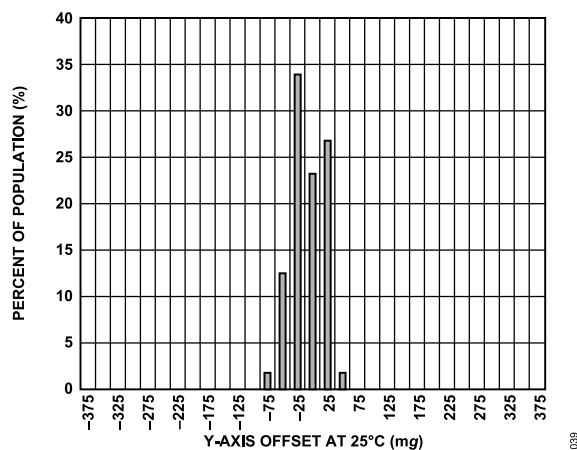


Figure 41. Zero g Offset Histogram at 25°C, Y-Axis

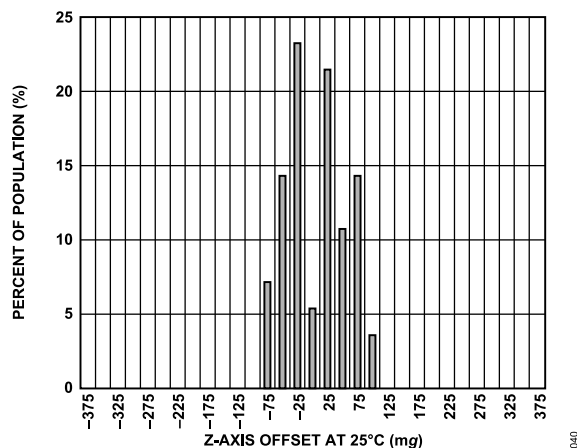


Figure 42. Zero g Offset Histogram at 25°C, Z-Axis

## TYPICAL PERFORMANCE CHARACTERISTICS

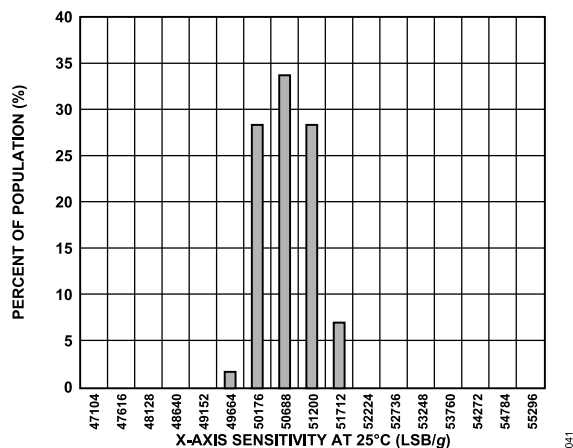


Figure 43. Sensitivity Histogram at 25°C, X-Axis

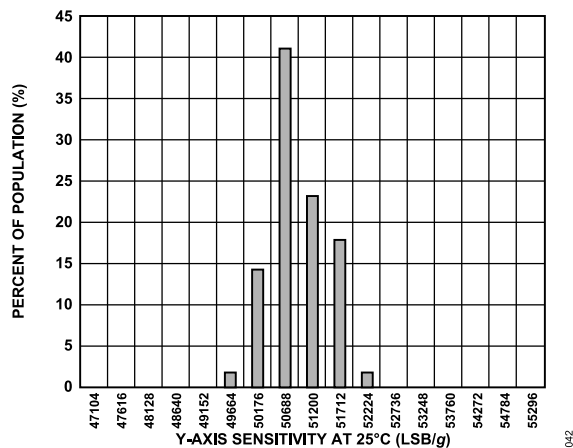


Figure 44. Sensitivity Histogram at 25°C, Y-Axis

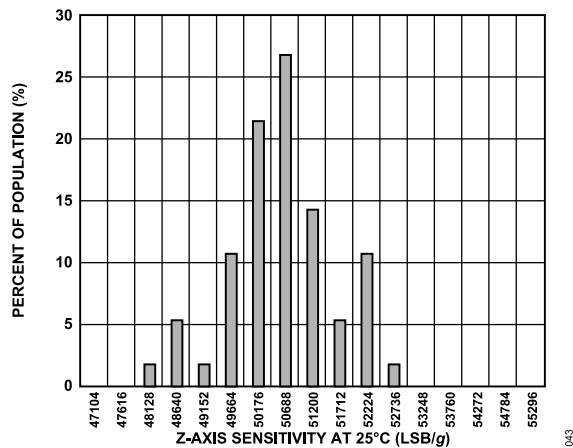
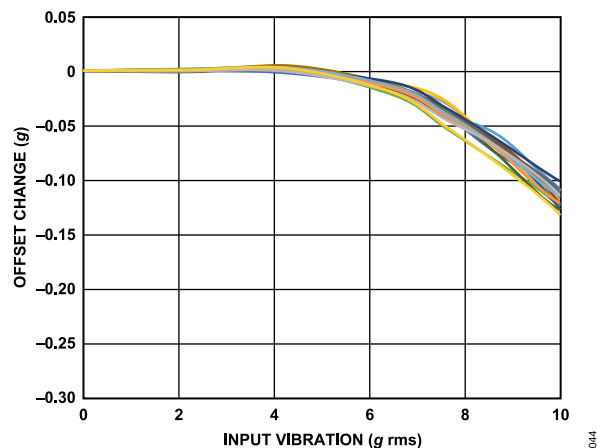
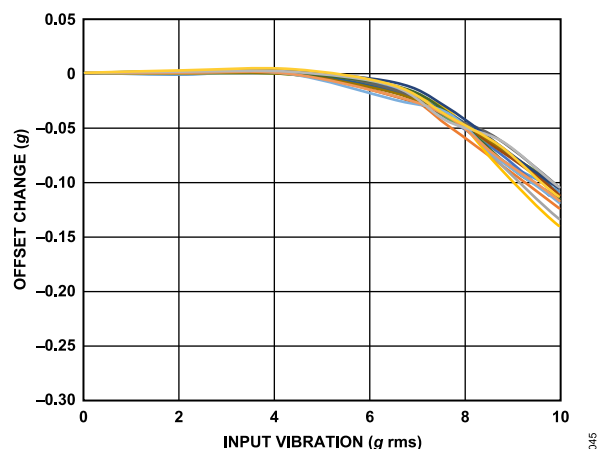
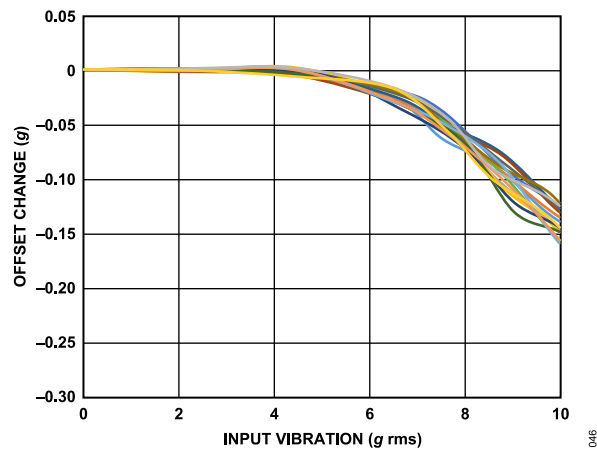


Figure 45. Sensitivity Histogram at 25°C, Z-Axis

Figure 46. VRE, X-Axis Offset from +1 g,  $\pm 10$  g Range, X-Axis Orientation = +1 gFigure 47. VRE, Y-Axis Offset from +1 g,  $\pm 10$  g Range, Y-Axis Orientation = +1 gFigure 48. VRE, Z-Axis Offset from +1 g,  $\pm 10$  g Range, Z-Axis Orientation = +1 g

## TYPICAL PERFORMANCE CHARACTERISTICS

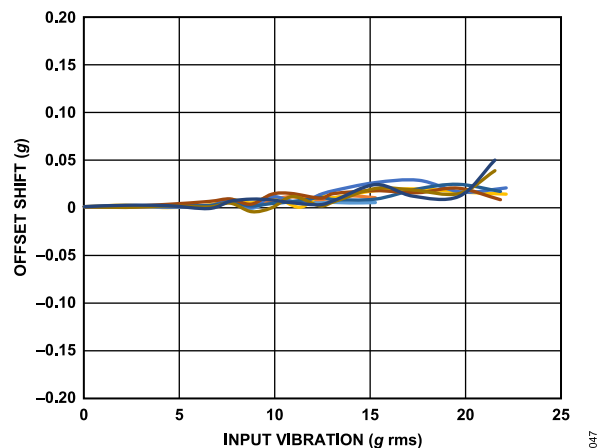


Figure 49. VRE, X-Axis Offset from  $-1\text{ g}$ ,  $\pm 40\text{ g}$  Range, X-Axis Orientation =  $-1\text{ g}$

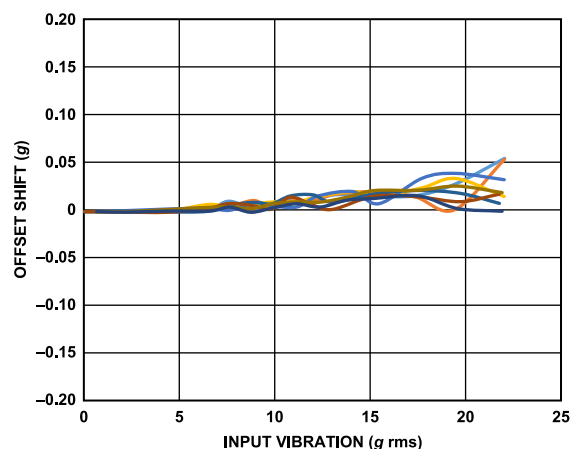


Figure 50. VRE, Y-Axis Offset from  $-1\text{ g}$ ,  $\pm 40\text{ g}$  Range, Y-Axis Orientation =  $-1\text{ g}$

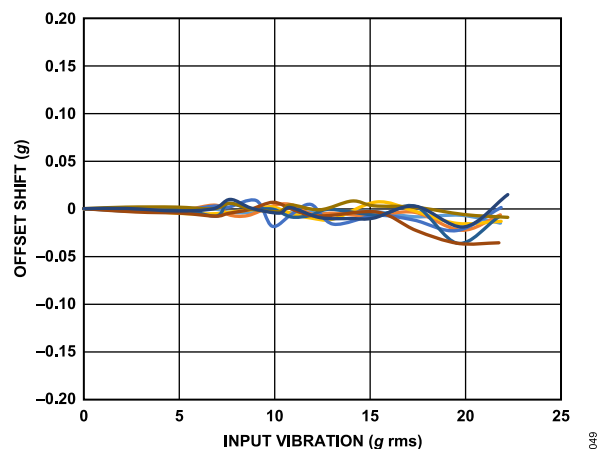


Figure 51. VRE, Z-Axis Offset from  $+1\text{ g}$ ,  $\pm 40\text{ g}$  Range, Z-Axis Orientation =  $+1\text{ g}$

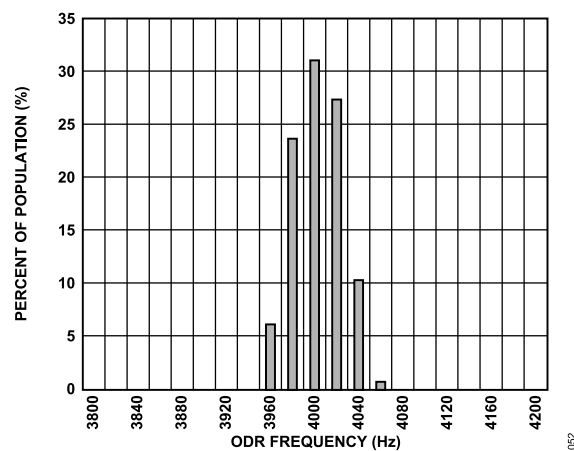


Figure 52. Output Data Rate (Internal Clock) Histogram

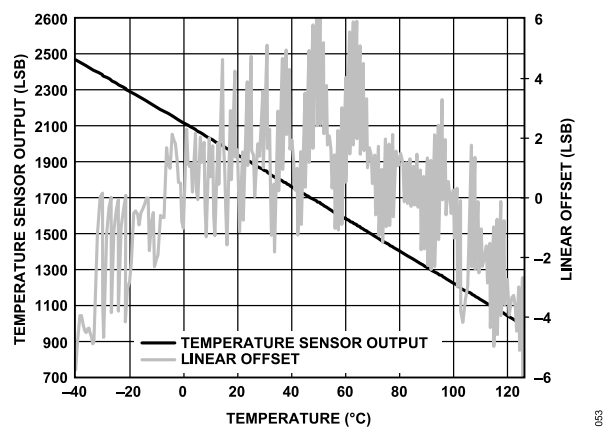


Figure 53. Temperature Sensor Output and Linear Offset vs. Temperature

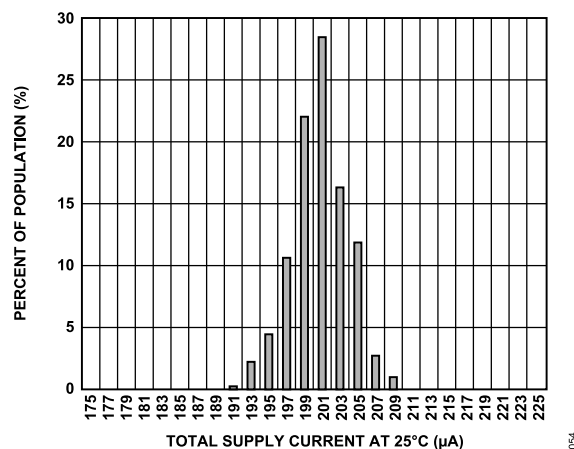


Figure 54. Total Supply Current, 3.3 V

## TYPICAL PERFORMANCE CHARACTERISTICS

## ADXL357B

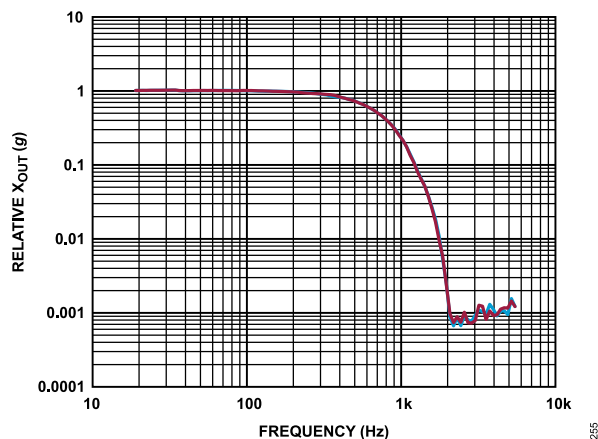


Figure 55. Frequency Response for X-Axis at 4 kHz ODR

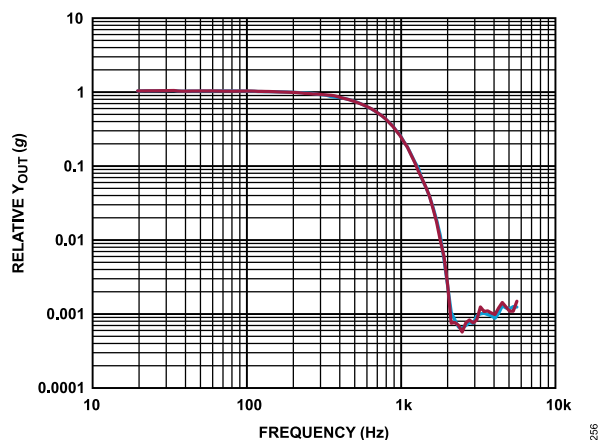


Figure 56. Frequency Response for Y-Axis at 4 kHz ODR

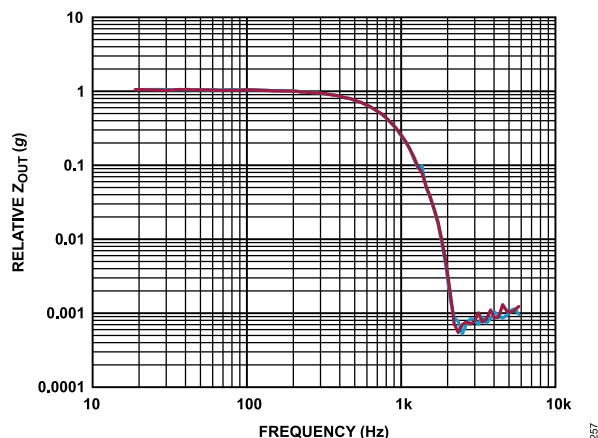


Figure 57. Frequency Response for Z-Axis at 4 kHz ODR

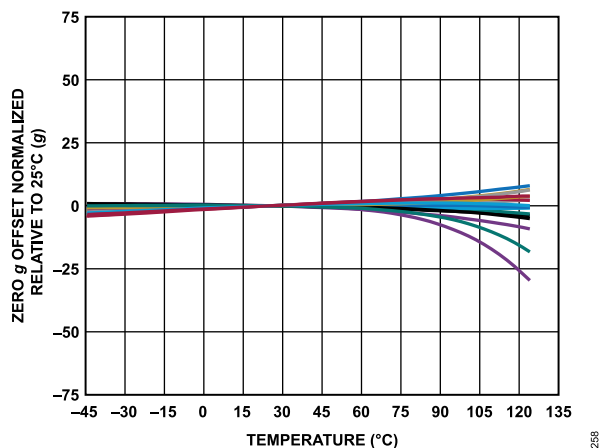


Figure 58. Zero g Offset Normalized Relative to 25°C vs. Temperature, X-Axis

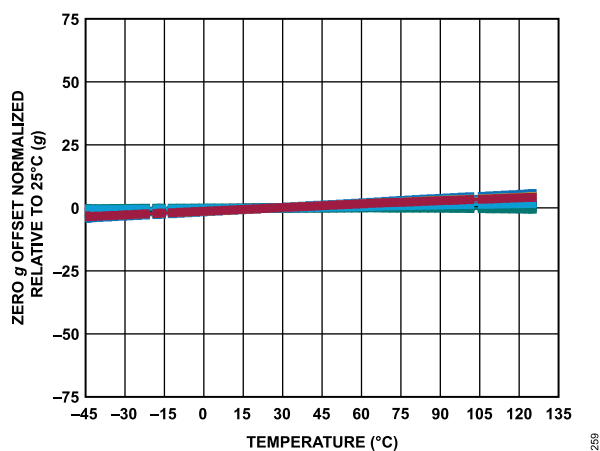


Figure 59. Zero g Offset Normalized Relative to 25°C vs. Temperature, Y-Axis

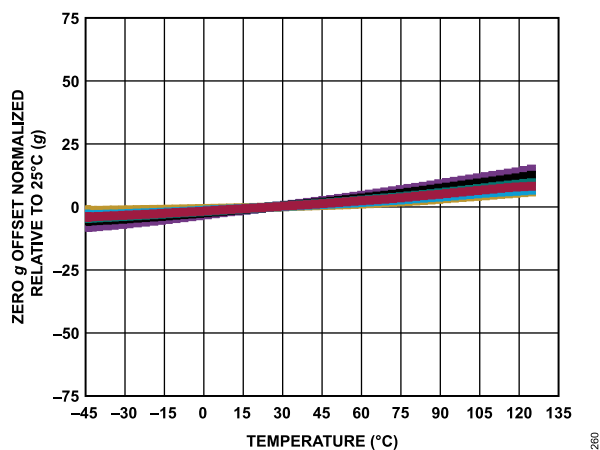


Figure 60. Zero g Offset Normalized Relative to 25°C vs. Temperature, Z-Axis

## TYPICAL PERFORMANCE CHARACTERISTICS

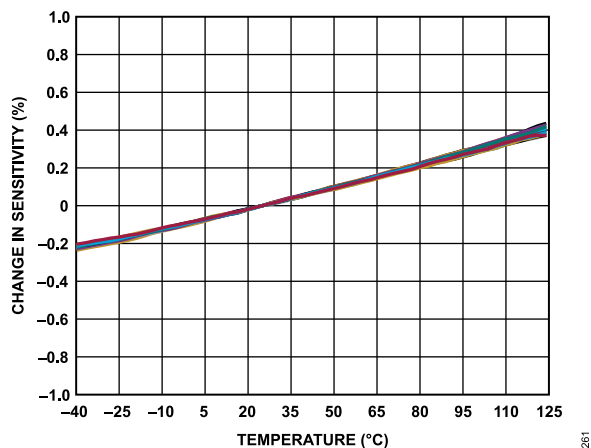


Figure 61. Sensitivity Normalized Relative to 25°C vs. Temperature X-Axis

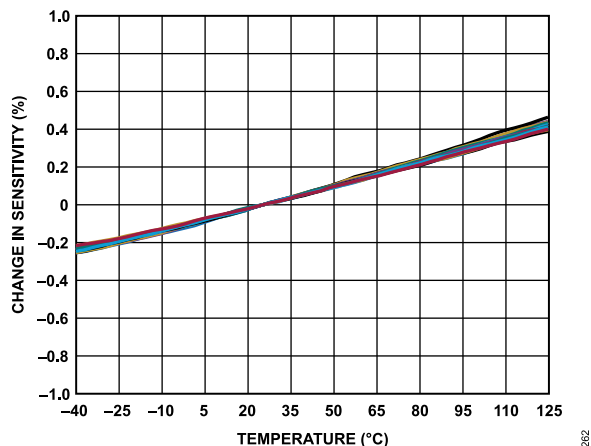


Figure 62. Sensitivity Normalized Relative to 25°C vs. Temperature Y-Axis

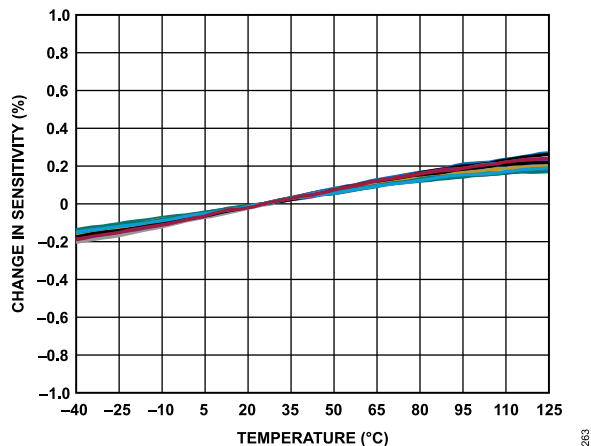


Figure 63. Sensitivity Normalized Relative to 25°C vs. Temperature Z-Axis

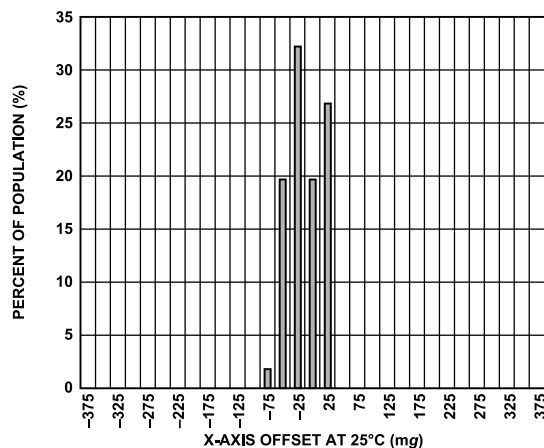


Figure 64. Zero g Offset Histogram at 25°C, X-Axis

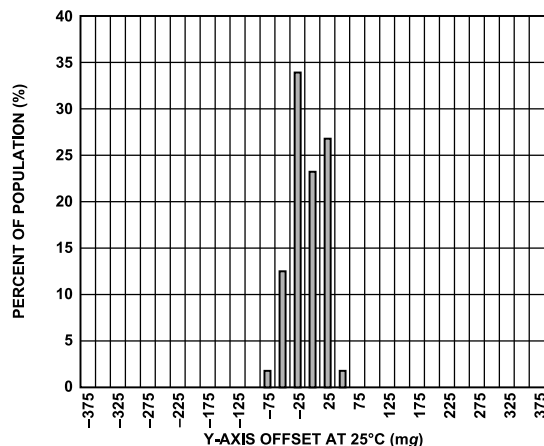


Figure 65. Zero g Offset Histogram at 25°C, Y-Axis

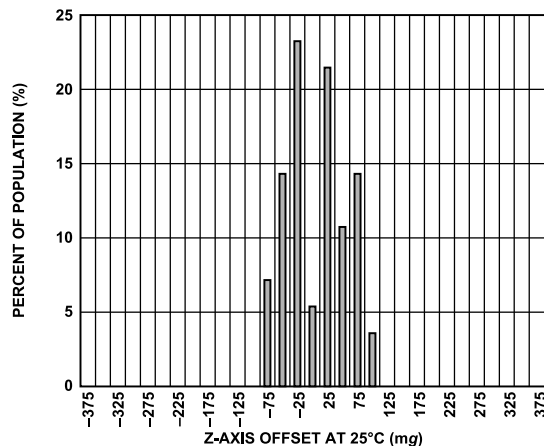


Figure 66. Zero g Offset Histogram at 25°C, Z-Axis



## TYPICAL PERFORMANCE CHARACTERISTICS

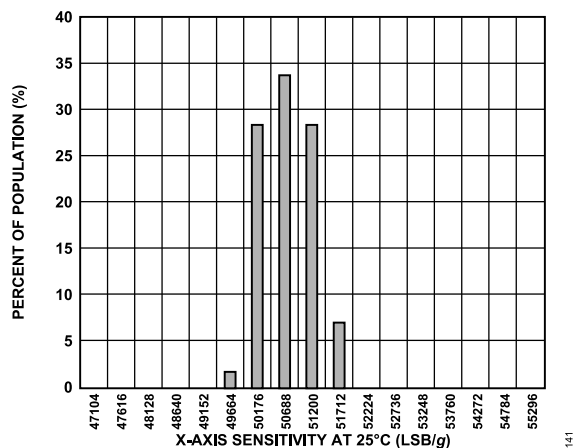


Figure 67. Sensitivity Histogram at 25°C, X-Axis

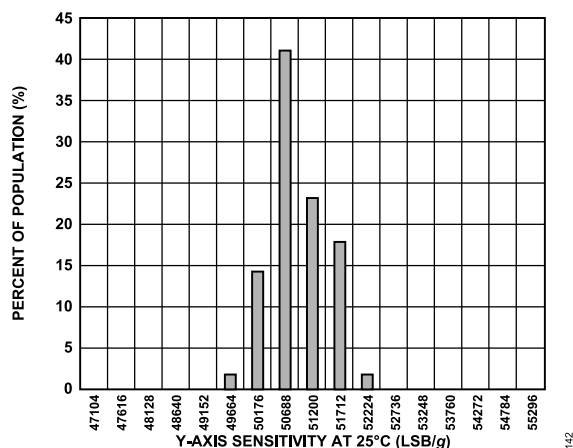


Figure 68. Sensitivity Histogram at 25°C, Y-Axis

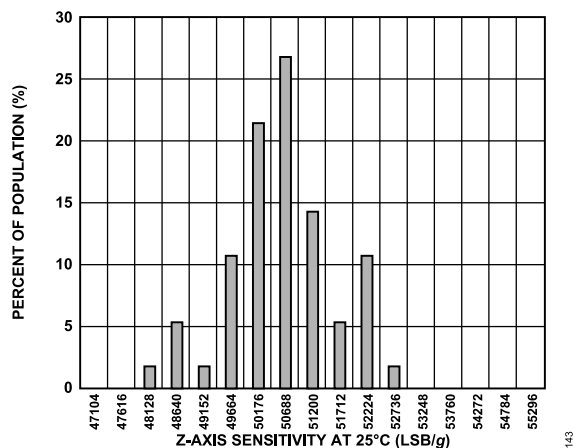
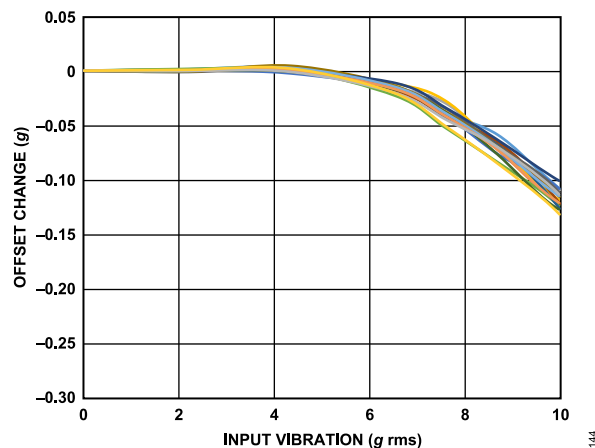
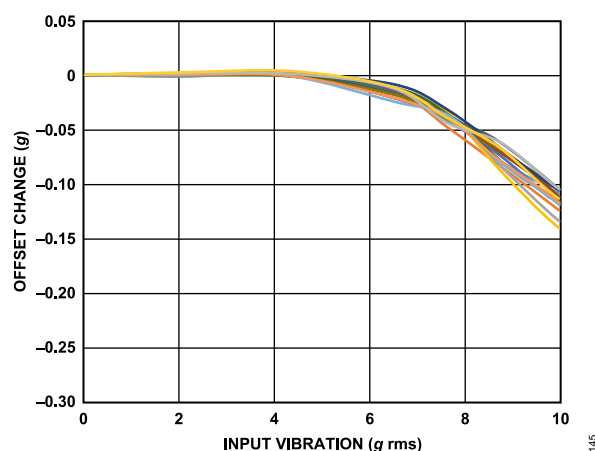
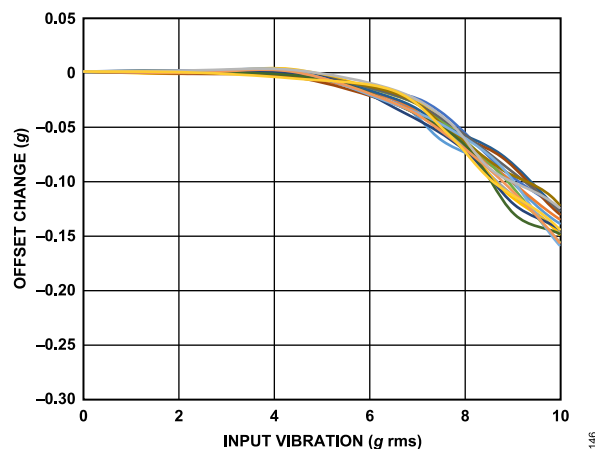


Figure 69. Sensitivity Histogram at 25°C, Z-Axis

Figure 70. VRE, X-Axis Offset from +1 g,  $\pm 10$  g Range, X-Axis Orientation = +1 gFigure 71. VRE, Y-Axis Offset from +1 g,  $\pm 10$  g Range, Y-Axis Orientation = +1 gFigure 72. VRE, Z-Axis Offset from +1 g,  $\pm 10$  g Range, Z-Axis Orientation = +1 g

## TYPICAL PERFORMANCE CHARACTERISTICS

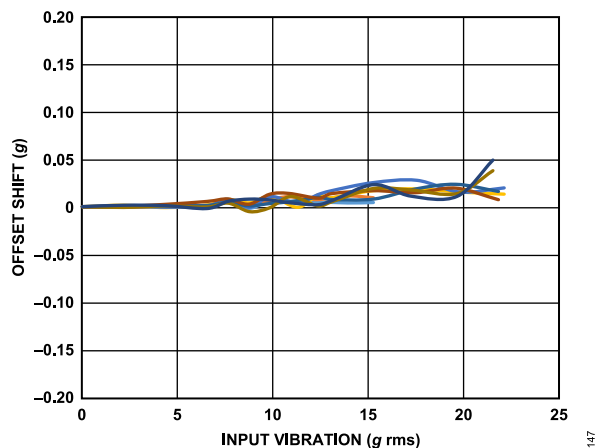


Figure 73. VRE, X-Axis Offset from  $-1\text{ g}$ ,  $\pm 40\text{ g}$  Range, X-Axis Orientation =  $-1\text{ g}$

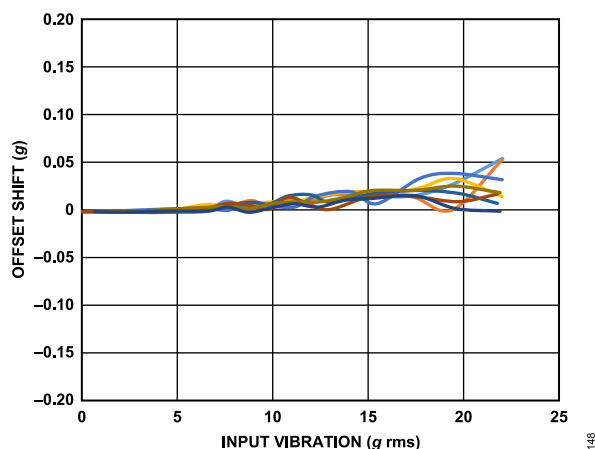


Figure 74. VRE, Y-Axis Offset from  $-1\text{ g}$ ,  $\pm 40\text{ g}$  Range, X-Axis Orientation =  $-1\text{ g}$

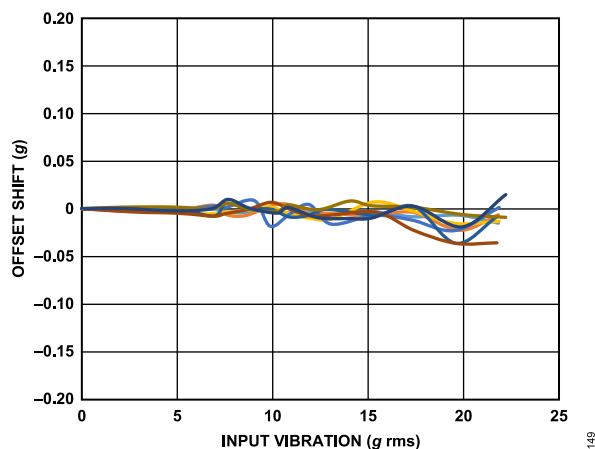


Figure 75. VRE, Z-Axis Offset from  $-1\text{ g}$ ,  $\pm 40\text{ g}$  Range, Z-Axis Orientation =  $-1\text{ g}$

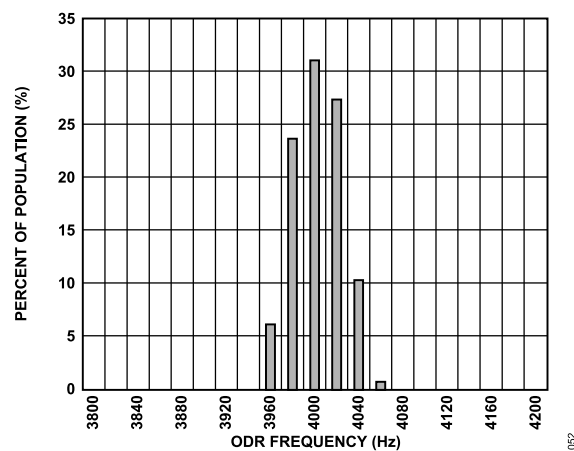


Figure 76. Output Data Rate (Internal Clock) Histogram

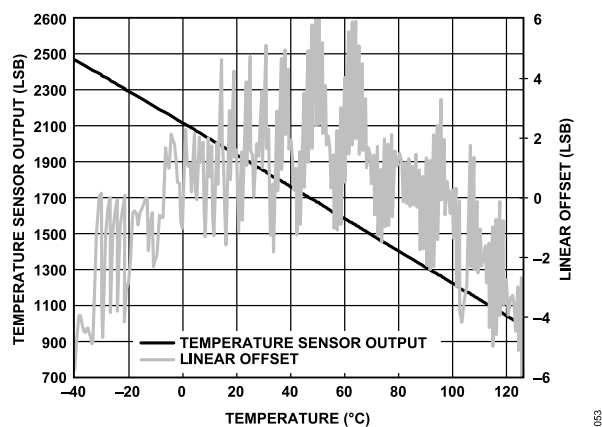


Figure 77. Temperature Sensor Output and Linear Offset vs. Temperature

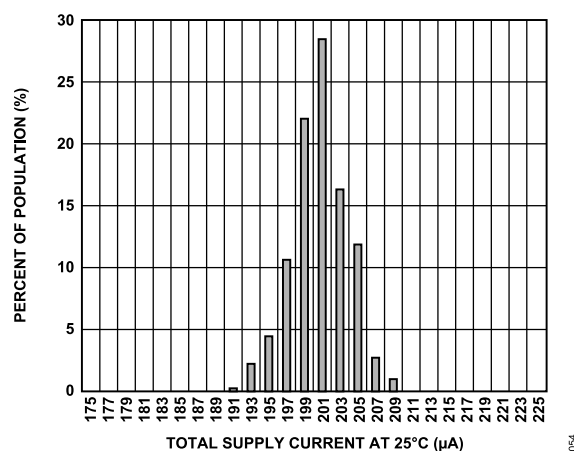


Figure 78. Total Supply Current, 3.3 V

## TYPICAL PERFORMANCE CHARACTERISTICS

## ROOT ALLAN VARIANCE (RAV) ADXL357/ADXL357B CHARACTERISTICS

Figure 79 to Figure 81 include data for multiple devices and multiple lots, and they were taken in the  $\pm 10\text{ g}$  range, unless otherwise noted.

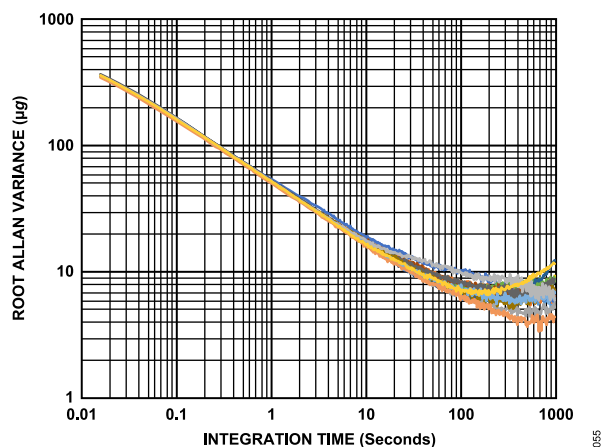


Figure 79. RAV, X-Axis

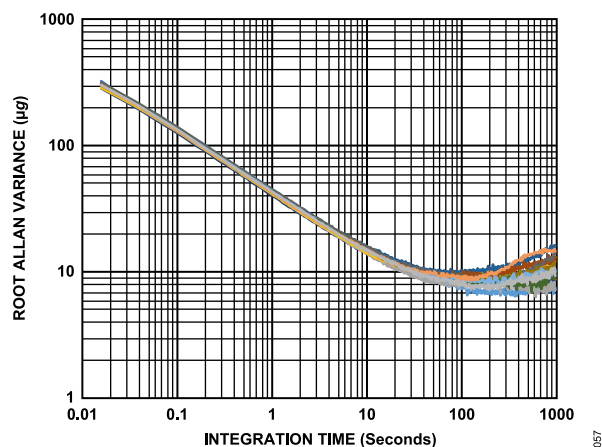


Figure 81. RAV, Z-Axis

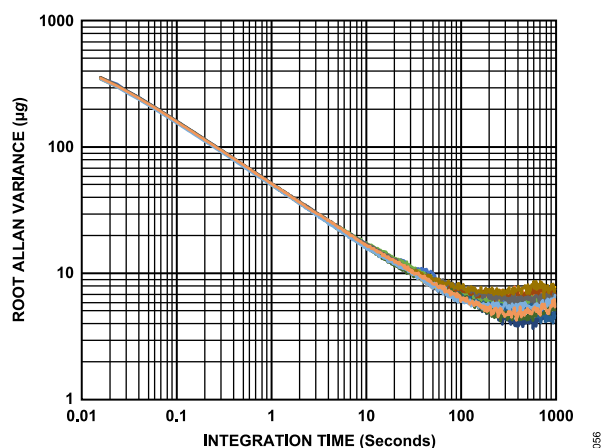


Figure 80. RAV, Y-Axis

## THEORY OF OPERATION

The ADXL356 is a complete 3-axis, ultralow noise and ultrastable offset microelectromechanical systems (MEMS) accelerometer with outputs ratiometric to the analog 1.8 V supply,  $V_{1P8ANA}$ . The ADXL357 and ADXL357B add three high resolution ADCs that use the analog 1.8 V supply as a reference to provide digital outputs insensitive to the supply voltage. The ADXL356B is pin selectable for  $\pm 10\text{ g}$  or  $\pm 20\text{ g}$  full scale, the ADXL356C is pin selectable for  $\pm 10\text{ g}$  or  $\pm 40\text{ g}$  full scale, and the ADXL357 and ADXL357B are programmable for  $\pm 10\text{ g}$ ,  $\pm 20\text{ g}$ , or  $\pm 40\text{ g}$  full scale and offer both SPI and I<sup>2</sup>C communications ports.

The micromachined, sensing elements are fully differential, comprising the lateral x-axis and y-axis sensors and the vertical, teeter totter z-axis sensors. The x-axis and y-axis sensors and the z-axis sensors go through separate signal paths that minimize offset

drift and noise. The signal path is fully differential, except for a differential to single-ended conversion at the analog outputs of the ADXL356.

The analog accelerometer outputs of the ADXL356 are ratiometric to  $V_{1P8ANA}$ . Therefore, digitize them carefully. The temperature sensor output is not ratiometric. The  $X_{OUT}$ ,  $Y_{OUT}$ , and  $Z_{OUT}$  analog outputs are filtered internally with an antialiasing filter. These analog outputs also have an internal 32 k $\Omega$  series resistor that can be used with an external capacitor to set the bandwidth of the output.

The ADXL357 includes antialias filters before and after the high resolution  $\Sigma$ - $\Delta$  ADC. User-selectable output data rates and filter corners are provided. The temperature sensor is digitized with a 12-bit successive approximation register (SAR) ADC.

## APPLICATIONS INFORMATION

## ANALOG OUTPUT

Figure 82 shows the ADXL356 application circuit. The analog outputs ( $X_{OUT}$ ,  $Y_{OUT}$ , and  $Z_{OUT}$ ) are ratiometric to the 1.8 V analog voltage from the  $V_{1P8ANA}$  pin.  $V_{1P8ANA}$  can be powered with an on-chip LDO regulator that is powered from  $V_{SUPPLY}$ .  $V_{1P8ANA}$  can also be supplied externally by forcing  $V_{SUPPLY}$  to  $V_{SS}$ , which disables the LDO regulator. Due to the ratiometric response, the analog output requires referencing to the  $V_{1P8ANA}$  supply when digitizing to achieve the inherent noise and offset performance of the ADXL356. The 0 g bias output is nominally equal to  $V_{1P8ANA}/2$ . The recommended option is to use the ADXL356 with a ratiometric ADC (for example, the Analog Devices, Inc., [AD7682](#)) and  $V_{1P8ANA}$  providing the voltage reference. This configuration results in self cancellation of errors due to minor supply variations.

The ADXL356 outputs two forms of filtering: internal anti-aliasing filtering with a cutoff frequency of approximately 1.5 kHz, and external filtering. The external filter uses a fixed, on-chip, 32 k $\Omega$  resistance in series with each output in conjunction with the external capacitors to implement the low-pass filter antialiasing and noise reduction prior to the external ADC. The antialias filter

cutoff frequency must be significantly higher than the desired signal bandwidth. If the antialias filter corner is too low, ratiometricity can degrade where the signal attenuation is different from the reference attenuation.

## DIGITAL OUTPUT

Figure 83 shows the ADXL357 and ADXL357B application circuit with the recommended bypass capacitors. The communications interface is either SPI or I<sup>2</sup>C (see the [Serial Communications](#) section for additional information).

The ADXL357/ADXL357B include an internal configurable digital band-pass filter. Both the high-pass and low-pass poles of the filter are adjustable, as detailed in the [Filter Settings Register](#) section and [Table 45](#). At power-up, the default conditions for the filters are as follows:

- High-pass filter (HPF) = dc (off)
- Low-pass filter (LPF) = 1000 Hz
- Output data rate = 4000 Hz

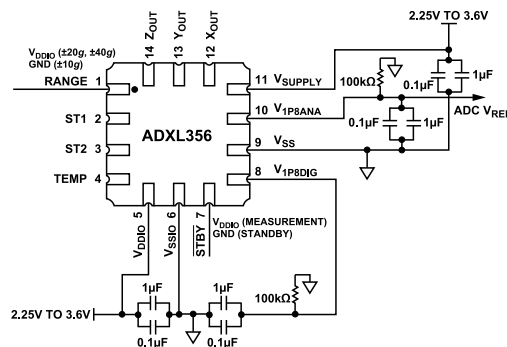


Figure 82. ADXL356 Application Circuit

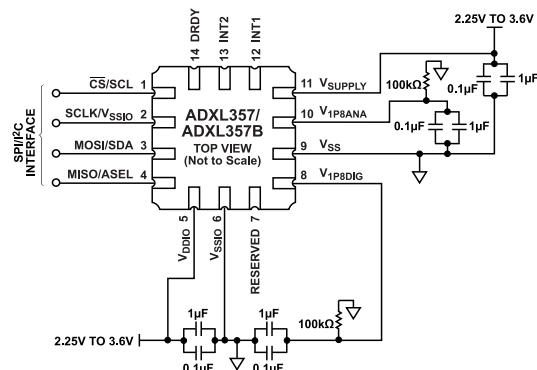


Figure 83. ADXL357/ADXL357B Application Circuit

## APPLICATIONS INFORMATION

## AXES OF ACCELERATION SENSITIVITY

Figure 84 shows the axes of acceleration sensitivity. Note that the output voltage increases when accelerated along the sensitive axis.

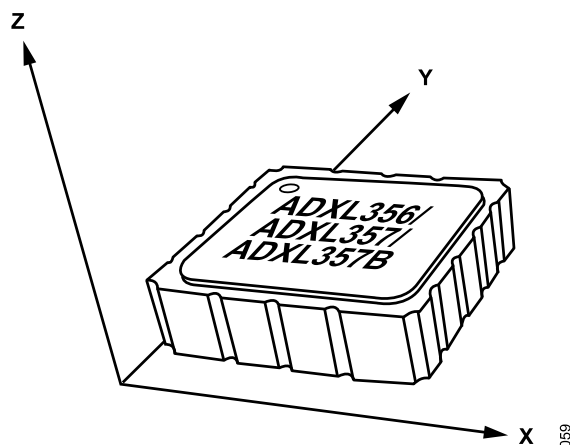


Figure 84. Axes of Acceleration Sensitivity

## POWER SEQUENCING

There are two methods for applying power to the device. Typically, internal LDO regulators generate the 1.8 V power for the analog and digital supplies,  $V_{1P8ANA}$  and  $V_{1P8DIG}$ , respectively. Optionally, the internal LDO regulators can be disabled and  $V_{1P8ANA}$  and  $V_{1P8DIG}$  are driven by external 1.8 V supplies.

When using the internal LDO regulators, connect  $V_{SUPPLY}$  to a voltage source between 2.25 V and 3.6 V. In this case, the recommended power sequence is to apply power to  $V_{DDIO}$ , followed by applying power to  $V_{SUPPLY}$  approximately 10  $\mu$ s later. If necessary,  $V_{SUPPLY}$  and  $V_{DDIO}$  can be powered from the same voltage source, so that both are powered at the same time. However,  $V_{SUPPLY}$  cannot be powered before  $V_{DDIO}$ .

To disable the internal LDO regulators, tie  $V_{SUPPLY}$  to ground and use external 1.8 V supplies to power  $V_{1P8ANA}$  and  $V_{1P8DIG}$ .  $V_{1P8ANA}$  and  $V_{1P8DIG}$  must have the same voltage level. The maximum acceptable tolerance between the external  $V_{1P8ANA}$  and  $V_{1P8DIG}$  voltage levels is 50 mV. In the case of bypassing the LDO regulators, the recommended power sequence is to apply power to  $V_{DDIO}$ , followed by applying power to  $V_{1P8DIG}$  approximately 10  $\mu$ s later, and then applying power to  $V_{1P8ANA}$  approximately 10  $\mu$ s later. If necessary,  $V_{1P8DIG}$  and  $V_{DDIO}$  can be powered from the same external 1.8 V supply, which can also be tied to  $V_{1P8ANA}$  with proper isolation, so that all are powered at the same time. In this case, proper decoupling and low frequency isolation are important to maintain the noise performance of the sensor.

When power cycling the ADXL356/ADXL357/ADXL357B, it is highly recommended to fully discharge the device to ground level ( $V_{SUPPLY}$  and  $V_{DDIO} = 0$  V), for at least 200 ms, on each power cycle. It is also highly recommended to help the LDO discharge properly by adding external resistors of approximately 100 k $\Omega$  between the LDO

outputs ( $V_{1P8ANA}$  and  $V_{1P8DIG}$ ) and ground (0 V), especially when power cycling at temperatures below  $-20^{\circ}\text{C}$ . See the recommended application circuits shown in Figure 82 and Figure 83 for reference.

## POWER SUPPLY DESCRIPTION

The ADXL356/ADXL357/ADXL357B have four different power supply domains:  $V_{SUPPLY}$ ,  $V_{1P8ANA}$ ,  $V_{1P8DIG}$ , and  $V_{DDIO}$ . The internal analog and digital circuitry operates at 1.8 V nominal.

 $V_{SUPPLY}$ 

$V_{SUPPLY}$  is 2.25 V to 3.6 V, which is the input range to the two LDO regulators that generate the nominal 1.8 V outputs for  $V_{1P8ANA}$  and  $V_{1P8DIG}$ . Connect  $V_{SUPPLY}$  to  $V_{SS}$  to disable the LDO regulators, which allows driving  $V_{1P8ANA}$  and  $V_{1P8DIG}$  from an external source.

 $V_{1P8ANA}$ 

All sensor and analog signal processing circuitry operates in this domain. Offset and sensitivity of the analog output ADXL356 are ratiometric to this supply voltage. When using external ADCs, use  $V_{1P8ANA}$  as the reference voltage. The ADXL357/ADXL357B include ADCs that are ratiometric to  $V_{1P8ANA}$ , thereby rendering the offset and sensitivity of the ADXL357/ADXL357B digital output is insensitive to the value of  $V_{1P8ANA}$ .  $V_{1P8ANA}$  can be an input or an output as defined by the state of the  $V_{SUPPLY}$  voltage.

 $V_{1P8DIG}$ 

$V_{1P8DIG}$  is the supply voltage for the internal logic circuitry. A separate LDO regulator decouples the digital supply noise from the analog signal path.  $V_{1P8ANA}$  can be an input or an output as defined by the state of the  $V_{SUPPLY}$  voltage. If driven externally,  $V_{1P8DIG}$  must be the same voltage as the  $V_{1P8ANA}$  voltage.

 $V_{DDIO}$ 

The  $V_{DDIO}$  value determines the logic high levels. On the analog output ADXL356,  $V_{DDIO}$  sets the logic high level for the self test pins, ST1 and ST2, as well as the STBY pin. On the digital output ADXL357/ADXL357B,  $V_{DDIO}$  sets the logic high level for communications interface ports, as well as the interrupt and DRDY outputs.

The LDO regulators are operational when  $V_{SUPPLY}$  is between 2.25 V and 3.6 V.  $V_{1P8ANA}$  and  $V_{1P8DIG}$  are the regulator outputs in this mode. Alternatively, when tying  $V_{SUPPLY}$  to  $V_{SS}$ ,  $V_{1P8ANA}$  and  $V_{1P8DIG}$  are supply voltage inputs with a 1.62 V to 1.98 V range.

## OVERRANGE PROTECTION

To avoid electrostatic capture of the proof mass when the accelerometer is subject to input acceleration beyond its full-scale range, all sensor drive clocks turn off for 0.5 ms. In the  $\pm 10$  g range setting, the overrange protection activates for input signals beyond approximately  $\pm 40$  g ( $\pm 25\%$ ), and for the  $\pm 20$  g and  $\pm 40$  g range settings, the threshold corresponds to about  $\pm 80$  g ( $\pm 25\%$ ).

## APPLICATIONS INFORMATION

When overrange protection occurs, the  $X_{OUT}$ ,  $Y_{OUT}$ , and  $Z_{OUT}$  pins on the ADXL356 begin to drive to midscale, whereas the ADXL357/ADXL357B float toward zero LSB, and the first in, first out (FIFO) buffer begins filling with this data. Figure 85 shows the ADXL356 overrange behavior for a 150 g, 5 ms width, half sine-wave shock event. The ADXL356 range is set to  $\pm 20$  g.

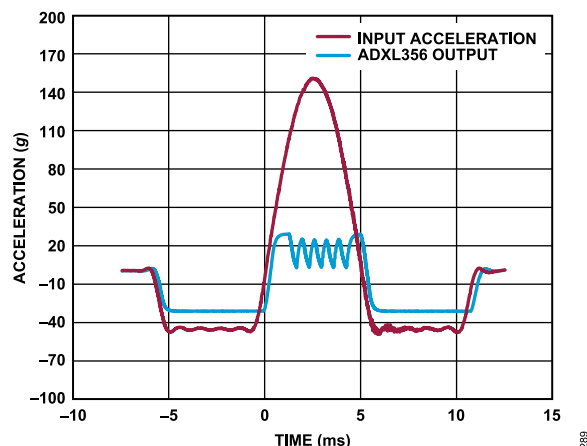


Figure 85. ADXL356 Overrange Behavior for  $\pm 20$  g Range Setting

## MECHANICAL HEADROOM VS. FREQUENCY

The behavior of the ADXL356, ADXL357, and ADXL357B MEMS sensor over frequency can be modeled as an undamped linear second order system.

The ADXL357B MEMS sensor is damped much heavily compared to the ADXL356 and ADXL357, which makes it more robust on applications where the accelerometer is exposed to harsh vibration environments. Refer to the vibration operation limits in the [Absolute Maximum Ratings](#) section for more details.

For the ADXL356 and ADXL357, avoiding vibrations near the accelerometer resonant frequency is particularly important because the input acceleration is amplified approximately 18 times for x-axis and y-axis, and approximately 5 times for z-axis. In essence, this reduces the minimum acceleration input necessary. Therefore, the gap between the MEMS mobile and fixed structures is closed (mechanical headroom). Continuous exposure to high amplitude vibrations in the vicinity of the sensor resonant frequency can result in durability issues.

Figure 86 and Figure 87 show a comparison of the mechanical headroom for the ADXL356, ADXL357, and ADXL357B over frequency.

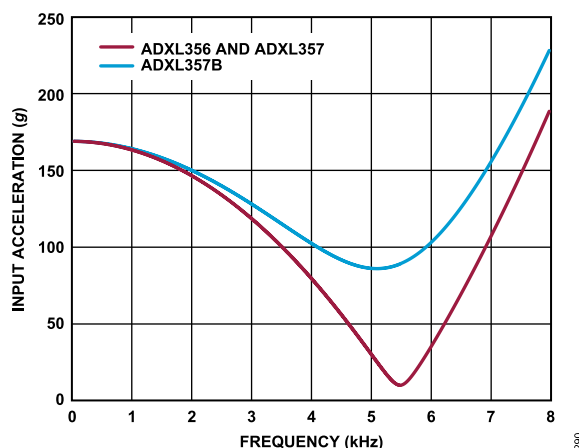


Figure 86. X-Axis and Y-Axis Mechanical Headroom vs. Frequency Comparison for ADXL356/ADXL357/ADXL357B

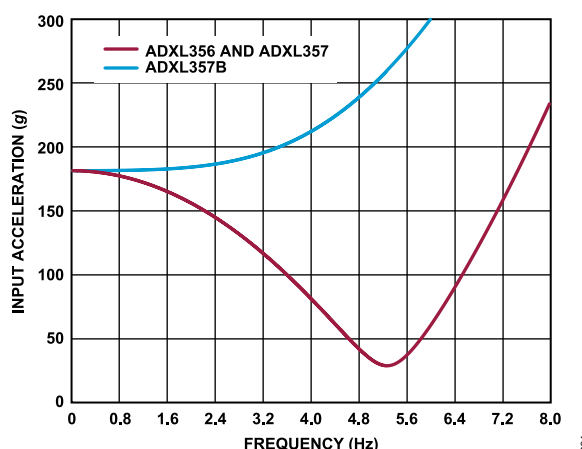


Figure 87. Z-Axis Mechanical Headroom vs. Frequency Comparison for ADXL356/ADXL357/ADXL357B

## SELF TEST

The ADXL356 and ADXL357/ADXL357B incorporate a self test feature that effectively tests the mechanical and electronic system. Enabling self test stimulates the sensor electrostatically to produce an output corresponding to the test signal applied as well as the mechanical force exerted. Only the z-axis response is specified to validate device functionality.

In the ADXL356, drive the ST1 pin to  $V_{DDIO}$  to invoke self test mode. Then, by driving the ST2 pin to  $V_{DDIO}$ , the ADXL356 applies an electrostatic force to the mechanical sensor and induces a change in output in response to the force. The self test delta (or response) is the difference in output voltage in the z-axis when ST2 is high vs. ST2 is low, while ST1 is asserted. After the self test measurement is complete, bring both pins low to resume normal operation.

The self test operation is similar in the ADXL357/ADXL357B, except ST1 and ST2 can be accessed through the SELF\_TEST register (Register 0x2E).



## APPLICATIONS INFORMATION

The self test feature rejects externally applied acceleration and only responds to the self test force, which allows an accurate measurement of the self test, even in the presence of external mechanical noise. When the self test feature is not used, both ST1 and ST2 must be kept low.

### FILTER

The ADXL356/ADXL357/ADXL357B use an analog, low-pass, anti-aliasing filter to reduce out of band noise and to limit bandwidth. The ADXL357/ADXL357B provide further digital filtering options to maintain optimal noise performance at various ODRs.

The analog, low-pass antialiasing filter in the ADXL356/ADXL357/ADXL357B provides a fixed 3 dB bandwidth of approximately 1.5 kHz, the frequency at which the voltage output response is attenuated by approximately 30%. The shape of the filter response in the frequency domain is that of a sinc filter. While the analog antialiasing filter attenuates the output response around and above its cutoff frequency, the MEMS sensor has a resonance at 5.5 kHz and mechanically amplifies the output response at around 2 kHz and above. These competing trends are apparent in the overall transfer function of the ADXL356, as shown in [Figure 8](#) to [Figure 10](#). Therefore, the overall -3 dB bandwidth of the ADXL356 is 2.4 kHz, and the overall bandwidth with  $\pm 4$  dB flatness is about 4.4 kHz.

The ADXL356 x-axis, y-axis, and z-axis analog outputs include an amplifier followed by a series 32 k $\Omega$  resistor, and output to the X<sub>OUT</sub>, the Y<sub>OUT</sub>, and the Z<sub>OUT</sub> pins, respectively.

The ADXL357/ADXL357B provide an internal 20-bit,  $\Sigma$ - $\Delta$  ADC to digitize the filtered analog signal. Additional digital filtering (beyond the analog, low-pass, antialiasing filter) consists of a low-pass digital decimation filter and a bypassable high-pass filter that supports output data rates between 4 kHz and 3.906 Hz. The decimation filter consists of two stages. The first stage is fixed decimation with a 4 kHz ODR and a low-pass filter cutoff (3 dB) at about 1 kHz. A variable second stage decimation filter is used for the 2 kHz output data rate and below (it is bypassed for 4 kHz ODR). [Figure 88](#) shows the low-pass filter response with a 1 kHz corner (4 kHz ODR) for the ADXL357/ADXL357B. Note that [Figure 88](#) does not include the fixed frequency analog, low-pass, antialiasing filter with a fixed 3 dB bandwidth of approximately 1.5 kHz.

The ADXL357/ADXL357B pass band of the signal path relates to the combined filter responses, including the analog filter previously described, and the digital decimation filter/ODR setting. [Table 11](#) shows the delay associated with the decimation filter for each setting and provides the attenuation at the ODR/4 corner.

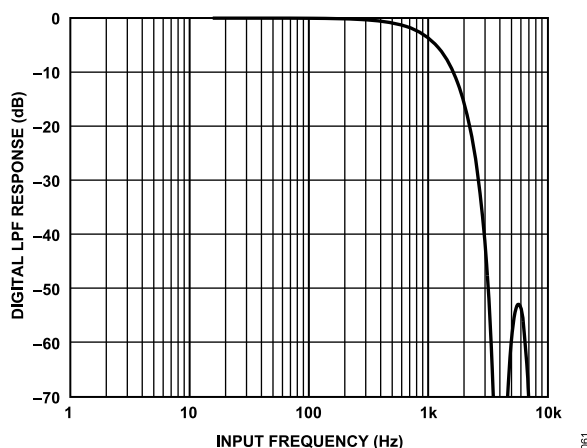


Figure 88. ADXL357 Digital LPF Response for 4 kHz ODR

The ADXL357/ADXL357B also include an optional digital high-pass filter with a programmable corner frequency. By default, the high-pass filter is disabled. The high-pass corner frequency, where the output is attenuated by 3 dB, is related to the ODR, and the HPF\_CORNER setting in the filter register (Register 0x28, Bits[6:4]). [Table 12](#) shows the HPF\_CORNER response. [Figure 89](#) and [Figure 90](#) show the simulated high-pass filter pass-band and delay responses for a 9.88 Hz cutoff.

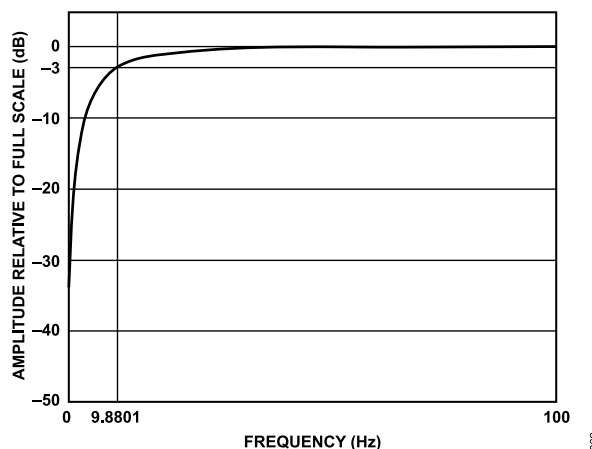


Figure 89. High-Pass Filter Pass-Band Response for a 4 kHz ODR and an HPF\_CORNER Setting of 001 (Register 0x28, Bits[6:4])



## APPLICATIONS INFORMATION

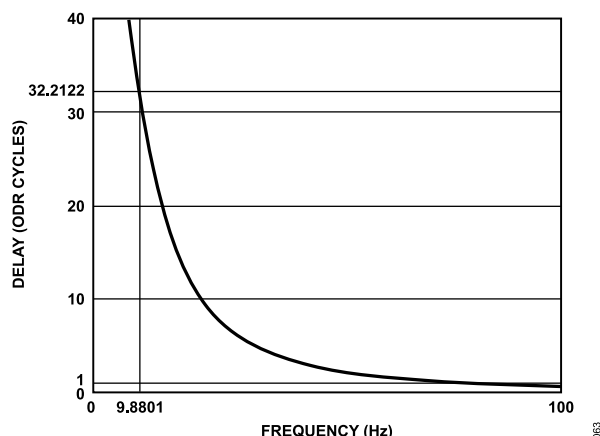


Figure 90. High-Pass Filter Delay Response for a 4 kHz ODR and an HPF\_CORNER Setting of 001 (Register 0x28, Bits[6:4])

The ADXL357/ADXL357B also include an interpolation filter after the decimation filters that produces oversampled/upconverted data and provides an external synchronization option. See the [Data Synchronization](#) section for more details. Table 13 shows the delay and attenuation relative to the programmed ODR.

Group delay is the digital filter delay from the input to the ADC until data is available at the interface (see the [Filter](#) section). This delay is the largest component of the total delay from sensor to serial interface.

Table 11. Digital Filter Group Delay and Profile

Programmed ODR (Hz)	Delay		Attenuation	
	ODR (Cycles)	Time (ms)	Decimator at ODR/4 (dB)	Full Path at ODR/4 (dB)
4000	2.52	0.63	-3.44	-3.63
4000/2 = 2000	2.00	1.00	-2.21	-2.26
4000/4 = 1000	1.78	1.78	-1.92	-1.93
4000/8 = 500	1.63	3.26	-1.83	-1.83
4000/16 = 250	1.57	6.27	-1.83	-1.83
4000/32 = 125	1.54	12.34	-1.83	-1.83
4000/64 = 62.5	1.51	24.18	-1.83	-1.83
4000/128 ≈ 31	1.49	47.59	-1.83	-1.83
4000/256 ≈ 16	1.50	96.25	-1.83	-1.83
4000/512 ≈ 8	1.50	189.58	-1.83	-1.83
4000/1024 ≈ 4	1.50	384.31	-1.83	-1.83

Table 12. Digital High-Pass Filter Response

HPF_CORNER Register Setting (Register 0x28, Bits[6:4])	HPF_CORNER Frequency, -3 dB Point Relative to ODR Setting	-3 dB at 4 kHz ODR (Hz)
000	Not applicable, no high-pass filter enabled	Off
001	$24.7 \times 10^{-4} \times \text{ODR}$	9.88
010	$6.2084 \times 10^{-4} \times \text{ODR}$	2.48
011	$1.5545 \times 10^{-4} \times \text{ODR}$	0.62
100	$0.3862 \times 10^{-4} \times \text{ODR}$	0.1545
101	$0.0954 \times 10^{-4} \times \text{ODR}$	0.03816
110	$0.0238 \times 10^{-4} \times \text{ODR}$	0.00952

Table 13. Combined Digital Interpolation Filter and Decimation Filter Response

Interpolator Data Rate Resolution Relative to $64 \times \text{ODR}$ (Hz)	Combined Interpolator/ Decimator Delay (ODR Cycles)	Combined Interpolator/ Decimator Delay (ms)	Combined Interpolator/Decimator Output Attenuation at ODR/4 (dB)
$64 \times 4000 = 256,000$	3.51661	0.88	-6.18
$64 \times 2000 = 128,000$	3.0126	1.51	-4.93
$64 \times 1000 = 64,000$	2.752	2.75	-4.66
$64 \times 500 = 32,000$	2.6346	5.27	-4.58
$64 \times 250 = 16,000$	2.5773	10.31	-4.55
$64 \times 125 = 8000$	2.5473	20.38	-4.55

## APPLICATIONS INFORMATION

Table 13. Combined Digital Interpolation Filter and Decimation Filter Response (Continued)

Interpolator Data Rate Resolution Relative to $64 \times \text{ODR}$ (Hz)	Combined Interpolator/ Decimator Delay (ODR Cycles)	Combined Interpolator/ Decimator Delay (ms)	Combined Interpolator/Decimator Output Attenuation at ODR/4 (dB)
$64 \times 62.5 = 4000$	2.53257	40.52	-4.55
$64 \times 31.25 = 2000$	2.52452	80.78	-4.55
$64 \times 15.625 = 1000$	2.52045	161.31	-4.55
$64 \times 7.8125 = 500$	2.5194	322.48	-4.55
$64 \times 3.90625 = 250$	2.51714	644.39	-4.55

## SERIAL COMMUNICATIONS

The 4-wire serial interface communicates in either the SPI or I<sup>2</sup>C protocol. The interface affectively autodetects the format being used, requiring no configuration control to select the format.

The ADXL357/ADXL357B multifunction pins are referred to by a single function of the pin, for example,  $\overline{CS}$ , when only that function is relevant.

### SPI PROTOCOL

Wire the ADXL357/ADXL357B for SPI communication as shown in the connection diagram in Figure 91. The SPI protocol timing is shown in Figure 93 to Figure 96. The timing scheme follows the clock polarity (CPOL) = 0 and clock phase (CPHA) = 0. The SPI clock speed ranges from 100 kHz to 10 MHz.

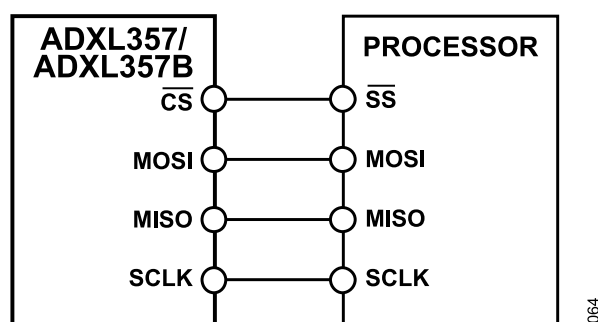


Figure 91. 4-Wire SPI Connection

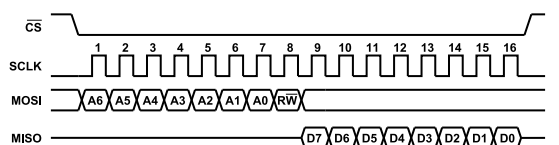


Figure 93. SPI Timing Diagram—Single-Byte Read

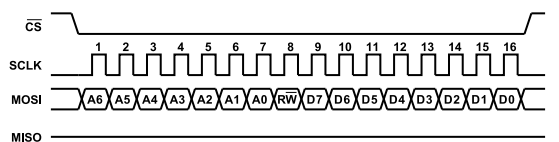


Figure 94. SPI Timing Diagram—Single-Byte Write

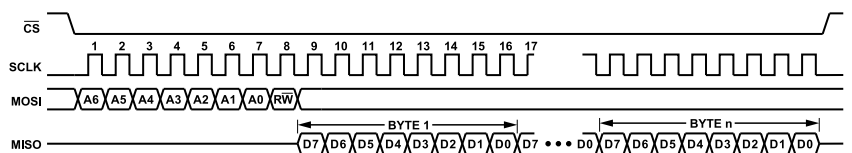


Figure 95. SPI Timing Diagram—Multibyte Read

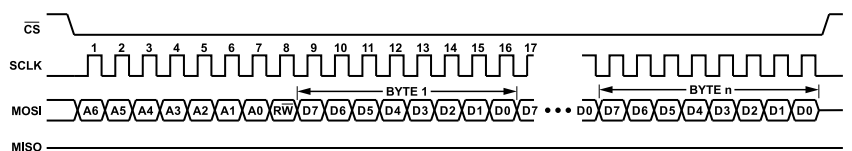


Figure 96. SPI Timing Diagram—Multibyte Write

### SPI BUS SHARING

Use a gated buffer on the SCLK line for the ADXL357/ADXL357B device to achieve the ultralow noise performance and possibly offset shift when the ADXL357/ADXL357B must share a SPI bus with another slave device. This gated SCLK allows the clock signal through only when the chip select ( $\overline{CS}$ ) line is low. See Figure 92 for the example circuit that provides this type of protection.

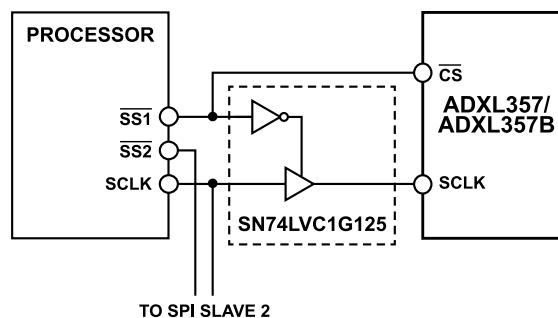


Figure 92. SCLK Protection Example

## SERIAL COMMUNICATIONS

I<sup>2</sup>C PROTOCOL

The ADXL357/ADXL357B support point to point I<sup>2</sup>C communication. However, when sharing an SDA bus, the ADXL357 may prevent communication with other devices on that bus. If at any point, even when the ADXL357 is not being addressed, the 0x3A and 0x3B bytes (when the ADXL357 device address is set to 0x1D), or the 0xA6 and 0xA7 bytes (when the ADXL357 device address is set to 0x53) are transmitted on the SDA bus, the ADXL357 responds with an acknowledge bit and pulls the SDA line down. For example, this response can occur when reading or writing the data bytes (0x3A/0x3B or 0xA6/0xA7) to another sensor on the bus. When the ADXL357 pulls the SDA line down, communication with other devices on the bus may be interrupted. To resolve this interruption, the ADXL357 must be connected to a separate SDA bus, or the CS/SCL pin must be switched high when communication with the ADXL357 is not desired (it is normally grounded).

The ADXL357/ADXL357B support standard (100 kHz), fast (up to 1 MHz) and high speed (up to 3.4 MHz) data transfer modes when the bus parameters in Table 5 are met. There is no minimum SCL frequency, with the exception that, when reading data, the clock must be fast enough to read an entire sample set before new data overwrites it. Single-byte or multiple byte reads/writes are supported. With the MISO/ASEL pin low, the I<sup>2</sup>C address for the device is 0x1D and an alternate I<sup>2</sup>C address of 0x53 can be chosen by pulling the MISO/ASEL pin high.

There are no internal pull-up or pull-down resistors for any unused pins. Therefore, there is no known state or default state for the pins if left floating or unconnected. SCLK/V<sub>SSIO</sub> must be connected to ground when communicating to the ADXL357/ADXL357B using I<sup>2</sup>C.

Due to communication speed limitations, the maximum output data rate when using the 400 kHz I<sup>2</sup>C mode is 800 Hz, and it scales linearly with a change in the I<sup>2</sup>C communication speed. For example, using I<sup>2</sup>C at 100 kHz limits the maximum ODR to 200 Hz. Operation at an output data rate above the recommended maximum may result in an undesirable effect on the acceleration data, including missing samples or additional noise. Figure 97 to Figure 99 show the detail of the I<sup>2</sup>C protocol timing. The I<sup>2</sup>C interface can be used

on most buses operating in I<sup>2</sup>C standard mode (100 kHz), fast mode (400 kHz), fast mode plus (1 MHz), and high speed mode (3.4 MHz). The ADXL357/ADXL357B I<sup>2</sup>C device ID is as follows:

- MISO/ASEL pin = 0, device address = 0x1D
- MISO/ASEL pin = 1, device address = 0x53

If other devices are connected to the same I<sup>2</sup>C bus, the nominal operating voltage level of these other devices cannot exceed V<sub>DDIO</sub> by more than 0.3 V. External pull-up resistors, R<sub>p</sub>, are necessary for proper I<sup>2</sup>C operation.

## READING ACCELERATION OR TEMPERATURE DATA FROM THE INTERFACE

Acceleration data is left justified and has a register address order of most significant data to least significant data, which allows the user to use multibyte transfers and to take only as much data as required—8 bits, 16 bits, or 20 bits, plus the marker. Temperature data is 12 bits unsigned, right justified. The ADXL357/ADXL357B temperature value is split over two bytes, but is not double buffered, meaning the value can update between readings of the two registers. The data in XDATA, YDATA, and ZDATA is always the most recent available. It is not guaranteed that XDATA, YDATA, and ZDATA form a set corresponding to one sample point in time. The routine used to retrieve the data from the device controls this data set continuity. If data transfers are initiated when the DATA\_RDY bit goes high and completes in a time approximately equal to 1/ODR, XDATA, YDATA, and ZDATA apply to the same data set.

For multibyte read or write transactions through either serial interface, the internal register address auto-increments. When the top of the register address range, 0x3FF, is reached, the auto-increment stops and does not wrap back to Address 0x00.

The address auto-increment function disables when the FIFO address is used, so that data can be read continuously from the FIFO as a multibyte transaction. In cases where the starting address of a multibyte transaction is less than the FIFO address, the address auto-increments until reaching the FIFO address, and then stops at the FIFO address.

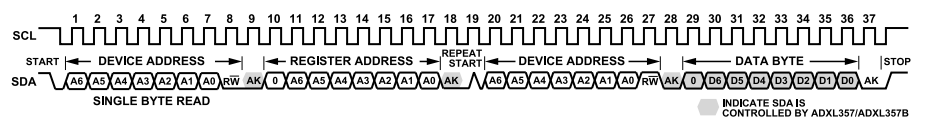


Figure 97. I<sup>2</sup>C Timing Diagram—Single-Byte Read



Figure 98. I<sup>2</sup>C Timing Diagram—Single-Byte Write

SERIAL COMMUNICATIONS

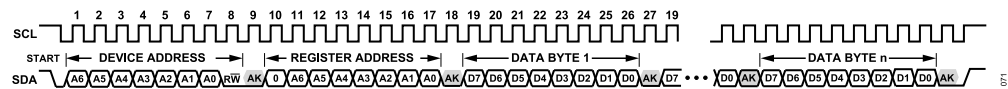


Figure 99. I<sup>2</sup>C Timing Diagram—Multibyte Write

FIFO

The FIFO operates in a stream mode. That is, when the FIFO overruns, new data overwrites the oldest data in the FIFO. A read from the FIFO address guarantees that the three bytes associated with the acceleration measurement on an axis all pertain to the same measurement. The FIFO never overflows, and the data is always taken out in sets (multiples of three data points).

There are 96 21-bit locations in the FIFO. Each location contains 20 bits of data and a marker bit for the x-axis data. A single-byte read from the FIFO address pops one location from the FIFO. A multibyte read to the FIFO location pops the FIFO on the read of the first byte and every third byte read thereafter.

logic inserts the two virtual bits (0b00) between the data bits and the empty indicator bit. Bit 1 indicates that an attempt was made to read an empty FIFO, and that the data is not valid acceleration data. Bit 0 is a marker bit to identify the x-axis, which allows a user to verify that the FIFO data was correctly read. An acceleration data point for a given axis occupies one FIFO location. The read pointer, RD\_PTR, points to the oldest stored data that was not read already from the interface (see Figure 100). There are no physical x-acceleration, y-acceleration, or z-acceleration data registers. The data read from data registers (Register 0x08 to Register 0x10) also comes directly from the most recent data set in the FIFO, which is pointed to by the z pointer, Z\_PTR (see Figure 100).

Figure 100 shows the organization of the data in the FIFO. The acceleration data is twos complement, 20-bit data. The FIFO control

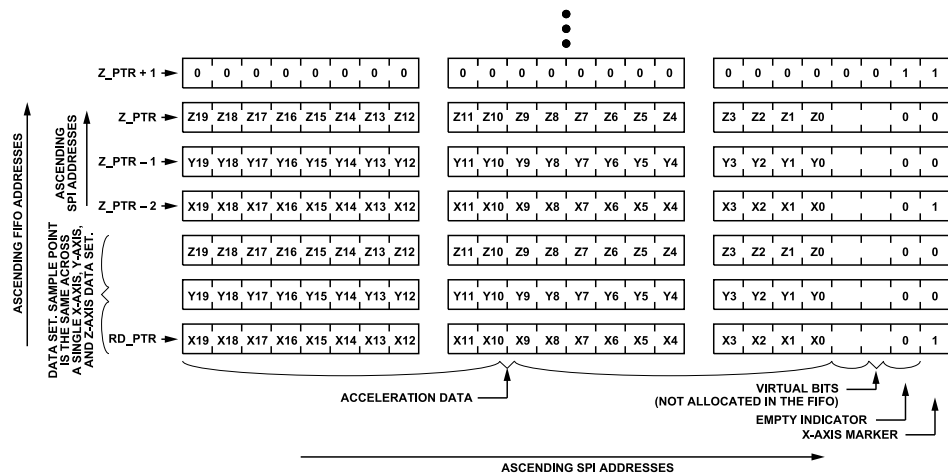


Figure 100. FIFO Data Organization

## INTERRUPTS

The status register (Register 0x04) contains five individual bits, four of which can be mapped to either the INT1 pin, the INT2 pin, or both. The polarity of the interrupt, active high or active low, is also selectable via the INT\_POL bit in the range (Register 0x2C) register. In general, the status register clears when read, but this is not the case if the condition that caused the interrupt persists after the read of the register. The definition of persist varies slightly in each case, but it is described in the [DATA\\_RDY](#), [DRDY Pin](#), [FIFO\\_FULL](#), [FIFO\\_OVR](#), and [Activity](#) sections. The DRDY pin is similar to an interrupt pins (INTx) but clears differently. This case is also described.

### DATA\_RDY

The DATA\_RDY bit is set when new acceleration data is available to the interface and clears on a read of the status register. This bit is not set again until acceleration data that is newer than the status register read is available.

Special logic on the clearing of the DATA\_RDY bit covers the corner case where new data arrives during the read of the status register. In this case, the data ready condition may be missed completely. This logic results in a delay of the clearing of DATA\_RDY of up to four 512 kHz cycles.

### DRDY PIN

The DRDY pin is not a status register bit. DRDY instead behaves similar to an unmaskable interrupt. DRDY is set when new acceleration data is available to the interface. DRDY clears on a read of the FIFO, on a read of XDATA, YDATA, or ZDATA, or by an autoclear function that occurs approximately halfway between output acceleration data sets.

DRDY is always active high. The INT\_POL bit does not affect DRDY. In external synchronization modes (EXT\_SYNC = 01, EXT\_SYNC = 10), the first few DRDY pulses after initial synchronization can be lost or corrupted. The length of this potential corruption is equal to or less than the group delay. Therefore, the samples within one group delay is lost or corrupted after the first synchronization signal. Depending on the decimation setting and interpolation setting (see [Table 13](#)), between one and three samples after the first synchronization pulse is lost, provided that all the restrictions set in the [External Synchronization and Interpolation](#) section is met.

### FIFO\_FULL

The FIFO\_FULL bit is set when the entries in the FIFO are equal to the setting of the FIFO\_SAMPLES bits. FIFO\_FULL clears as follows:

- ▶ If the number of entries in the FIFO is less than the number of samples indicated by the FIFO\_SAMPLES bits, which is only the case if sufficient data is read from the FIFO.
- ▶ On a read of the status register, but only when the entries in the FIFO are less than the FIFO\_SAMPLES bits.

### FIFO\_OVR

The FIFO\_OVR bit is set when the FIFO is so far overrange that data is lost. The specified size of the FIFO is 96 locations. The FIFO\_OVR bit is set only when there is an attempt to write past this 96-location limit.

A read of the status register clears FIFO\_OVR. FIFO\_OVR is not set again until data is lost subsequent to this status register read.

### ACTIVITY

The activity bit (Register 0x04, Bit 3) is set when the measured acceleration on any axis is above the value set in the ACT\_THRESH bits for ACT\_COUNT consecutive measurements. An overthreshold condition can shift from one axis to another on successive measurements and is still counted toward the consecutive ACT\_COUNT count.

A read of the status register clears the activity bit (Register 0x04, Bit 3), but the bit sets again at the end of the next measurement if the activity bit (Register 0x04, Bit 3) conditions are still satisfied.

### NVM\_BUSY

The NVM\_BUSY bit indicates that the nonvolatile memory (NVM) controller is busy and, therefore, the NVM cannot be accessed to read or write. The interrupt functionality requires the NVM\_BUSY bit to be cleared to function.

A status register read that occurs after the NVM controller is no longer busy clears NVM\_BUSY.

## EXTERNAL SYNCHRONIZATION AND INTERPOLATION

There are four possible synchronization options for the ADXL357/ADXL357B, three of which are shown in [Figure 101](#) to [Figure 103](#). For clarity, the clock frequencies and delays are drawn to scale. The labels in [Figure 101](#) to [Figure 103](#) are defined as follows:

- ▶ Internal ODR is the alignment of the decimated output data based on the internal clock.
- ▶ ADC modulator clock shows the internal master clock rate.
- ▶ DRDY is an output indicator signaling a sample is ready.

The four possible synchronization options are as follows:

- ▶ No external synchronization (internal clocks used)
- ▶ Synchronization with an external synchronization signal and internal clock, interpolation filter enabled
- ▶ Synchronization with external synchronization and clock signals, no interpolation filter
- ▶ Synchronization with external synchronization and clock signals, interpolation filter enabled



## INTERRUPTS

### EXT\_SYNC = 00, EXT\_CLK = 0—No External Synchronization or Interpolation

This is the default mode of operation for the device. The sensor runs on an internal ODR and an internal clock that is generated by an internal oscillator. The internal ODR serves as the synchronization master, which generates the data. Register 0x28 is used to program the ODR. No external signals are required, and this mode is used typically when the external processor retrieves data from the device asynchronously and absolute synchronization to an external source is not required.

The device outputs a DRDY (active high) to signal that a new sample is available, and data is retrieved from the real-time registers or the FIFO. The group delay is based on the decimation setting, as shown in Table 11. This mode is shown in Figure 101.

### EXT\_SYNC = 10, EXT\_CLK = 0—External Synchronization With Interpolation

Synchronization using interpolation filters and an external ODR clock is commonly used when the external processor can provide a synchronization signal, SYNC, that is asynchronous to the internal clock at the desired ODR. In this case, an interpolation filter provides additional time resolution of 64 times the programmed ODR (see Table 13). Synchronization with the interpolation filter enabled (EXT\_SYNC = 10) allows the sensor to operate on an internal clock and output data most closely associated with the SYNC rising edge.

The advantage of this mode is that data is available at an arbitrary user defined SYNC sample rate and is asynchronous to the internal clock oscillator. The maximum sample rate cannot exceed 4000 SPS. The disadvantage of this mode is that the group delay is increased, with increased attenuation at the band edge. Additionally, because there is a limit to the time resolution, there is some distortion related to the mismatch of the external synchronization relative to the internal clock oscillator. This mismatch degrades spectral performance. The group delay is based on the decimation setting and interpolation setting (see Table 13). Figure 102 schematically shows the timings in this mode, and Table 14 shows the delay between the SYNC signal (input) to DRDY (output).

Table 14. EXT\_SYNC = 10, DRDY Delay

ODR_LPF	SYNC to DRDY Delay (Oscillator Cycles)
0x0	8
0x1	10
0x2	14
0x3	22
0x4	38
0x5	70
0x6	134
0x7	262

Table 14. EXT\_SYNC = 10, DRDY Delay (Continued)

ODR_LPF	SYNC to DRDY Delay (Oscillator Cycles)
0x8	1031
0x9	2054
0xA	4102

### EXT\_SYNC = 01, EXT\_CLK = 1—External Synchronization and External Clock, No Interpolation Filter

When configured for EXT\_SYNC = 01 and EXT\_CLK = 1 (sync register, see Table 48), the user must supply an external clock (enabled via the EXT\_CLK bit) at 1.024 MHz on the INT2 pin (Pin 13) and an external synchronization signal, SYNC, on the DRDY pin (Pin 14), as shown in Table 15. If configured in this mode and an external clock is not supplied, the device does not process any data and reading from the output results in null values. This mode is schematically shown in Figure 103.

Special restrictions when using this mode include the following:

- The external clock frequency on INT2 (Pin 13, see Table 15) must be 1.024 MHz.
- The pulse width of the SYNC signal must be at least 3.91  $\mu$ s, which represents four cycles of the external clock ( $4 \div 1.024$  MHz =  $\sim 3.91$   $\mu$ s).
- The phase of SYNC must meet an approximate 25 ns setup time to the external clock rising edge.

When using the EXT\_SYNC mode and without providing the SYNC signal, the device runs on its own internal ODR. Similarly, after external synchronization, the device continues to run synchronized to the last SYNC pulse it received, which means that EXT\_SYNC = 01 mode can be used with only a single synchronization pulse.

For more information about the lost sample in Figure 103, see the DRDY Pin section.

### EXT\_SYNC = 10, EXT\_CLK = 1—External Synchronization and External Clock, With Interpolation Filter

This mode can be used to run the device on an external clock and synchronization with an arbitrary sample rate set by the SYNC signal rate. Conditions for external SYNC and external clock signals is the same as EXT\_SYNC = 01, EXT\_CLK = 1 mode. The interpolation filter provides a frequency resolution related to the ODR (see Table 13). In this case, the data provided corresponds to the external SYNC signal, which can be greater than the set ODR and less than 4000 SPS, but the output pass band remains the same it was prior to the interpolation filter.



## INTERRUPTS

Table 15. Multiplexing of INT2 and DRDY

Register or Bit Fields			Pins		Comments
EXT_CLK	EXT_SYNC, Bits[1:0]	INT_MAP, Bits[7:4]	INT2 (Pin 13)	DRDY (Pin 14)	
0	00	0000	Low	DRDY	Synchronization is to the internal clocks, and there is no external clock synchronization.
0	00	Not 0000	INT2	DRDY	
1	00	0000	EXT_CLK	DRDY	
1	00	Not 0000 <sup>1</sup>	EXT_CLK	DRDY	
0	01	0000	DRDY <sup>2</sup>	SYNC	These options reset the digital filters on every synchronization pulse and are not recommended.
0	01 <sup>3</sup>	Not 0000	INT2	SYNC	
1	01 <sup>3</sup>	0000	EXT_CLK	SYNC	External synchronization, no interpolation filter, and DRDY (active high) signals that data is ready. Data represents a sample point group delay earlier in time.
1	01 <sup>3</sup>	Not 0000 <sup>1</sup>	EXT_CLK	SYNC	
0	10	0000	DRDY <sup>2</sup>	SYNC	External synchronization, interpolation filter, and DRDY (active high) signals that data is ready. Data sample group delay earlier in time.
0	10 <sup>3</sup>	Not 0000	INT2	SYNC	
1	10 <sup>3</sup>	0000	EXT_CLK	SYNC	
1	10 <sup>3</sup>	Not 0000	EXT_CLK	SYNC	

<sup>1</sup> No INT2, even though it is enabled.

<sup>2</sup> DRDY routing through the INT\_MAP register takes precedence over the default, per Table 14.

<sup>3</sup> No DRDY.

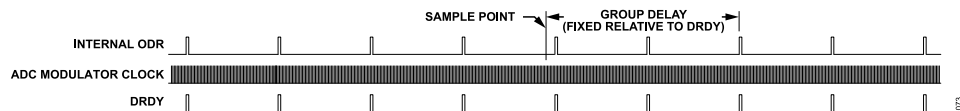


Figure 101. EXT\_SYNC = 00, EXT\_CLK = 0, Internal Synchronization, Internal Clock

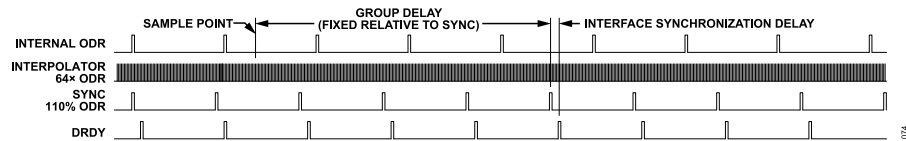


Figure 102. EXT\_SYNC = 10, EXT\_CLK = 0, External Synchronization, Internal Clock, Interpolation Filter

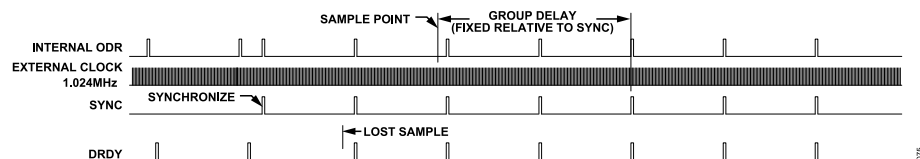


Figure 103. EXT\_SYNC = 01, EXT\_CLK = 1, External Synchronization, External Clock, No Interpolation Filter

## ADXL357/ADXL357B REGISTER MAP

Note that while configuring the ADXL357/ADXL357B in an application, all configuration registers must be programmed before enabling measurement mode in the POWER\_CTL register. When the ADXL357/ADXL357B are in measurement mode, only the following configurations can change: the HPF\_CORNER bits in the filter register, the INT\_MAP register, the ST1 and ST2 bits in the SELF\_TEST register, and the reset register.

Table 16. ADXL357/ADXL357B Register Map

Hex. Addr.	Register Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Reset	R/W		
0x00	DEVID_AD	DEVID_AD								0xAD	R		
0x01	DEVID_MST	DEVID_MST								0x1D	R		
0x02	PARTID	PARTID								0xED <sup>1</sup>	R		
										0xE9 <sup>2</sup>	R		
0x03	REVID	REVID								0x01 <sup>1</sup>	R		
										0x00 <sup>2</sup>	R		
0x04	Status	Reserved			NVM_BUSY	Activity	FIFO_OVR	FIFO_FULL	DATA_RDY	0x00	R		
0x05	FIFO_ENTRIES	Reserved	FIFO_ENTRIES									0x00	R
0x06	TEMP2	Reserved				Temperature, Bits[11:8]					0x00	R	
0x07	TEMP1					Temperature, Bits[7:0]				0x00	R		
0x08	XDATA3					XDATA, Bits[19:12]				0x00	R		
0x09	XDATA2					XDATA, Bits[11:4]				0x00	R		
0x0A	XDATA1	XDATA, Bits[3:0]				Reserved				0x00	R		
0x0B	YDATA3					YDATA, Bits[19:12]				0x00	R		
0x0C	YDATA2					YDATA, Bits[11:4]				0x00	R		
0x0D	YDATA1	YDATA, Bits[3:0]				Reserved				0x00	R		
0x0E	ZDATA3					ZDATA, Bits[19:12]				0x00	R		
0x0F	ZDATA2					ZDATA, Bits[11:4]				0x00	R		
0x10	ZDATA1	ZDATA, Bits[3:0]				Reserved				0x00	R		
0x11	FIFO_DATA					FIFO_DATA				0x00	R		
0x1E	OFFSET_X_H					OFFSET_X, Bits[15:8]				0x00	R/W		
0x1F	OFFSET_X_L					OFFSET_X, Bits[7:0]				0x00	R/W		
0x20	OFFSET_Y_H					OFFSET_Y, Bits[15:8]				0x00	R/W		
0x21	OFFSET_Y_L					OFFSET_Y, Bits[7:0]				0x00	R/W		
0x22	OFFSET_Z_H					OFFSET_Z, Bits[15:8]				0x00	R/W		
0x23	OFFSET_Z_L					OFFSET_Z, Bits[7:0]				0x00	R/W		
0x24	ACT_EN	Reserved					ACT_Z	ACT_Y	ACT_X	0x00	R/W		
0x25	ACT_THRESH_H					ACT_THRESH, Bits[15:8]				0x00	R/W		
0x26	ACT_THRESH_L					ACT_THRESH, Bits[7:0]				0x00	R/W		
0x27	ACT_COUNT					ACT_COUNT				0x01	R/W		
0x28	Filter	Reserved	HPF_CORNER				ODR_LPF				0x00	R/W	
0x29	FIFO_SAMPLES	Reserved	FIFO_SAMPLES									0x60	R/W
0x2A	INT_MAP	ACT_EN2	OVR_EN2	FULL_EN2	RDY_EN2	ACT_EN1	OVR_EN1	FULL_EN1	RDY_EN1	0x00	R/W		
0x2B	Sync	Reserved					EXT_CLK	EXT_SYNC		0x00	R/W		
0x2C	Range	I2C_HS	INT_POL	Reserved				Range		0x81	R/W		
0x2D	POWER_CTL	Reserved					DRDY_OFF	TEMP_OFF	Standby	0x01	R/W		
0x2E	SELF_TEST	Reserved						ST2	ST1	0x00	R/W		
0x2F	Reset	Reset								0x00	W		

<sup>1</sup> For the ADXL357.<sup>2</sup> For the ADXL357B.

## REGISTER DEFINITIONS

This section describes the functions of the ADXL357/ADXL357B registers. The ADXL357/ADXL357B power up with the default register values, as shown in the reset column of [Table 16](#).

### ANALOG DEVICES ID REGISTER

This register contains the Analog Devices ID, 0xAD.

**Address: 0x00, Reset: 0xAD, Name: DEVID\_AD**

*Table 17. Bit Descriptions for DEVID\_AD*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	DEVID_AD		Analog Devices ID	0xAD	R

### ANALOG DEVICES MEMS ID REGISTER

This register contains the Analog Devices MEMS ID, 0x1D.

**Address: 0x01, Reset: 0x1D, Name: DEVID\_MST**

*Table 18. Bit Descriptions for DEVID\_MST*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	DEVID_MST		Analog Devices MEMS ID	0x1D	R

### DEVICE ID REGISTER

This register contains the device ID, 0xED (355 octal).

**Address: 0x02, Reset: 0xED/0xE9, Name: PARTID**

*Table 19. Bit Descriptions for PARTID*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	PARTID		Device ID	0xED (for the ADXL357), 0xE9 (for the ADXL357B)	R

### PRODUCT REVISION ID REGISTER

This register contains the product revision ID, beginning with 0x00 and incrementing for each subsequent revision.

**Address: 0x03, Reset: 0x01, Name: REVID**

*Table 20. Bit Descriptions for REVID*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	REVID		Mask revision	0x01 (for the ADXL357); 0x00 (for the ADXL357B)	R

### STATUS REGISTER

This register includes bits that describe the various conditions of the ADXL357/ADXL357B.

## REGISTER DEFINITIONS

**Address: 0x04, Reset: 0x00, Name: Status**

Table 21. Bit Descriptions for Status

Bits	Bit Name	Settings	Description	Reset	Access
[7:5]	Reserved		Reserved.	0x0	R
4	NVM_BUSY		NVM controller is busy with a refresh, programming, or a built in self test (BIST).	0x0	R
3	Activity		Activity, as defined in the ACT_THRESH_x and ACT_COUNT registers, is detected.	0x0	R
2	FIFO_OVR		FIFO has overrun, and the oldest data is lost.	0x0	R
1	FIFO_FULL		FIFO watermark is reached.	0x0	R
0	DATA_RDY		A complete x-axis, y-axis, and z-axis measurement was made and results can be read.	0x0	R

## FIFO ENTRIES REGISTER

This register indicates the number of valid data samples present in the FIFO buffer. This number ranges from 0 to 96.

**Address: 0x05, Reset: 0x00, Name: FIFO\_ENTRIES**

Table 22. Bit Descriptions for FIFO\_ENTRIES

Bits	Bit Name	Settings	Description	Reset	Access
7	Reserved		Reserved	0x0	R
[6:0]	FIFO_ENTRIES		Number of data samples stored in the FIFO	0x0	R

## TEMPERATURE DATA REGISTERS

These two registers contain the uncalibrated temperature data. The nominal intercept is 1885 LSB at 25°C and the nominal slope is -9.05 LSB/°C. TEMP2 contains the four most significant bits, and TEMP1 contains the eight least significant bits of the 12-bit value. The ADXL357/ADXL357B temperature value is not double buffered, meaning the value can update between reading of the two registers.

**Address: 0x06, Reset: 0x00, Name: TEMP2**

Table 23. Bit Descriptions for TEMP2

Bits	Bit Name	Settings	Description	Reset	Access
[7:4]	Reserved		Reserved		
[3:0]	Temperature, Bits[11:8]		Uncalibrated temperature data	0x0	R

**Address: 0x07, Reset: 0x00, Name: TEMP1**

Table 24. Bit Descriptions for TEMP1

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	Temperature, Bits[7:0]		Uncalibrated temperature data	0x00	R

## X-AXIS DATA REGISTERS

These three registers contain the x-axis acceleration data. Data is left justified and formatted as twos complement.

**Address: 0x08, Reset: 0x00, Name: XDATA3**

Table 25. Bit Descriptions for XDATA3

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	XDATA, Bits[19:12]		X-axis data	0x00	R

## REGISTER DEFINITIONS

**Address: 0x09, Reset: 0x00, Name: XDATA2**

Table 26. Bit Descriptions for XDATA2

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	XDATA, Bits[11:4]		X-axis data	0x00	R

**Address: 0x0A, Reset: 0x00, Name: XDATA1**

Table 27. Bit Descriptions for XDATA1

Bits	Bit Name	Settings	Description	Reset	Access
[7:4]	XDATA, Bits[3:0]		X-axis data	0x0	R
[3:0]	Reserved		Reserved	0x0	R

## Y-AXIS DATA REGISTERS

These three registers contain the y-axis acceleration data. Data is left justified and formatted as twos complement.

**Address: 0x0B, Reset: 0x00, Name: YDATA3**

Table 28. Bit Descriptions for YDATA3

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	YDATA, Bits[19:12]		Y-axis data	0x00	R

**Address: 0x0C, Reset: 0x00, Name: YDATA2**

Table 29. Bit Descriptions for YDATA2

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	YDATA, Bits[11:4]		Y-axis data	0x00	R

**Address: 0x0D, Reset: 0x00, Name: YDATA1**

Table 30. Bit Descriptions for YDATA1

Bits	Bit Name	Settings	Description	Reset	Access
[7:4]	YDATA, Bits[3:0]		Y-axis data	0x0	R
[3:0]	Reserved		Reserved	0x0	R

## Z-AXIS DATA REGISTERS

These three registers contain the z-axis acceleration data. Data is left justified and formatted as twos complement.

**Address: 0x0E, Reset: 0x00, Name: ZDATA3**

Table 31. Bit Descriptions for ZDATA3

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	ZDATA, Bits[19:12]		Z-axis data	0x00	R

**Address: 0x0F, Reset: 0x00, Name: ZDATA2**

Table 32. Bit Descriptions for ZDATA2

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	ZDATA, Bits[11:4]		Z-axis data	0x00	R

## REGISTER DEFINITIONS

**Address: 0x10, Reset: 0x00, Name: ZDATA1***Table 33. Bit Descriptions for ZDATA1*

Bits	Bit Name	Settings	Description	Reset	Access
[7:4]	ZDATA, Bits[3:0]		Z-axis data	0x0	R
[3:0]	Reserved		Reserved	0x0	R

## FIFO ACCESS REGISTER

**Address: 0x11, Reset: 0x00, Name: FIFO\_DATA**

Read this register to access data stored in the FIFO.

*Table 34. Bit Descriptions for FIFO\_DATA*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	FIFO_DATA		FIFO data is formatted to 24 bits, three bytes, most significant byte first. A read to this address pops an effective three equal byte words of axis data from the FIFO. Two subsequent reads or a multibyte read completes the transaction of this data onto the interface. Continued reading or a sustained multibyte read of this field continues to pop the FIFO every third byte. Multibyte reads to this address do not increment the address pointer. If this address is read due to an auto-increment from the previous address, it does not pop the FIFO. Instead, it returns zeros and increments on to the next address.	0x0	R

## X-AXIS OFFSET TRIM REGISTERS

**Address: 0x1E, Reset: 0x00, Name: OFFSET\_X\_H***Table 35. Bit Descriptions for OFFSET\_X\_H*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	OFFSET_X, Bits[15:8]		Offset added to x-axis data after all other signal processing. Data is in two's complement format. The significance of OFFSET_X, Bits[15:0] matches the significance of XDATA, Bits[19:4].	0x0	R/W

**Address: 0x1F, Reset: 0x00, Name: OFFSET\_X\_L***Table 36. Bit Descriptions for OFFSET\_X\_L*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	OFFSET_X, Bits[7:0]		Offset added to x-axis data after all other signal processing. Data is in two's complement format. The significance of OFFSET_X, Bits[15:0] matches the significance of XDATA, Bits[19:4].	0x0	R/W

## Y-AXIS OFFSET TRIM REGISTERS

**Address: 0x20, Reset: 0x00, Name: OFFSET\_Y\_H***Table 37. Bit Descriptions for OFFSET\_Y\_H*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	OFFSET_Y, Bits[15:8]		Offset added to y-axis data after all other signal processing. Data is in two's complement format. The significance of OFFSET_Y, Bits[15:0] matches the significance of YDATA, Bits[19:4].	0x0	R/W

**Address: 0x21, Reset: 0x00, Name: OFFSET\_Y\_L***Table 38. Bit Descriptions for OFFSET\_Y\_L*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	OFFSET_Y, Bits[7:0]		Offset added to y-axis data after all other signal processing. Data is in two's complement format. The significance of OFFSET_Y, Bits[15:0] matches the significance of YDATA, Bits[19:4].	0x0	R/W

## REGISTER DEFINITIONS

## Z-AXIS OFFSET TRIM REGISTERS

**Address: 0x22, Reset: 0x00, Name: OFFSET\_Z\_H***Table 39. Bit Descriptions for OFFSET\_Z\_H*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	OFFSET_Z, Bits[15:8]		Offset added to z-axis data after all other signal processing. Data is in two's complement format. The significance of OFFSET_Z, Bits[15:0] matches the significance of ZDATA, Bits[19:4].	0x0	R/W

**Address: 0x23, Reset: 0x00, Name: OFFSET\_Z\_L***Table 40. Bit Descriptions for OFFSET\_Z\_L*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	OFFSET_Z, Bits[7:0]		Offset added to z-axis data after all other signal processing. Data is in two's complement format. The significance of OFFSET_Z, Bits[15:0] matches the significance of ZDATA, Bits[19:4].	0x0	R/W

## ACTIVITY ENABLE REGISTER

**Address: 0x24, Reset: 0x00, Name: ACT\_EN***Table 41. Bit Descriptions for ACT\_EN*

Bits	Bit Name	Settings	Description	Reset	Access
[7:3]	Reserved		Reserved.	0x0	R
2	ACT_Z		Z-axis data is a component of the activity detection algorithm.	0x0	R/W
1	ACT_Y		Y-axis data is a component of the activity detection algorithm.	0x0	R/W
0	ACT_X		X-axis data is a component of the activity detection algorithm.	0x0	R/W

## ACTIVITY THRESHOLD REGISTERS

**Address: 0x25, Reset: 0x00, Name: ACT\_THRESH\_H***Table 42. Bit Descriptions for ACT\_THRESH\_H*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	ACT_THRESH, Bits[15:8]		Threshold for activity detection. Acceleration magnitude must be above ACT_THRESH to trigger the activity counter. ACT_THRESH is an unsigned magnitude. The significance of ACT_THRESH, Bits[15:0] matches the significance of Bits[18:3] of XDATA, YDATA, and ZDATA.	0x0	R/W

**Address: 0x26, Reset: 0x00, Name: ACT\_THRESH\_L***Table 43. Bit Descriptions for ACT\_THRESH\_L*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	ACT_THRESH, Bits[7:0]		Threshold for activity detection. The acceleration magnitude must be greater than the value in ACT_THRESH to trigger the activity counter. ACT_THRESH is an unsigned magnitude. The significance of ACT_THRESH, Bits[15:0] matches the significance of Bits[18:3] of XDATA, YDATA, and ZDATA.	0x0	R/W

## REGISTER DEFINITIONS

## ACTIVITY COUNT REGISTER

**Address: 0x27, Reset: 0x01, Name: ACT\_COUNT**

*Table 44. Bit Descriptions for ACT\_COUNT*

Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	ACT_COUNT		Number of consecutive events above threshold (from ACT_THRESH) required to detect activity	0x1	R/W

## FILTER SETTINGS REGISTER

**Address: 0x28, Reset: 0x00, Name: Filter**

Use this register to specify parameters for the internal high-pass and low-pass filters.

*Table 45. Bit Descriptions for Filter*

Bits	Bit Name	Settings	Description	Reset	Access
7	Reserved		Reserved	0x0	R
[6:4]	HPF_CORNER		−3 dB filter corner for the first-order, high-pass filter relative to the ODR	0x0	R/W
		000	Not applicable, no high-pass filter enabled		
		001	$24.7 \times 10^{-4} \times \text{ODR}$		
		010	$6.2084 \times 10^{-4} \times \text{ODR}$		
		011	$1.5545 \times 10^{-4} \times \text{ODR}$		
		100	$0.3862 \times 10^{-4} \times \text{ODR}$		
		101	$0.0954 \times 10^{-4} \times \text{ODR}$		
		110	$0.0238 \times 10^{-4} \times \text{ODR}$		
[3:0]	ODR_LPF		ODR and low-pass filter corner	0x0	R/W
		0000	4000 Hz and 1000 Hz		
		0001	2000 Hz and 500 Hz		
		0010	1000 Hz and 250 Hz		
		0011	500 Hz and 125 Hz		
		0100	250 Hz and 62.5 Hz		
		0101	125 Hz and 31.25 Hz		
		0110	62.5 Hz and 15.625 Hz		
		0111	31.25 Hz and 7.813 Hz		
		1000	15.625 Hz and 3.906 Hz		
		1001	7.813 Hz and 1.953 Hz		
		1010	3.906 Hz and 0.977 Hz		

## FIFO SAMPLES REGISTER

**Address: 0x29, Reset: 0x60, Name: FIFO\_SAMPLES**

Use the FIFO\_SAMPLES value to specify the number of samples to store in the FIFO. The default value of this register is 0x60 to avoid triggering the FIFO watermark interrupt.

*Table 46. Bit Descriptions for FIFO\_SAMPLES*

Bits	Bit Name	Settings	Description	Reset	Access
7	Reserved		Reserved.	0x0	R
[6:0]	FIFO_SAMPLES		Watermark number of samples stored in the FIFO that triggers a FIFO_FULL condition. Values range from 1 to 96.	0x60	R/W



## REGISTER DEFINITIONS

## INTERRUPT PIN (INTX) FUNCTION MAP REGISTER

**Address: 0x2A, Reset: 0x00, Name: INT\_MAP**

The INT\_MAP register configures the interrupt pins. Bits[7:0] select which functions generate an interrupt on the INT1 and INT2 pins. Multiple events can be configured. If the corresponding bit is set to 1, the function generates an interrupt on the interrupt pins.

**Table 47. Bit Descriptions for INT\_MAP**

Bits	Bit Name	Settings	Description	Reset	Access
7	ACT_EN2		Activity interrupt enable on INT2	0x0	R/W
6	OVR_EN2		FIFO_OVR interrupt enable on INT2	0x0	R/W
5	FULL_EN2		FIFO_FULL interrupt enable on INT2	0x0	R/W
4	RDY_EN2		DATA_RDY interrupt enable on INT2	0x0	R/W
3	ACT_EN1		Activity interrupt enable on INT1	0x0	R/W
2	OVR_EN1		FIFO_OVR interrupt enable on INT1	0x0	R/W
1	FULL_EN1		FIFO_FULL interrupt enable on INT1	0x0	R/W
0	RDY_EN1		DATA_RDY interrupt enable on INT1	0x0	R/W

## DATA SYNCHRONIZATION

**Address: 0x2B, Reset: 0x00, Name: Sync**

Use this register to control the external timing triggers.

**Table 48. Bit Descriptions for Sync**

Bits	Bit Name	Settings	Description	Reset	Access
[7:3]	Reserved		Reserved.	0x0	R
2	EXT_CLK		Enable external clock. See <a href="#">Table 15</a> for configuration details.	0x0	R/W
[1:0]	EXT_SYNC		Enable external synchronization control.	0x0	R/W
		00	Internal synchronization.		
		01	External synchronization, no interpolation filter. After synchronization, and for EXT_SYNC within specification, DATA_RDY occurs on EXT_SYNC.		
		10	External synchronization, interpolation filter, next available data indicated by DATA_RDY 14 to 8204 oscillator cycles later (longer delay for higher ODR_LPF setting), data represents a sample point group delay earlier in time.		
		11	Reserved.		

I<sup>2</sup>C SPEED, INTERRUPT POLARITY, AND RANGE REGISTER**Address: 0x2C, Reset: 0x81, Name: Range****Table 49. Bit Descriptions for Range**

Bits	Bit Name	Settings	Description	Reset	Access
7	I2C_HS		I <sup>2</sup> C speed. 1 High speed mode. 0 Fast mode.	0x1	R/W
6	INT_POL		Interrupt polarity. 0 INT1 and INT2 are active low. 1 INT1 and INT2 are active high.	0x0	R/W
[5:2]	Reserved		Reserved.	0x0	R
[1:0]	Range		Range. 01 ±10 g.	0x1	R/W

## REGISTER DEFINITIONS

Table 49. Bit Descriptions for Range (Continued)

Bits	Bit Name	Settings	Description	Reset	Access
		10	$\pm 20\text{ g}$ .		
		11	$\pm 40\text{ g}$ .		

## POWER CONTROL REGISTER

Address: 0x2D, Reset: 0x01, Name: POWER\_CTL

Table 50. Bit Descriptions for POWER\_CTL

Bits	Bit Name	Settings	Description	Reset	Access
[7:3]	Reserved		Reserved.	0x0	R
2	DRDY_OFF		Set to 1 to force the DRDY output to 0 in modes where it is normally signal data ready.	0x0	R/W
1	TEMP_OFF		Set to 1 to disable temperature processing. Temperature processing is also disabled when standby = 1.	0x0	R/W
0	Standby		Standby or measurement mode.  Standby mode. In standby mode, the device is in a low power state, and the temperature and acceleration datapaths are not operating. In addition, digital functions, including FIFO pointers, reset. Changes to the configuration setting of the device must be made when standby = 1. An exception is a high-pass filter that can be changed when the device is operating.  1 0 Measurement mode.	0x1	R/W

## SELF TEST REGISTER

Address: 0x2E, Reset: 0x00, Name: SELF\_TEST

Refer to the [Mechanical Headroom vs. Frequency](#) section for more information on the operation of the self test feature.

Table 51. Bit Descriptions for SELF\_TEST

Bits	Bit Name	Settings	Description	Reset	Access
[7:2]	Reserved		Reserved.	0x0	R
1	ST2		Set to 1 to enable self test force	0x0	R/W
0	ST1		Set to 1 to enable self test mode	0x0	R/W

## RESET REGISTER

Address: 0x2F, Reset: 0x00, Name: Reset

Table 52. Bit Descriptions for Reset

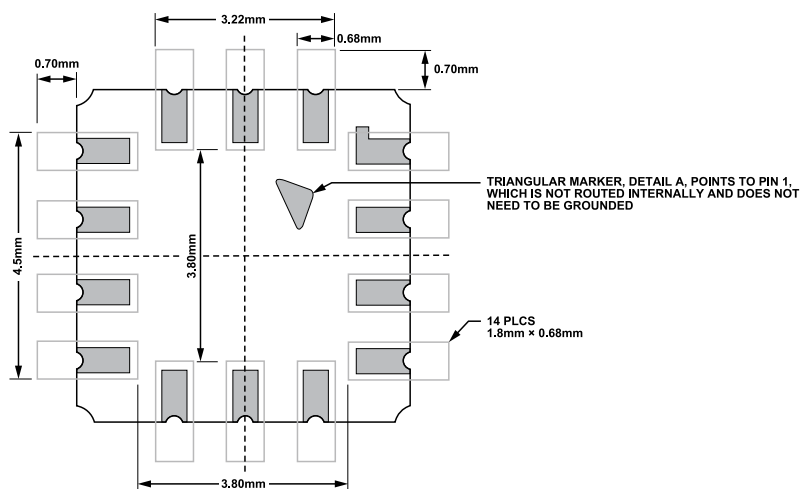
Bits	Bit Name	Settings	Description	Reset	Access
[7:0]	Reset		Write Code 0x52 to reset the device, similar to a power-on reset (POR)	0x0	W

For the ADXL357 (fixed on the ADXL357B), in case of a software reset, an unlikely race condition may occur. If the race condition occurs, some factory settings in the NVM load incorrectly to shadow registers (the registers from which the internal logic configures the sensor and calculates the output after a power-on or a software reset). The incorrect loading of the NVM affects overall performance of the sensor, such as an incorrect  $0\text{ g}$  bias and other performance issues. The incorrect loading of NVM does not occur from a power-on or after a power cycle. To guarantee reliable operation of the sensor after a software reset, the user can access the shadow registers after a power-on, read and store the values on the host microprocessor, and compare the values read from the same shadow registers after a software reset. This method guarantees proper operation in all devices and under all conditions. The recommended steps are as follows:

1. Read the shadow registers, Register 0x50 to Register 0x54 (five 8-bit registers) after power-up, but before any software reset.
2. Store these values in a host device (for example, a host microprocessor).
3. After each software reset, read the same five registers. If the values differ, perform a software reset again until they match.

## PCB FOOTPRINT PATTERN

Figure 104 shows the PCB footprint pattern and dimensions in millimeters.



**Figure 104. PCB Footprint Pattern and Dimensions in Millimeters**

## OUTLINE DIMENSIONS

Package Drawing (Option)	Package Type	Package Description
E-14-1	LCC	14-Terminal Ceramic Leadless Chip Carrier

For the latest package outline information and land patterns (footprints), go to [Package Index](#).

Updated: March 13, 2024

## ORDERING GUIDE

Model <sup>1</sup>	Temperature Range	Package Description	Packing Quantity	Package Option
ADXL356BEZ	+40°C to +125°C	14-Lead LCC (6mm x 6mm)		E-14-1
ADXL356BEZ-RL	+40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 2000	E-14-1
ADXL356BEZ-RL7	+40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 500	E-14-1
ADXL356CEZ	+40°C to +125°C	14-Lead LCC (6mm x 6mm)		E-14-1
ADXL356CEZ-RL	+40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 2000	E-14-1
ADXL356CEZ-RL7	+40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 500	E-14-1
ADXL357BEZ	-40°C to +125°C	14-Lead LCC (6mm x 6mm)		E-14-1
ADXL357BEZ-RL	-40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 2000	E-14-1
ADXL357BEZ-RL7	-40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 500	E-14-1
ADXL357BBEZ	-40°C to +125°C	14-Lead LCC (6mm x 6mm)		E-14-1
ADXL357BBEZ-RL	-40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 2000	E-14-1
ADXL357BBEZ-RL7	-40°C to +125°C	14-Lead LCC (6mm x 6mm)	Reel, 500	E-14-1

<sup>1</sup> Z = RoHS Compliant Part.

## OUTPUT MODE, MEASUREMENT RANGE, AND SPECIFIED VOLTAGE OPTIONS

Model <sup>1</sup>	Output Mode	Measurement Range (g)	Specified Voltage (V)
ADXL356BEZ	Analog	±10, ±20	3.3
ADXL356BEZ-RL	Analog	±10, ±20	3.3
ADXL356BEZ-RL7	Analog	±10, ±20	3.3
ADXL356CEZ	Analog	±10, ±40	3.3
ADXL356CEZ-RL	Analog	±10, ±40	3.3
ADXL356CEZ-RL7	Analog	±10, ±40	3.3
ADXL357BEZ	Digital	±10, ±20, ±40	3.3
ADXL357BEZ-RL	Digital	±10, ±20, ±40	3.3
ADXL357BEZ-RL7	Digital	±10, ±20, ±40	3.3
ADXL357BBEZ	Digital	±10, ±20, ±40	3.3
ADXL357BBEZ-RL	Digital	±10, ±20, ±40	3.3
ADXL357BBEZ-RL7	Digital	±10, ±20, ±40	3.3

<sup>1</sup> Z = RoHS Compliant Part.

OUTLINE DIMENSIONS

EVALUATION BOARDS

Model <sup>1</sup>	Description
EVAL-ADXL356BZ	Evaluation Board for ADXL356B
EVAL-ADXL356CZ	Evaluation Board for ADXL356C
EVAL-ADXL357Z	Evaluation Board for ADXL357
EVAL-ADXL357BZ	Evaluation Board for ADXL357B

<sup>1</sup> Z = RoHS Compliant Part.

I<sup>2</sup>C refers to a communications protocol originally developed by Philips Semiconductors (now NXP Semiconductors).