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The ADSP-TS101S TigerSHARC® On-chip SDRAM Controller

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1 Introduction

In systems where a Digital Signal Processor (DSP) is used to address Synchronous Dynamic Random Access Memory (SDRAM), additional hardware and software is needed to handle the multiplexed row and column addressing, as well as the refresh and precharge requirements of the SDRAM. The ADSP-TS101S TigerSHARC® processor uses a hardware intensive solution, an on-chip SDRAM controller.

This Engineer-to-Engineer Note introduces the ADSP-TS101S on-chip SDRAM controller's characteristics. The internal signal chain is shown with the necessary address-mapping scheme. The command truth table gives detailed information about execution in the SDRAM. The important power-up sequence summarizes detail information to start successful designs. A timing overview demonstrates the performance for different access modes. For basic understanding of SDRAM memories, refer to "*The ABC of SDRAMemory (EE-126)*" [3].

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3 Signal Chain of SDRAM

The signal chain between the ADSP-TS101S, the on-chip SDRAM controller, and the external memory device is illustrated in **Figure 1** for a 64-bit bus configuration:

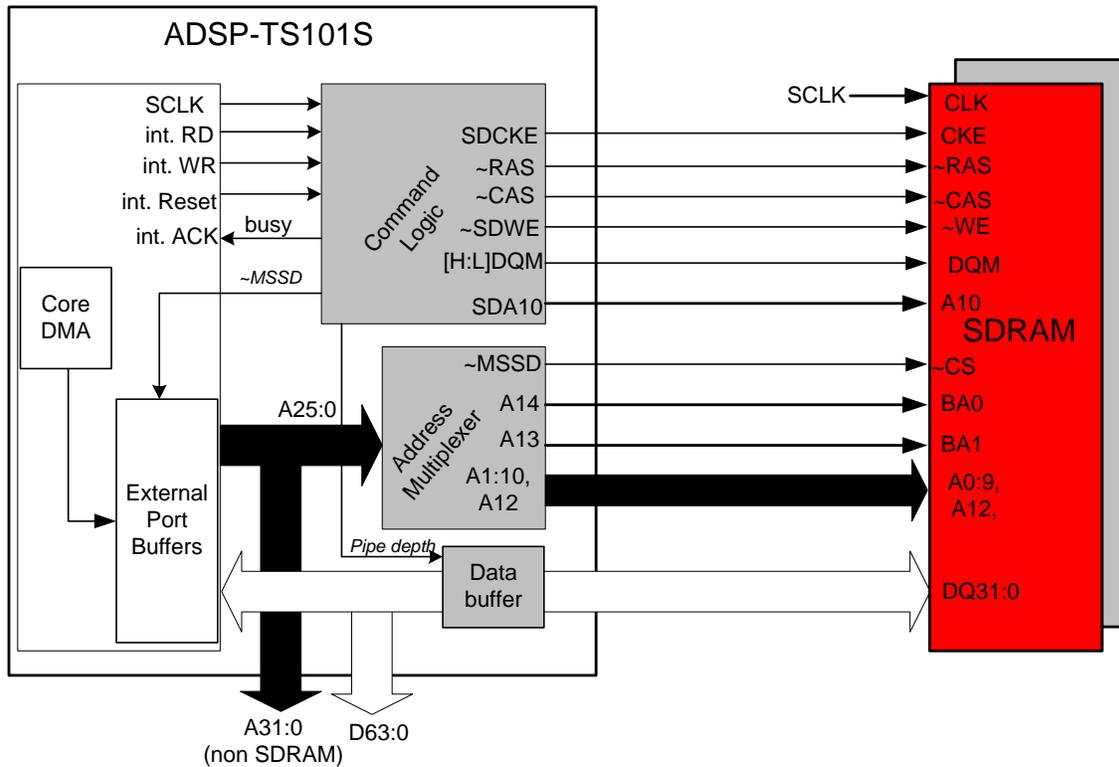


Figure 1. ADSP-TS101 to SDRAM Signal Chain (64-bit bus configuration)

The 3 parts for the signal flow to be considered are:

- ADSP-TS101S (core, DMA engine, I/O-Processor and the address buffer)
- The SDRAM Controller (control interface, delay buffer, and address multiplexer).
- SDRAM device.

4 On-Chip Controller Architecture

The synchronous interface between the ADSP-TS101S and the on-chip controller can be described in 4 basic parts:

4.1 Controller Command Interface

Because of the 2 different timing protocols, the internal TigerSHARC commands are converted to comply with the JEDEC standard for SDRAMs. The external clock, 100 MHz maximum, is used for synchronous operation. The TigerSHARC's internal request lines or strobes are used to access the SDRAM with pulsed commands. The controller's internal ACK line inserts variable wait states to the DSP during overhead cycles, caused by DRAM technology.

4.2 TigerSHARC Output FIFO

The TigerSHARC's output FIFO is active for external port addresses like SDRAM. With the FIFO depth 6, address pipelining for high-speed non-sequential read operations (CAS latency=3) is supported without performance losses.

4.3 Controller Address Multiplexer

Every first read or write action is issued in multiplexed mode. A maximum of 8192 rows (64-bit bus configuration) and 16384 rows (32-bit bus configuration) within 1024 columns can be addressed.

4.4 Controller Data Delay Buffer

If systems incorporate a heavy busload, an additional data buffers is used to decouple the input from the capacitive load. This delay buffer in conjunction with an external buffer for SDRAM control and address lines reduces additional logic to a minimum.

4.5 SDRAM Types

The ADSP-TS101S on-chip SDRAM controller interface supports various LVTTTL SDRAM devices depending on size and internal organization (I/O capability, number of rows, and page size). The following table summarizes all the supported types:

Size	I/O capability	Row x Page	Size	I/O capability	Row x Page
16 Mbits	1M x 16	2k x 256	256 Mbits	8M x 32	8k x 256
	2M x 8	2k x 512		16M x 16	8k x 512
	4M x 4	2k x 1024		32M x 8	8k x 1024
64 Mbits	2M x 32	2k x 256	512 Mbits	8M x 32	8k x 256
	4M x 16	4k x 256		16M x 16	8k x 512
	8M x 8	4k x 512		32M x 8	8k x 1024
	16M x 4	4k x 1024			
128 Mbits	4M x 32	4k x 256			
	8M x 16	4k x 512			
	16M x 8	4k x 1024			

5 Command Coding

5.1 Controller's Pin Definition

Pin	Type	Description
~MSSD	I/O/T (pu)	Memory select signal
~RAS	I/O/T (pu)	Row select signal
~CAS	I/O/T (pu)	Column select signal
~SDWE	I/O/T (pu)	Write enable signal
HDQM	O/T (pu)	Mask data high lane signal
LDQM	O/T (pu)	Mask data low lane signal
SDA10	O/T (pu)	Address10 /command select signal
SDCKE	I/O/T (pu/pd)	Clock enable signal
A[1:10,;12-15]	I/O/T	addresses for 64-bit
A[0:9,11-15]	I/O/T	addresses for 32-bit
A[11:15]	I/O/T	Bank select signal
D[63:0]	I/O/T	Data signals

I = input, O = output, T = Hi-Z, pd = pull-down, pu = pull-up

5.2 Controller Command Truth Table

This section provides a table to get an overview of all commands provided by the SDRAM controller. These commands are automatically handled by the interface.

SDCKE = high								
CMD	SDCKE(n-1)	SDCKE(n)	~MSSD	~RAS	~CAS	~SDWE	SDA10	ADDR
MRS	1	1	0	0	0	0	V	V
ACT	1	1	0	0	1	1	V	V
RD	1	1	0	1	0	1	0	V
WR	1	1	0	1	0	0	0	V

SDCKE = high, no validity of address								
CMD	SDCKE(n-1)	SDCKE(n)	~MSSD	~RAS	~CAS	~SDWE	SDA10	ADDR
NOP	1	1	1	x	x	x	x	x
BST	1	1	0	1	1	0	x	x
REF	1	1	0	0	0	1	1	x
PREA	1	1	0	0	1	0	1	x

Commands with SDCKE transition								
CMD	SDCKE(n-1)	SDCKE(n)	~MSSD	~RAS	~CAS	~SDWE	SDA10	ADDR
SREF E	1	0	0	0	0	1	x	x
SREF M	0	0	x	x	x	X	x	x
SREF X	0	1	1	x	x	x	x	x

X = don't care, v = valid data input, 0 = logic 0, 1 = logic 1, E = entry, M = maintain, X = exit

Although the SDCKE line toggles in an asynchronous manner, the commands are sampled synchronous to the CLK signal.

Note that *Power-down* and *Suspend* modes are not supported, and that the controller does not allow *auto precharge*. Lastly, keep in mind that *all SDRAM commands are fully transparent to the user*.

5.3 Setup and Hold Times

The synchronous operation uses the external clock as a reference. Commands, addresses, and data are latched at the rising edge of clock. The valid time margin around the rising edge is defined as setup time (time before rising edge) and hold time (time after rising edge) to guarantee that both the controller and the SDRAM are working reliably together. Signal's slew rates, propagation delays (PCB), and capacitive loads (devices) influence these parameters and should be taken into consideration. The SDRAM interface AC Signal Specifications can be found in the “*ADSP-TS101S Data Sheet*” [2].

5.4 Simplified State Diagram

The following state diagram (**Figure 2**) shows all possible SDRAM commands sequences to help analyze the controller's functionality.

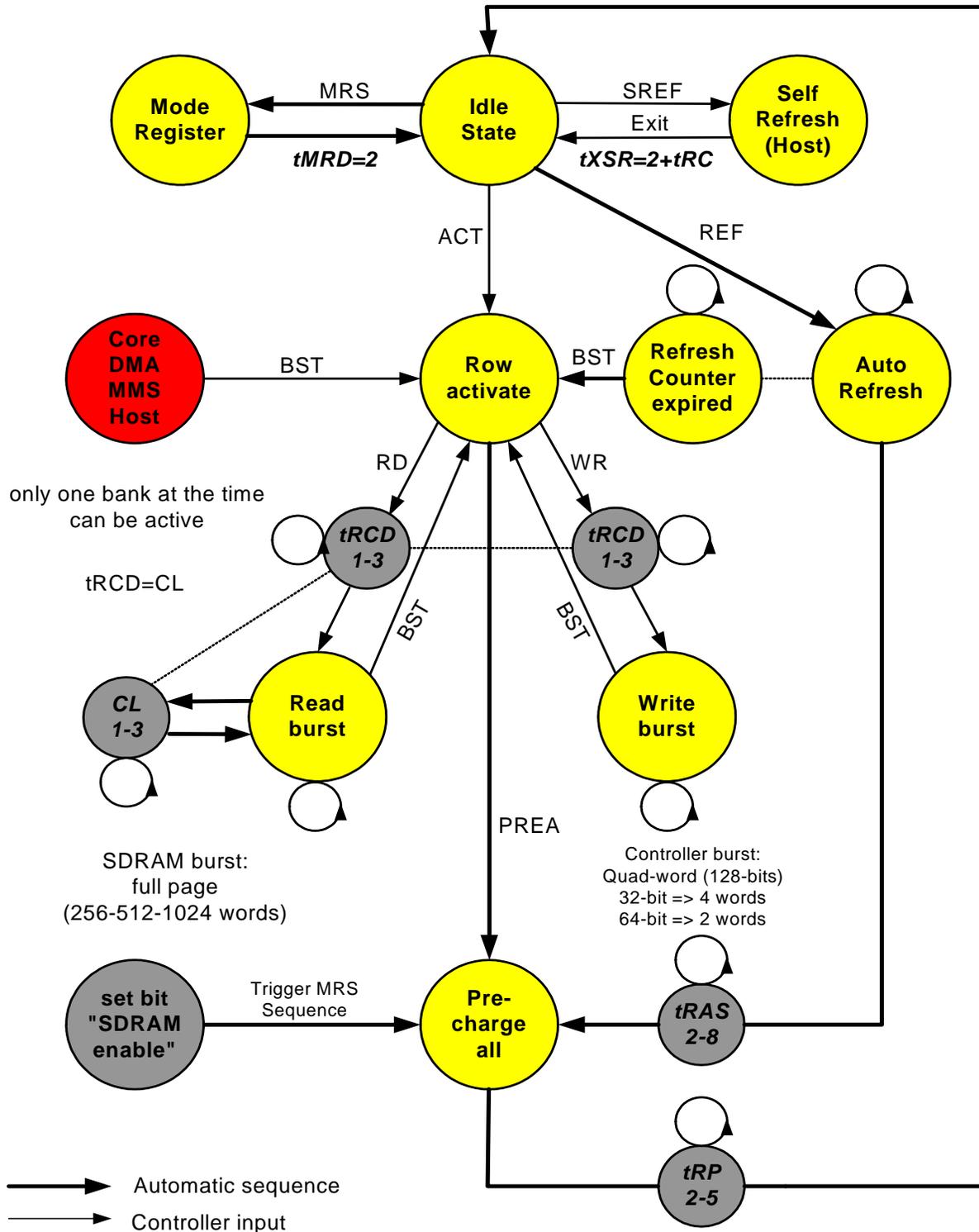


Figure 2 ADSP-TS101S SDRAM Controller Simplified State Diagram

6 SDRAM Controller Properties

Following, the ADSP-TS101S on-chip SDRAM controller properties are examined:

6.1 Address Mapping Scheme

There are various possibilities when accessing the SDRAM. For instance, all rows in a bank can be accessed sequentially, or all banks in a row. PC DIMM modules are accessed in a different manner compared to a typical DSP application. The ADSP-TS101S controller uses a hardware map scheme optimized for digital signal processing.

The address mapping scheme is decoded from the page size and the bus width (both configurable by software in the SDRCON and SYSCON registers respectively; refer to section 7 *SDRAM Programming*). For more information regarding the address mapping scheme, refer to the SDRAM chapter of the *ADSP-TS101 TigerSHARC Processor Hardware Reference* [1].

Figure 3 reproduces an example of the controller's address mapping for 64-bit data. In bank A, the SDRAM's columns are sequentially accessed until the end of the row. Similarly, the SDRAM's rows are sequentially selected until the bank's end.

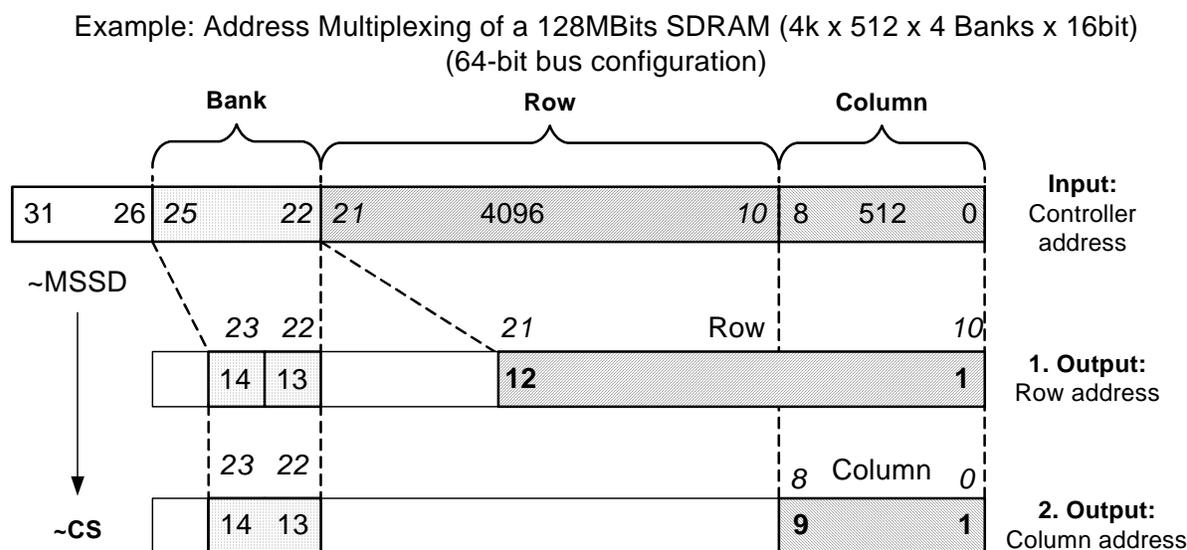


Figure 3 ADSP-TS101S SDRAM Controller Mapping Scheme Example

Note that only one bank at a time can be active, which results in some overhead cycles when switching between banks (off-bank accesses). Similarly, moving from one row to another (off-page access) results in the same overhead cycles. **Figure 4** shows how the ADSP-TS101S TigerSHARC on-chip SDRAM controller accesses SDRAM.

4M x 4bit x 4 Banks, 4096 Rows, Page size 1024 words

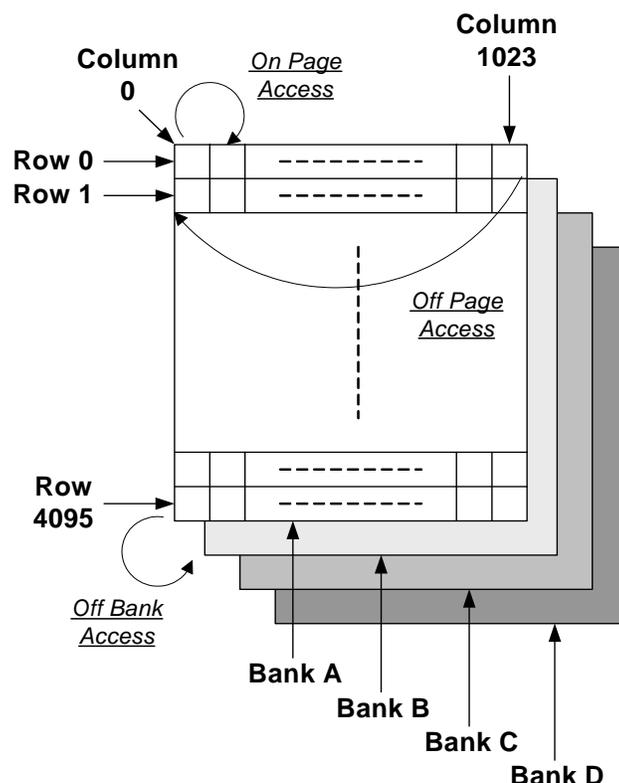


Figure 4 ADSP-TS101S SDRAM Controller Access Structure

6.2 TigerSHARC SDRAM Bank Select (~MSSD)

The fixed dedicated bank \sim MSSD ($A[31:26]=0x0b000001$) must be used for SDRAM accesses only. In this memory region, the controller's address multiplexer will be active.

6.3 Burst Stop (BST)

Although the controller works in burst mode, there is one way to interrupt the burst with the burst stop command. BST is issued if the next instruction is:

- Non external SDRAM access (access to another TigerSHARC bank)
- Core access (depending on the number of accesses, delay and external port FIFOs state)
- DMA operation (external port DMA to SDRAM interrupted by a higher priority DMA)
- Refresh counter expired (refresh period counter)
- SDRAM read to write and write to read transitions
- SDRAM off page/bank access
- \sim HBR asserted (host interface)

- During a Bus Transition Cycle (multiprocessing)

6.4 Data Mask Function ([H:L]DQM)

The [H:L]DQM pins are used by the controller to mask write operations. HDQM masks the SDRAM DQ buffers when performing 32-bit writes to even addresses in a 64-bit bus configuration. LDQM masks the SDRAM DQ buffers when performing writes to odd addresses in a 64-bit bus configuration. This data mask function does not apply for read operations, where the LDQM and HDQM pins are always low (inactive).

This is summarized in the following table:

Bus Width*	64-bit			32-bit
	32-bit Even	32-bit Odd	64-bit	32-bit Even/Odd
HDQM	1	0	0	x
LDQM	0	1	0	0

x = don't care 0 = logic 0, 1 = logic 1

*Bus Width bit setting in SYSCON

6.5 SDRAM Bank Select

The next tables show the address lines selection for the different banks:

2-banked access		
Banks	A[11:15]	SDA10
Bank_A	0	0
Bank_B	1	0
Banks A/B	x	1

x = don't care, 0 = logic 0, 1 = logic 1

Note: Any address line from address range A[11:15] can be used for bank select as long as they are not driven as a row or column address.

4-banked access			
Banks	A[11,13,15]	A[12,14]	SDA10
Bank_A	0	0	0
Bank_B	1	0	0
Bank_C	0	1	0
Bank_D	1	1	0
All Banks	X	X	1

x = don't care, 0 = logic 0, 1 = logic 1

Note: Any address line pair from address range A[11:15] can be used for bank select as long as they are not driven as a row- or column address.

6.6 Controller Address 10 (SDA10)

This pin provides a special solution to gain control of the SDRAM, even when the DSP operates as a slave (multiprocessing). The SDA10 pin allows access to all banks simultaneously during a refresh and precharge-all command. This pin must be connected to the A10 pin of the SDRAM.

Note that *the SDA10 pin replaces the DSP's A[10] and A[11] pins in a 32-bit and 64-bit bus width configuration, respectively*. Also, during access to the ~MSSD space, these pins are inactive.

6.7 Burst Mode

Although the SDRAM device is programmed for full page burst, the controller uses *quad-word* (128-bits) burst mode. For 32-bit bus width, the burst length is 4 words, and for 64-bit width, the burst length is 2 words. Only the first read or write command is accompanied with an external address, which is driven by the controller until the burst is interrupted by another address.

It's also important to note that *the SDRAM Controller burst mode cannot be changed*.

6.8 Precharge All (PREA)

This command precharges all SDRAM banks simultaneously (SDA10 high to select all banks), which brings the banks into idle state.

Although only one bank at a time can be active, *the controller does not support a single bank precharge*.

6.9 Circular Access

The controller supports circular accesses during sequential read or writes within a page, performing a fixed throughput of 1 cycle/word. At the end of the page (defined in the SDRCON register), the instructions `xR3:0=Q[j1+=last_word];;` followed by `xR7:4=Q[j1+=first_word];;` are also executed with a 1 cycle/word throughput.

This functionality is similar to the IALU's circular buffering mode supported by the TigerSHARC core.

6.10 Auto-Refresh (REF)

After the SDRAM registers the auto-refresh command, it internally asserts CAS and delays RAS to execute a row's refresh. The row interval is typically $t_{RC}=15,625 \mu s$, which is a good compromise between data access time and the refresh reliability. The limit of refresh period is given through the *tREFmax* spec.

Note that *the controller does not support burst refresh*.

6.11 Self-Refresh (SREF)

The self-refresh is a very effective way of reducing the application's power consumption to a minimum. When a host processor gains control of the cluster bus, the TigerSHARC SDRAM controller brings the SDRAM into self-refresh mode before the bus is relinquished to the host. The SDRAM starts refreshing itself triggered by an internal timer. The controller does not allow bringing the SDRAM into self-refresh mode by software, only during host accesses.

6.12 Mode Register Set (MRS)

During the MRS command, the SDRAM controller initializes the SDRAM with the following *fixed* settings:

- Burst length is hardwired to full page burst
- Burst type is hardwired to sequential burst
- Read Latency (CL) is user specified (1-3 cycles)

7 SDRAM Programming

Before external bus transactions to SDRAM start, the system and SDRAM control registers must be configured accordingly.

7.1 SYSCON Register

The SYSCON register is the system configuration register and must be configured after hardware reset at the beginning of the source code. This register may only be configured once, and additional write to SYSCON after the first are ignored.

This register is composed of different fields, although only the following applies for SDRAM:

Bus Width: For proper operation, ensure that the dedicated bus width bits settings is:

- 0: 32-bit bus
- 1: 64-bit bus

Note that if either the host or memory bus width is 64-bits, the multiprocessing width must also be 64-bits. Also, using 64-bit mode, the DSP address 0 pin becomes redundant, since its information is contained in the stobes. Therefore, this pin is not connected for SDRAM accesses in 64-bit mode (see Figure 1).

The following diagram shows the ADSP-TS101S external port data alignment for both, 32 and 64-bit bus configurations:

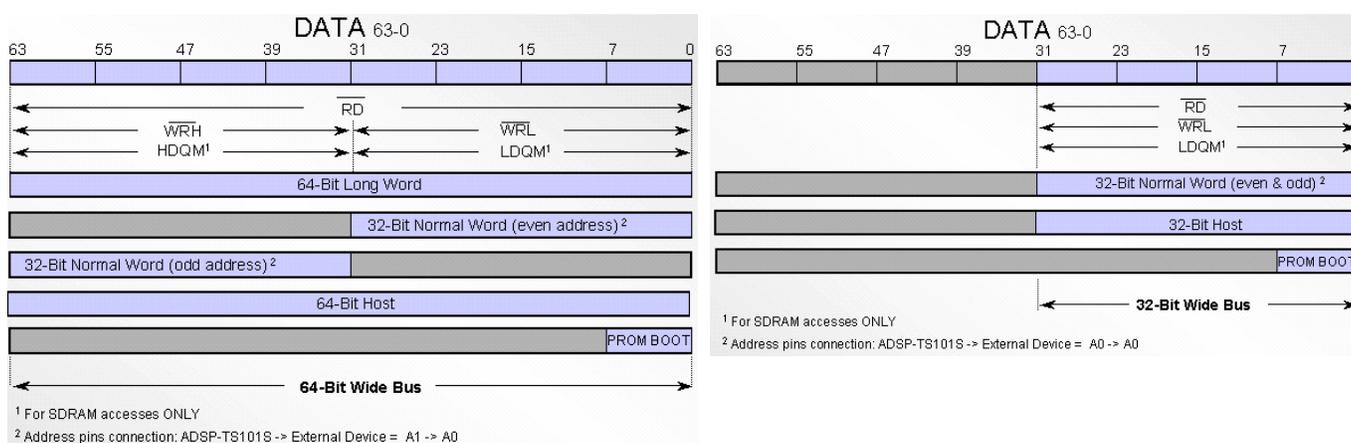


Figure 5 ADSP-TS101S External Port Data Alignment

7.2 SDRCON Register

The SDRAM programming is done by the SDRCON register. Similar to SYSCON, it can only be programmed once after reset and its value should remain unchanged during normal operation. In systems where more than one DSP is used, this register must be programmed to the same setup in every processor.

The SDRCON bit fields are:

SDRAM Enable: set whenever an SDRAM is present in the system.

CAS Latency: this bit is used to define the read latency (CL_{min}) related to the vendor's device. This value can be set from 1 to 3 cycles, depending on the SDRAM characteristics and the clock frequency.

Pipe depth: this bit allows the SDRAM address and control lines to be pipelined. Setting this bit introduces a one-cycle delay during read and write accesses.

Page Boundary: this bit determines the page size of the SDRAM, which can be set to 256, 512, or 1024 words.

Refresh Counters: these bits enable coordination of the clock rate with the SDRAM's required refresh rate. It is used to select between 4 different refresh rates calculated with the following equation:

$$Cycles = \left(Clock \cdot \frac{t_{REF}}{Rows} \right) = (Clock \cdot refresh_rate)$$

The ADSP-TS101S supports refresh rates of 600, 900, 1200 or 2400 cycles. The following table illustrates how to select between the different rates based on the formula above.

Clock frequency (MHz)	Refresh rate (cycles)	SDRCON Bits [8:7]
$f < 60$	600	00
$60 \leq f < 80$	900	01
$80 \leq f < 150^*$	1200	10
$f \geq 150^*$	2400	11

* Note that the maximum external port frequency supported by the ADSP-TS101S is 100 MHz.

Precharge to RAS delay (tRP): this bit defines the precharge time (t_{RPmin}) related to the vendor's device. This value can be set from 2 to 5 cycles. Note that this parameter is defined in most SDRAMs in nanoseconds delay rather than number of cycles.

RAS to precharge delay (tRAS): this bit is used to define the row active time (t_{RASmin}) related to the vendor's device. This value can be programmed from 2 to 8 cycles. Note that this parameter is also defined in nanoseconds rather than number of cycles.

Initialization Sequence: this bit determines the order of the MRS (Mode Register set) and refresh sequences. When set, MRS follows refresh in the SDRAM initialization sequence. Otherwise, the MRS precedes refresh:

Init Sequence = 0

- PREA command - brings the SDRAM in the defined idle state.
- 8 REF commands - charges SDRAM's internal nodes.
- MRS command - initializes the SDRAM's working mode.

Init Sequence = 1

- PREA command - brings the SDRAM in the defined idle state.
- MRS command - initializes the SDRAM's working mode.

- 8 REF commands - charges SDRAM's internal nodes.

$tRAS$ and tRP are used for the refresh cycle, which can be expressed as: $tRFC=tRAS+tRP$. Also, the specification requires that $tRAS_{min}$, tRP_{min} , and CL_{min} are defined as a integer fraction of the $SCLK$ period.

7.3 SDRAM Mode of Operation

During the MRS, address bits ADDR[13:0] of the SDRAM are used to program the device. The MRS is executed during power-up only and it is always issued by the TigerSHARC with ID=000. MRS initializes the following parameters:

- A[2:0] - Burst length is hardwired to full page burst.
- A[3] - Burst type is hardwired to sequential burst.
- A[6:4] - Latency mode set according to the CAS latency programmed in SDRCON (1-3 cycles).
- A[13:7]- hardwired to zero (reserved mode of operation for future needs).

7.4 ADSP-TS101S EZ-KIT Lite

The ADSP-TS101S TigerSHARC Processor EZ-KIT Lite™ includes an SDRAM PC-DIMM module, which is either 128-Mbyte or 32-Mbyte, depending on the EZ-KIT revision used. The following table shows the different SDRAM control register settings depending on the module used:

EZ-KIT Rev.	SDRAM Module	SDRAM CLK	SDRAM Settings	SDRCON
1.2	128 Mbyte	83.3 MHz	SDRAM Enable = 1	0x5323
			CAS Latency = 2	
			Pipe Depth = 0	
			Page size = 1k	
			Refresh rate = 1200	
			tRP = 3, tRAS = 4	
			Init Sequence = 1	
1.3	32 Mbyte	83.3 MHz	SDRAM Enable = 1	0x5303
			CAS Latency = 2	
			Pipe Depth = 0	
			Page Size = 256	
			Refresh rate = 1200	
			tRP = 3, tRAS = 4	
			Init Sequence = 1	

7.5 SDRAM Setting Overview

Timing spec	Controller	Description
<i>tCK</i>	50–100 MHz	clock cycle time, hardware
<i>tREFmax</i>	600–2400 cycles	refresh period
<i>CLmin</i>	1-3 cycles	read latency
<i>tRASmin</i>	2-8 cycles	activate to precharge
<i>tRPmin</i>	2-5 cycles	precharge period
<i>tRCD</i>	$tRCD = CL$ (fixed)	RAS to CAS delay
<i>tRFCmin</i>	$tRFC = tRAS + tRP$ (fixed)	Auto-refresh period
<i>tMRD</i>	2 cycles (fixed)	mode register to command
<i>tXSR</i>	$tRFC + tRP$ (fixed)	Exit self refresh to active

SDRAM (Mode Register Set)	
<i>Power up Mode</i>	PREA-MRS-REF or PREA-REF-MRS
<i>Burst Mode</i>	sequential (fixed)
<i>Burst Length</i>	full page (fixed)
<i>CL</i>	1-3 cycles

8 Power-up Sequence

The power-up procedure for the SDRAM interface is shown in **figure 6**:

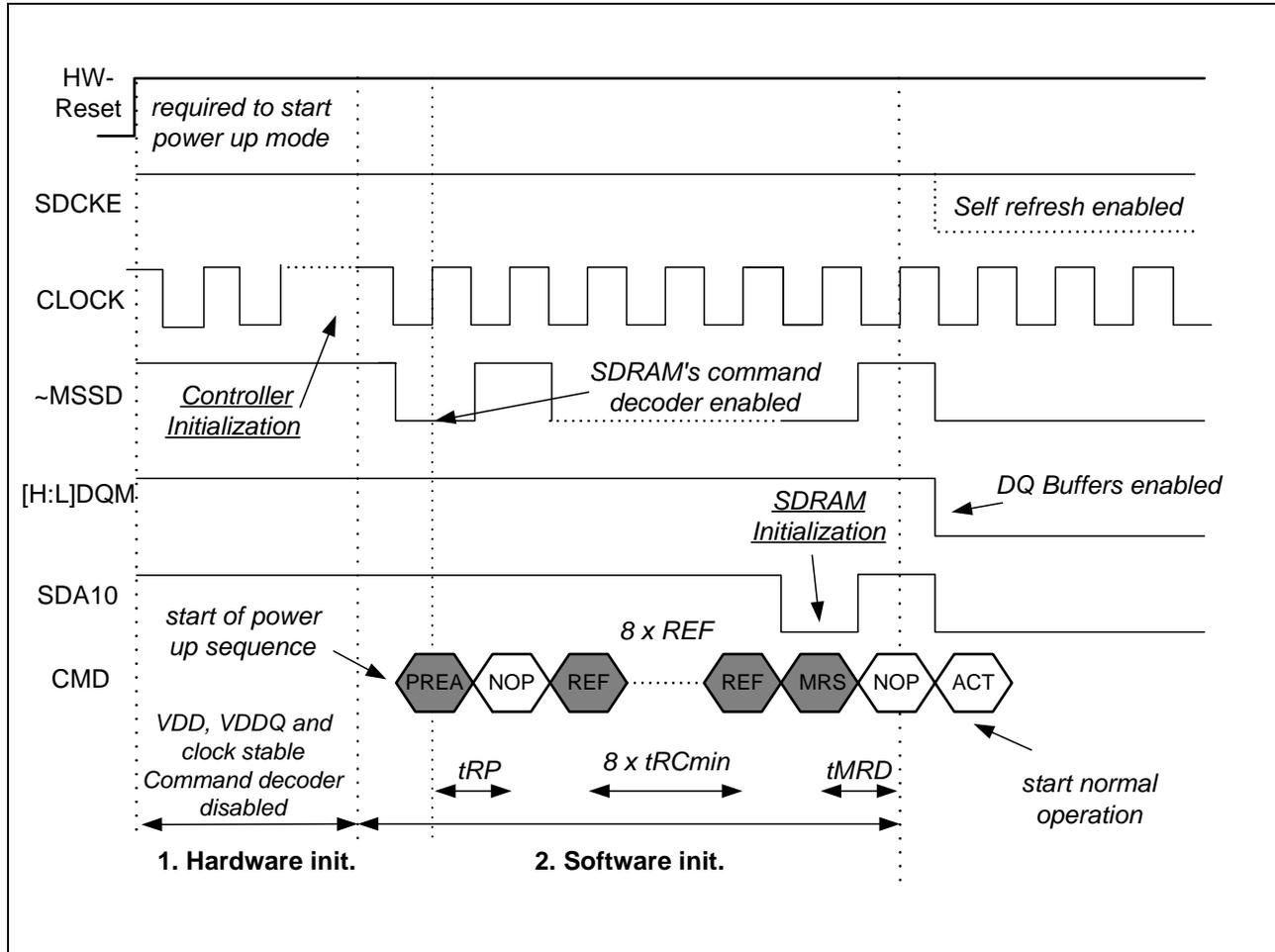


Figure 6 Power-up and Initialization: PREA-REF-MRS

8.1 Hardware Initialization

After hardware reset of the ADSP-TS101S, the clock and the SDRAM's power supply pins (VDD and VDDQ) must provide a stable signal for a typical minimum time of 200 μ s. The SDRCON register can be accessed only after this time has elapsed.

8.2 Software Initialization

The SDRAM controller is initialized by writing to the SDRCON register. As soon as ~MSSD is asserted, the SDRAM's command decoder is enabled.

The *Init Sequence* bit in SDRCON starts the initialization sequence (refer to section 7.2 *SDRCON Register*). During this time, the [H:L]DQM lines stay high, three-stating the SDRAM DQ buffers. Once

the initialization sequence is completed, the buffers are enabled and the SDRAM is ready for normal operation.

The time elapsed before the first access to SDRAM can be represented as:

$$t_{access} \approx t_{RP} + 8(t_{RAS} + t_{RP}) + t_{MRD} \text{ (SCLK cycles)}$$

Note that in order to properly initialize the SDRAM, the first access is delayed with the internal acknowledge until the power up sequence has finished.

8.3 SDRAM Initialization Example

This section shows how to initialize the SDRCON register according to the following device specifications.

- 2M x 32 bit SDRAM:
- 4 banks, page size: 256 words
- SDRAM Clock: 60 MHz
- Speed grade: 70
- Refresh cycles: 4096/64ms
- Power-up mode: PRE - MRS - REF
- CLmin=2 @60MHz, tRASmin=42 ns, tRPmin=20 ns
- No self refresh
- No buffering

Therefore, the SDRAM initialization code would look as follows:

```
j1 = j31 + 0x00004883;; // ENA=1, CL=2, pipedepth=0, page=256w
SDRCON = j1;; // refresh rate=900, trp=2, tras=3, init=1
```

Note that the minimum timing specifications for tRAS, tRP, and CL must be guaranteed. Setting these values larger than the minimum required causes a loss of performance. (e.g. longer delay cycles or unnecessary refresh cycles).

8.4 SDRAM Interface After Reset

After power up and when the ADSP-TS101S reset pin is deasserted, the SDRAM-lines are in the following state:

Pin	State	Description
SDCKE	1	SDRAM Clock enabled
~MSSD	1	command decoder disabled
~RAS	1	Deselected
~CAS	1	Deselected
~SDWE	1	Deselected
[H:L]DQM	1	SDRAM data buffers disabled
SDA10	1	Access all banks simultaneously

9 DMA Transfers

Direct Memory Access (DMA) is a mechanism for transferring data without core involvement. Transfers can be between internal and external memory, or between external memory and an external peripheral.

9.1 Internal Memory and SDRAM

The TigerSHARC DMA controller can be used to transfer data from the DSP's internal memory to external SDRAM. Similarly, the SDRAM can be used as the source and the internal memory as the destination.

9.2 External Device and SDRAM (FLY-BY)

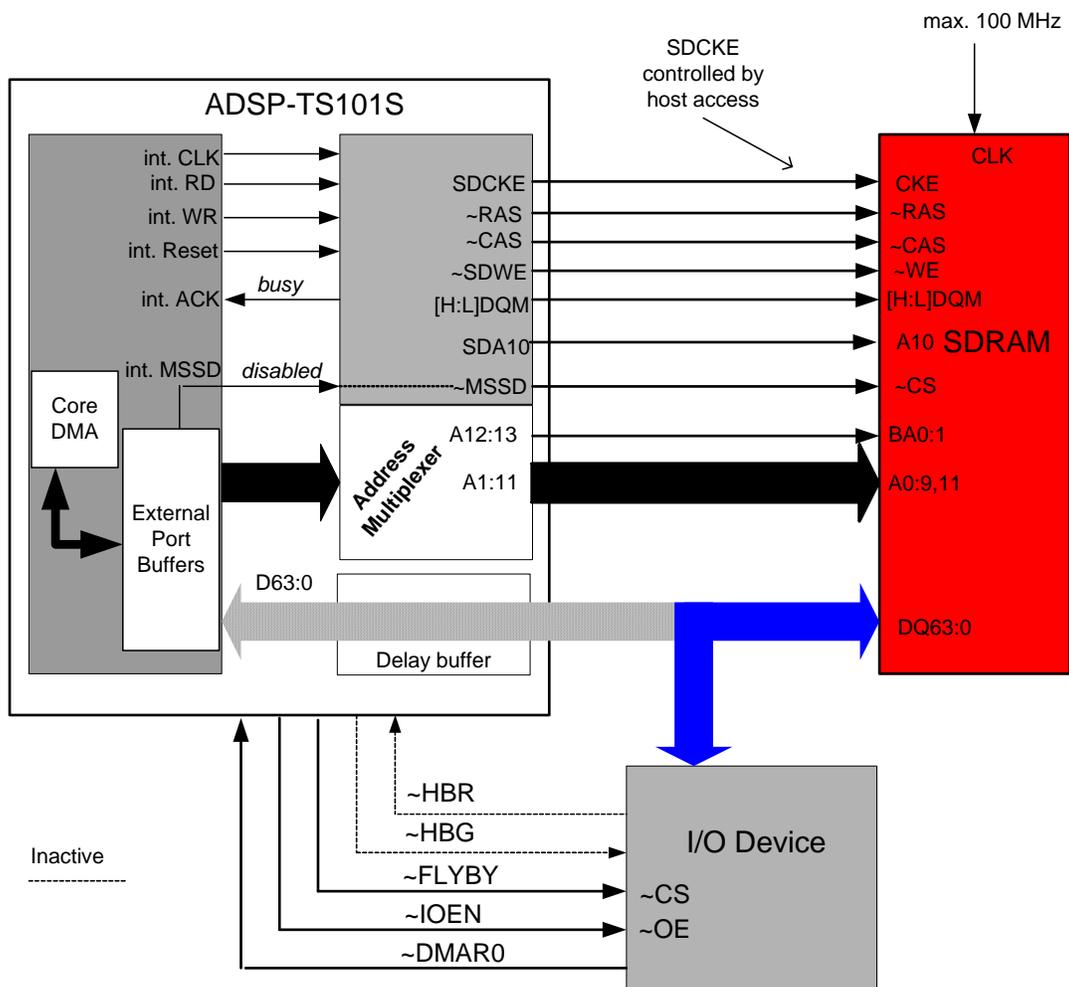


Figure 7 Signal chain: Fly-by DMA and SDRAM

Data transfers between external SDRAM and external I/O device is also supported. In this mode, there is no core intervention, and only the TigerSHARC DMA controller is used to perform the correspondent data transfer between the two external devices. This is called *fly-by mode*.

When the I/O-device wants to transfer data, it asserts the \sim DMAR0 pin – only DMA channel 0 can be used for fly-by transactions. Note that the \sim DMAR0 is *edge sensitive*. Each assertion (falling edge) of the \sim DMAR0 signal represents one transaction. The \sim FLBY is used as a chip select for the I/O device and \sim IOEN allows read transaction (output enable) from the I/O.

For more details on the TigerSHARC DMA controller and Fly-by Mode, refer to the “*Direct Memory Access*” and the “*Cluster Bus*” chapters of the *ADSPTS101 TigerSHARC Processor Hardware Reference Manual* [1]. Also, “*Understanding DMA on the ADSP-TS101S (EE-143)*” [4] is available.

10 SDRAM Interface in Host Mode

Figure 8 shows a possible signal chain between the ADSP-TS101S, SDRAM, and a host processor.

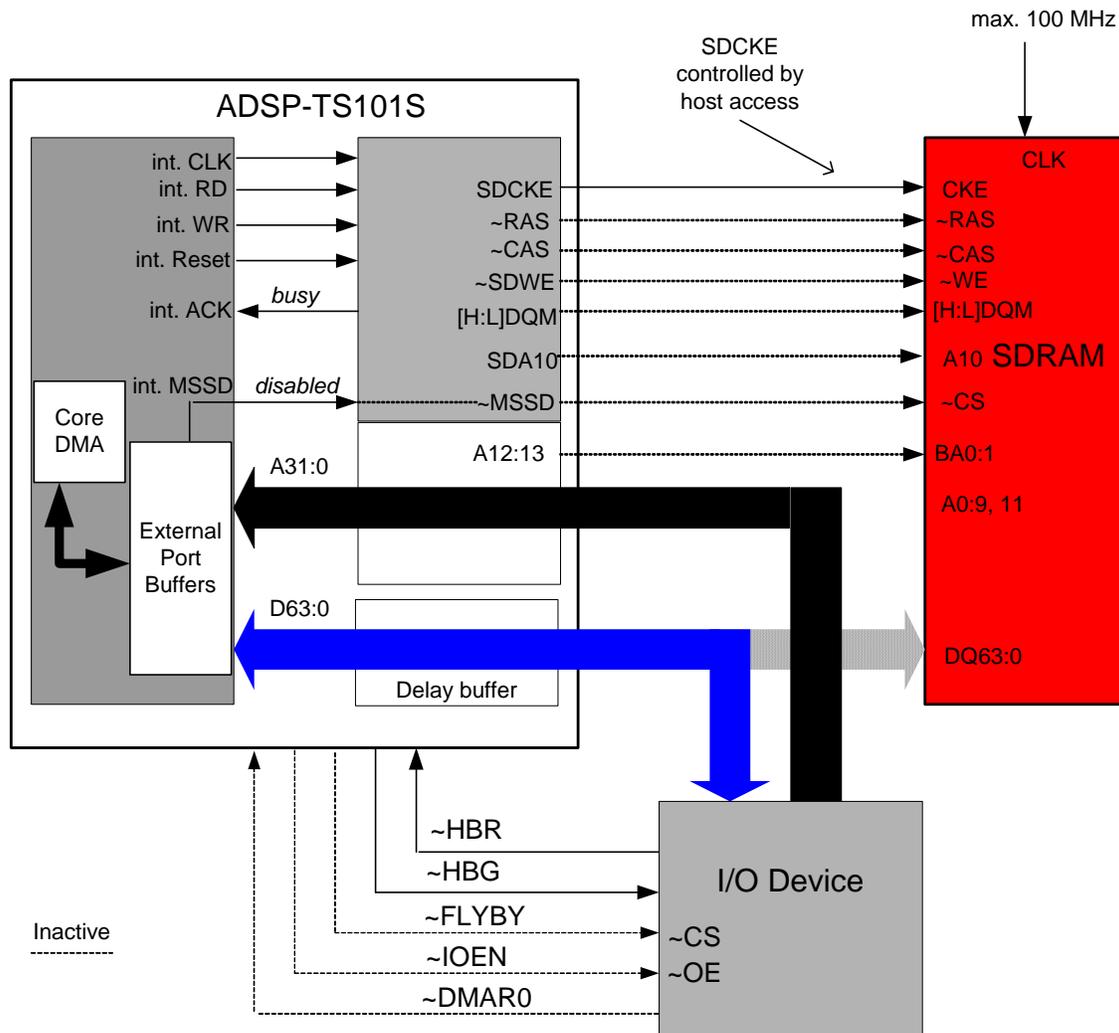


Figure 8 Signal Chain: Host to ADSP-TS101S

Host accesses are initiated by having the host assert \sim HBR to request the external bus to the TigerSHARC. Thereafter, the TigerSHARC responds to this request by asserting \sim HBG, granting the bus to the host. The host becomes bus master and can now access the internal TigerSHARC resources through its multiprocessor memory space.

Note that *during \sim HBG low, the SDRAM is in self-refresh mode. This limits the host device to not being able to access the SDRAM using the TigerSHARC's SDRAM controller (fly-by mode). Thus, during host transactions, the SDRAM can only be accessed by a host using its own SDRAM controller.*

However, a workaround for this may be possible if the host follows the given sequence:

1. Initially, the TigerSHARC is current bus master. To gain control of the bus, the host asserts $\sim HBR$ and waits for the TigerSHARC to grant the bus ($\sim HBG$)
2. The host becomes bus master and is now able to access the internal resources of the TigerSHARC.
3. The host sets up a DMA transfer to have the TigerSHARC transfer data to SDRAM upon host requests (DMA request line $\sim DMAR0$).
4. The host relinquishes the bus (de-asserts $\sim HBR$), which brings the SDRAM out of self-refresh mode
5. Finally, the host starts toggling $\sim DMAR0$, indicating to the TigerSHARC to start transmitting data to the SDRAM.

11 Multiprocessing

This section covers the arbitration logic used to guarantee multiprocessing systems.

11.1 Command Decoding

The ADSP-TS101S can be connected to a multiprocessor cluster of eight DSPs. Only one TigerSHARC can drive the bus at the time. To build a glueless hardware, the interface works in slave mode as well to detect commands. These commands are MRS, REF, and SREF. Following pins are necessary for detection:

Pin	Type	Description
SDCKE	I/O/T	SDRAM Clock Enable
~MSSD	I/O/T	Memory Select SDRAM (command input)
~RAS	I/O/T	Row Address Select (command input)
~CAS	I/O/T	Column Address Select (command input)
~SDWE	I/O/T	SDRAM Write Enable (command input)

I=input, O=output, T=three-state

11.2 MRS Decoding

In TigerSHARC multiprocessor systems where SDRAM is used, *the Mode Register Set sequence is only issued by the DSP with ID=000. Therefore, a DSP with ID=000 must be present in every multiprocessor system.* The MRS sequence is only issued once before accessing the SDRAM for the first time. It is also important to note that the set-up value initialized in the SDRCON register must be identical for all DSPs in the cluster.

11.3 REF Decoding

This detection helps to synchronize all eight refresh counters. The slave's refresh counter will be decremented each time the interface detects a refresh. This feature guarantees a periodic refresh and, similarly to the MRS sequence, it requires the exact same settings of SDRCON registers.

11.4 SREF Decoding

This detection helps to synchronize the self-refresh base. When a host requests the bus, the system master brings the SDRAM into self-refresh mode. The slave DSPs recognize the SREF freezing the refresh counters until the self-refresh mode is exited.

11.5 Bus Transition Cycle

The bus transition or turn-over cycle is issued when bus mastership changes between the different TigerSHARCs sharing the cluster bus. This also implies a mastership transition between the SDRAM controllers. Since the new bus master is not able to determine which row in the SDRAM is currently open, the current bus master automatically issues a precharge-all command (PREA) before the bus is

relinquished to the new bus master. This way, the new bus master can safely start accessing the SDRAM by simply issuing an activation command (ACT).

This sequence is shown in **Figure 9**:

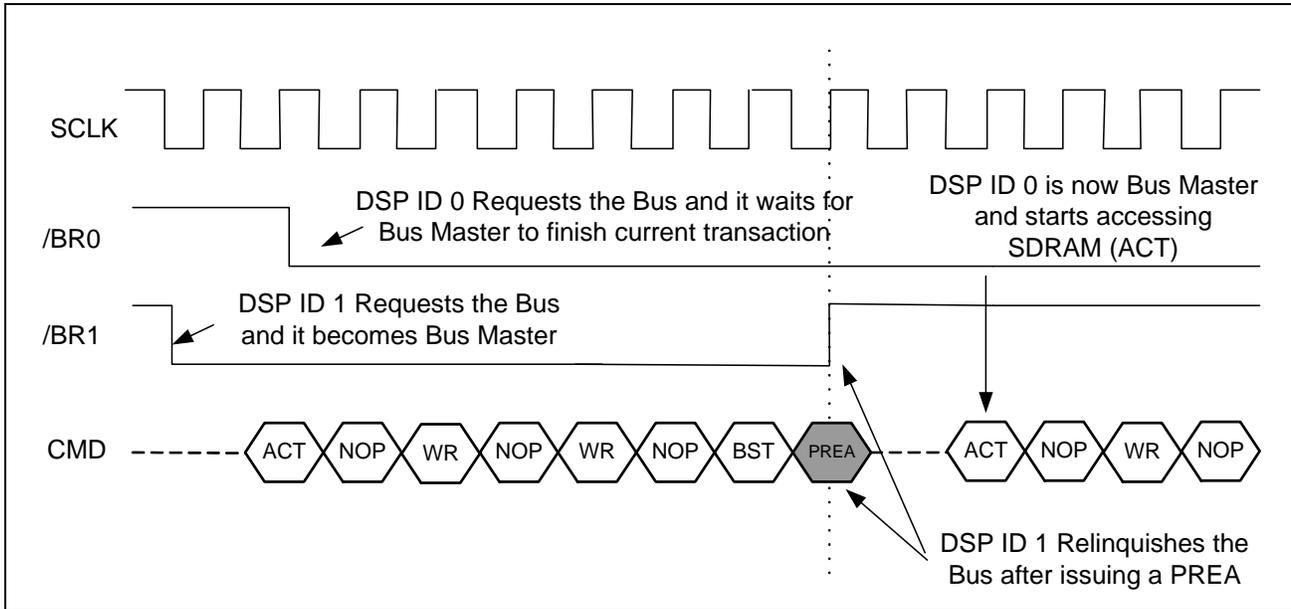


Figure 9 Bus Transition Cycle during SDRAM Accesses

12 SDRAM and Booting

12.1 Loader Kernel

Data must be loaded in SDRAM before runtime. This requires an initialized SDRAM before downloading the data during the boot scenario. The VisualDSP++® loader utility provides this capability.

Three standard loader executables are available for booting either an EPROM (via link ports) or a host processor. *Note that independently from the selected boot mode, the SDRAM control register (SDRCON) must be initialized before the user's application starts writing data to it. This must be done during the first 256-loader kernel words; otherwise the data placed in the SDRAM will get corrupted.*

The steps are:

1. ~RESET must be deasserted, ~BMS is sampled
2. Kernel (256 x 32 bit) is transmitted via DMA into the DSP (0x00-0xFF) (~BMS or ~HBG continuously asserted)
3. Interrupt generation starts kernel execution, SDRAM and controller is initialized (~MSSD asserted for SDRAM setup)
4. User's data is loaded into the DSP/SDRAM (~BMS, ~HBG and ~MSSD are toggling requesting the bus)
5. Kernel is now overwritten with user's code (~BMS or ~HBG continuously asserted)
6. DSP starts code execution at start address 0x00000000

12.2 Booting Modes

The boot mode is selected at the end of reset by sampling the ~BMS pin. At this stage, four options for beginning operation can be selected:

Mode	~BMS	DMA channel
EPROM	0 (Default)	Hardwired DMA Channel 0
Link Port	1	Link Port DMA Channels 8-11
Host	1	AutoDMA Channels 0-1
No boot	1	-

If ~BMS is sampled low during reset, EPROM boot is selected. However, if ~BMS is sampled high during reset, the DSP will go into an IDLE state, waiting for host or link boot. Refer to “ADSP-TS101S TigerSHARC Processor Boot Loader Kernels Operation (EE-174)” [5], which describes each mode of operation in more detail.

13 SDRAM Interface Throughput

This section investigates the SDRAM controller's performance during data transfers, where the SDRAM is used as either a source or destination. The following settings will be used:

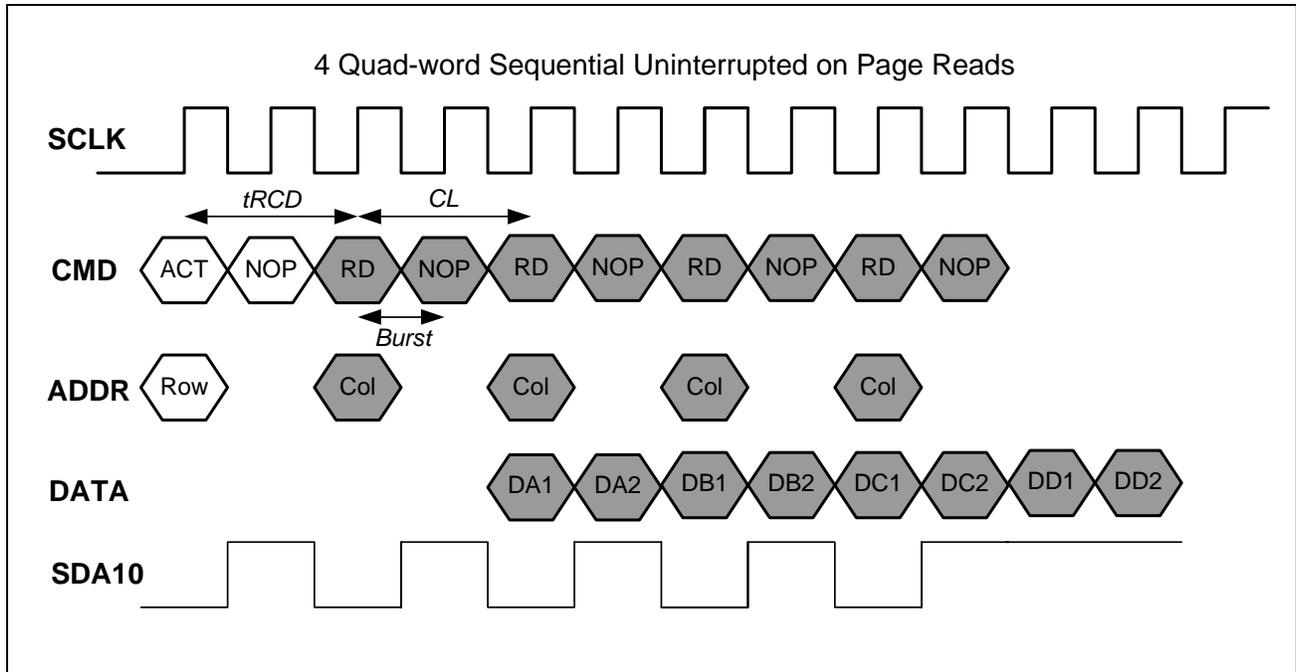
- Silicon = ADSP-TS101S Rev0.0
- SCLK = 60 MHz (external port and SDRAM clock)
- LCLK = 180 MHz (core clock)
- SYSCON = 0x3A79E7, for 64-bit bus configuration
- SDRCON = 0x4883, where Page Boundary = 256 words,
Refresh rate = 900 cycles,
 t_{RAS} =3 cycles,
 t_{RP} =2 cycle,
CL=2 cycles,

The following diagrams correspond to the different types of core/IOP accesses to SDRAM. Additionally, a listing of the instructions and SDRAM commands executed during each memory access are listed.

Note that the external bus is configured to 64-bits wide. Therefore, every data word read/written from/to the SDRAM will be 64-bits (i.e. DA1, DA2, DB1, etc.).

13.1 Sequential Reads without Interruption

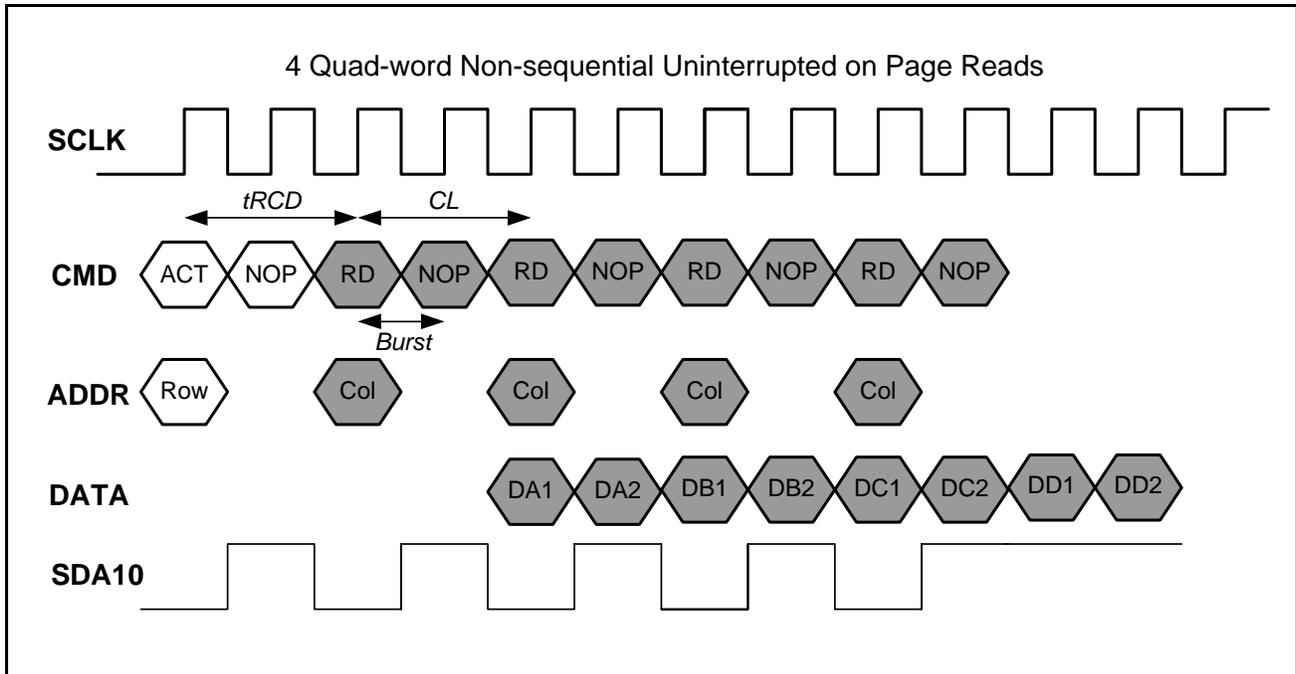
As shown below, uninterrupted sequential reads have a throughput of 1 word per cycle after the first word is read.



Nr. Cycles	Core	Controller	Data
1	xr3:0=q[j1+=j4];;	ACT	
2	int. ACK	NOP	
3	int. ACK	RD	
4	int. ACK	NOP	
5	xr7:4=q[j1+=j4];;	RD	DA1
6	int. ACK	NOP	DA2
7	xr11:8=q[j1+=j4];;	RD	DB1
8	int. ACK	NOP	DB2
9	xr15:12=q[j1+=j4];;	RD	DC1
10	int. ACK	NOP	DC2
11	int. ACK	...	DD1
12	int. ACK	...	DD2

13.2 Non Sequential Reads without Interruption

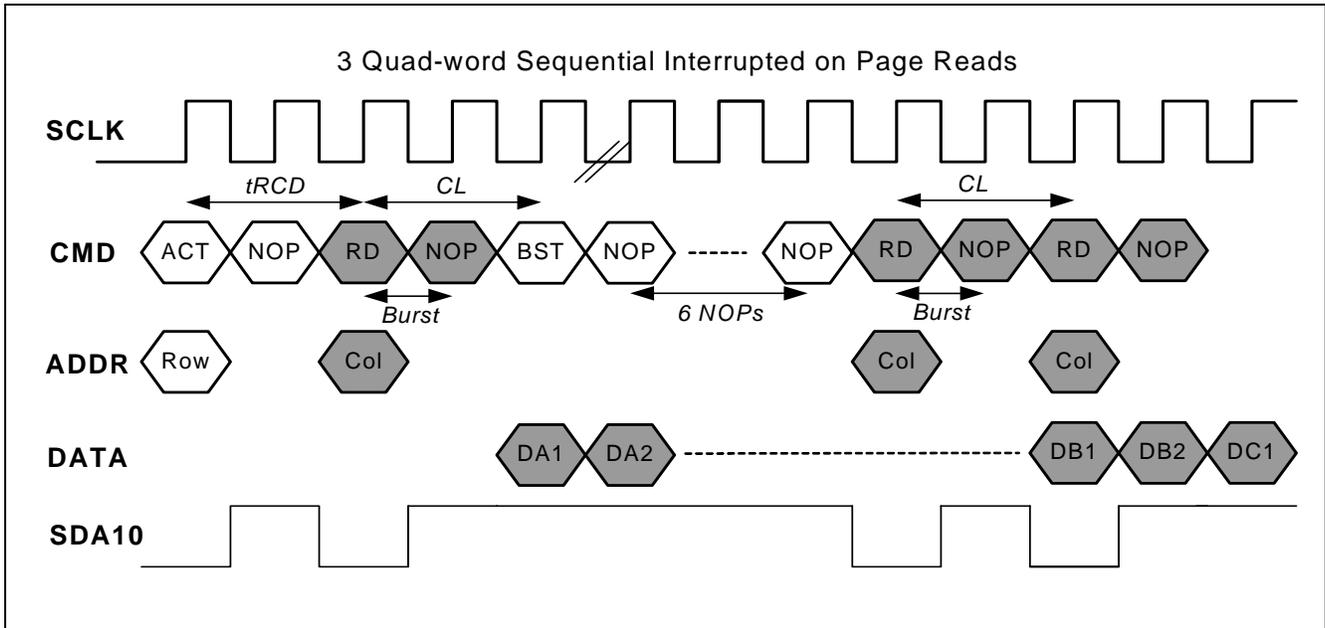
Similar to uninterrupted sequential reads, non-sequential reads also give a 1 word/cycle throughput.



Nr. Cycles	Core	Controller	Data
1	xr3:0=q[j1+=j4];;	ACT	
2	int. ACK	NOP	
3	int. ACK	RD	
4	int. ACK	NOP	
5	xr7:4=q[j1+=j4];;	RD	DA1
6	int. ACK	NOP	DA2
7	xr11:8=q[j1+=j4];;	RD	DB1
8	int. ACK	NOP	DB2
9	xr15:12=q[j1+=j4];;	RD	DC1
10	int. ACK	NOP	DC2
11	int. ACK	...	DD1
12	int. ACK	...	DD2

13.3 Sequential Reads with minimum Interruption

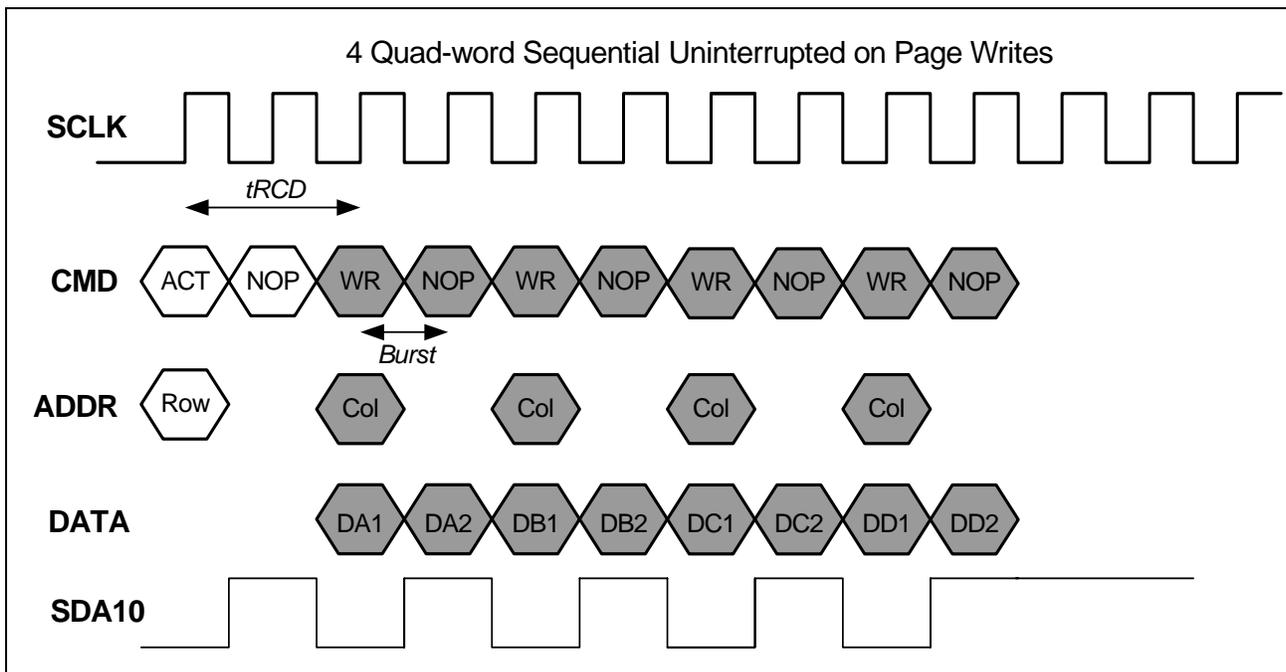
In this case, an additional core access is performed during a core read from SDRAM, interrupting the data transfer. Similarly, an IOP read is interrupted by a higher priority request to IOP. In this case, the throughput is 1 word every 8 cycles.



Nr. Cycles	Core	Controller	Data
1	xr3:0=q[j1+=j4];;	ACT	
2	int. ACK	NOP	
3	int. ACK	RD	
4	int. ACK	NOP	
5	xr0=r1+r2;;	BST	DA1
6	int. ACK	NOP	DA2
7	int. ACK	NOP	
8	int. ACK	NOP	
9	int. ACK	NOP	
10	int. ACK	NOP	
11	int. ACK	NOP	
12	xr7:4=q[j1+=j4];;	RD	
13	int. ACK	NOP	
14	xr15:12=q[j1+=j4];;	RD	DB1
15	int. ACK	NOP	DB2
16	int. ACK	NOP	DC1
17	int. ACK	NOP	DC2

13.4 Sequential Writes without Interruption

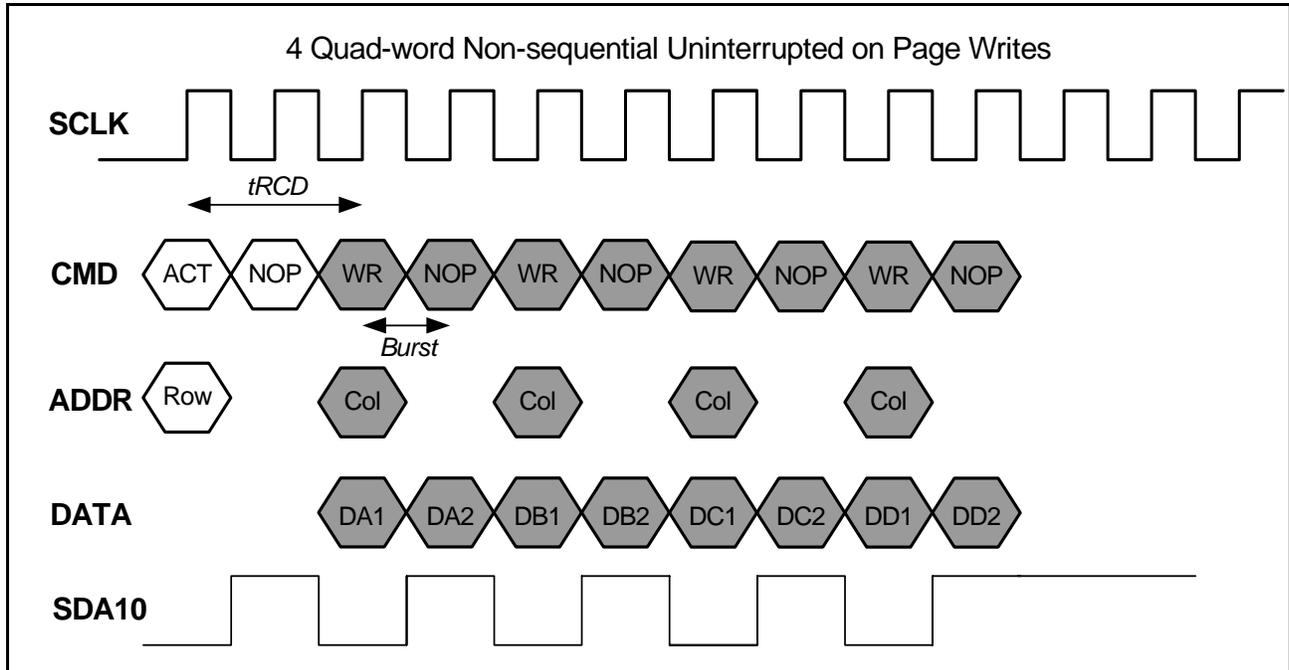
As shown below, uninterrupted sequential writes have a throughput of 1 word per cycle.



Nr. Cycles	Core	Controller	Data
1	<code>q[j1+=j4]=xr3:0;;</code>	ACT	
2	int. ACK	NOP	
3	int. ACK	WR	DA1
4	int. ACK	NOP	DA2
5	<code>q[j1+=j4]=xr7:4;;</code>	WR	DB1
6	int. ACK	NOP	DB2
7	<code>q[j1+=j4]=xr11:8;;</code>	WR	DC1
8	int. ACK	NOP	DC2
9	<code>q[j1+=j4]=xr15:12;;</code>	WR	DD1
10	int. ACK	NOP	DD2

13.5 Non Sequential Writes without Interruption

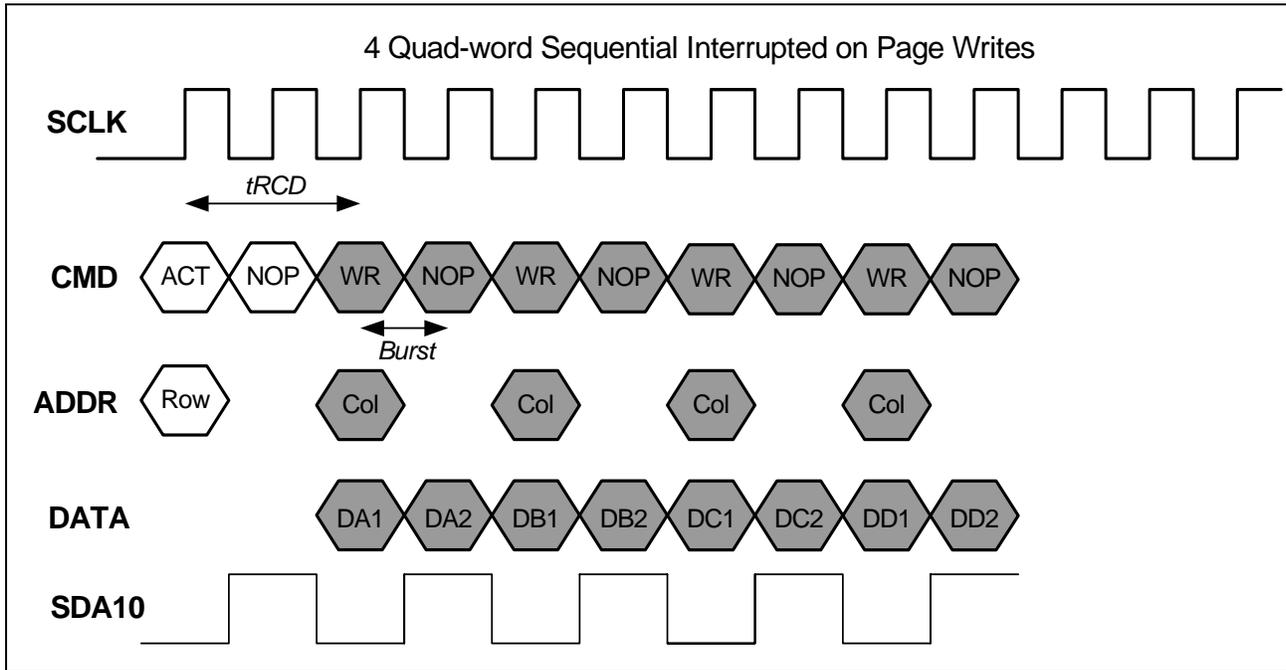
Similar to uninterrupted sequential writes, non-sequential writes also give a 1 word/cycle throughput.



Nr. Cycles	Core	Controller	Data
1	q[j1+=j4]=xr3:0;;	ACT	
2	int. ACK	NOP	
3	int. ACK	WR	DA1
4	int. ACK	NOP	DA2
5	q[j1+=j4]=xr7:4;;	WR	DB1
6	int. ACK	NOP	DB2
7	q[j1+=j4]=xr11:8;;	WR	DC1
8	int. ACK	NOP	DC2
9	q[j1+=j4]=xr15:12;;	WR	DD1
10	int. ACK	NOP	DD2

13.6 Sequential Writes with minimum Interruption

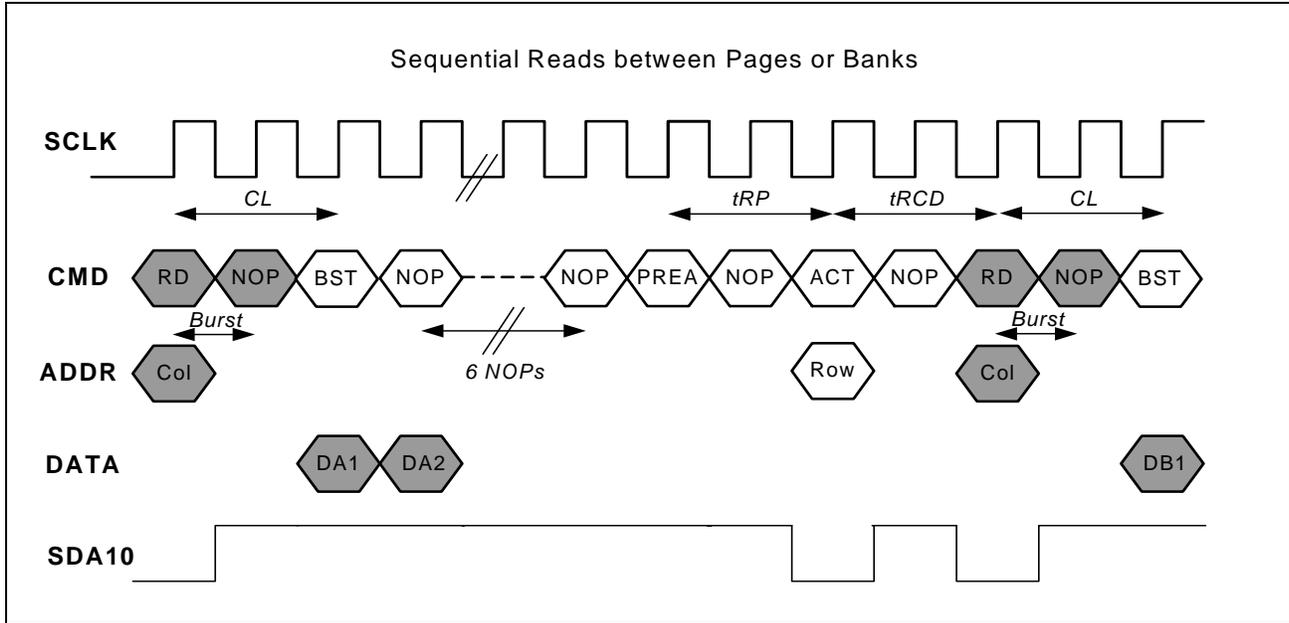
In this case, an additional core access is performed during a core/IOP write to SDRAM. As shown below, the SDRAM access does not get interrupted, and it operates as for un-interrupted writes, 1 word/cycle. Note that a higher number of core accesses during a write to SDRAM may cause a throughput decrease depending upon system speed and core activity.



Nr. Cycles	Core	Controller	Data
1	<code>q[j1+=j4]=xr3:0;;</code>	ACT	
2	int. ACK	NOP	
3	int. ACK	WR	DA1
4	int. ACK	NOP	DA2
5	<code>q[j1+=j4]=xr7:4;;</code>	WR	DB1
6	<code>xr0 = r1 + r2;;</code>	NOP	DB2
7	<code>q[j1+=j4]=xr11:8;;</code>	WR	DC1
8	int. ACK	NOP	DC2
9	<code>q[j1+=j4]=xr15:12;;</code>	WR	DD1
10	int. ACK	NOP	DD2

13.7 Reads between Page/Bank

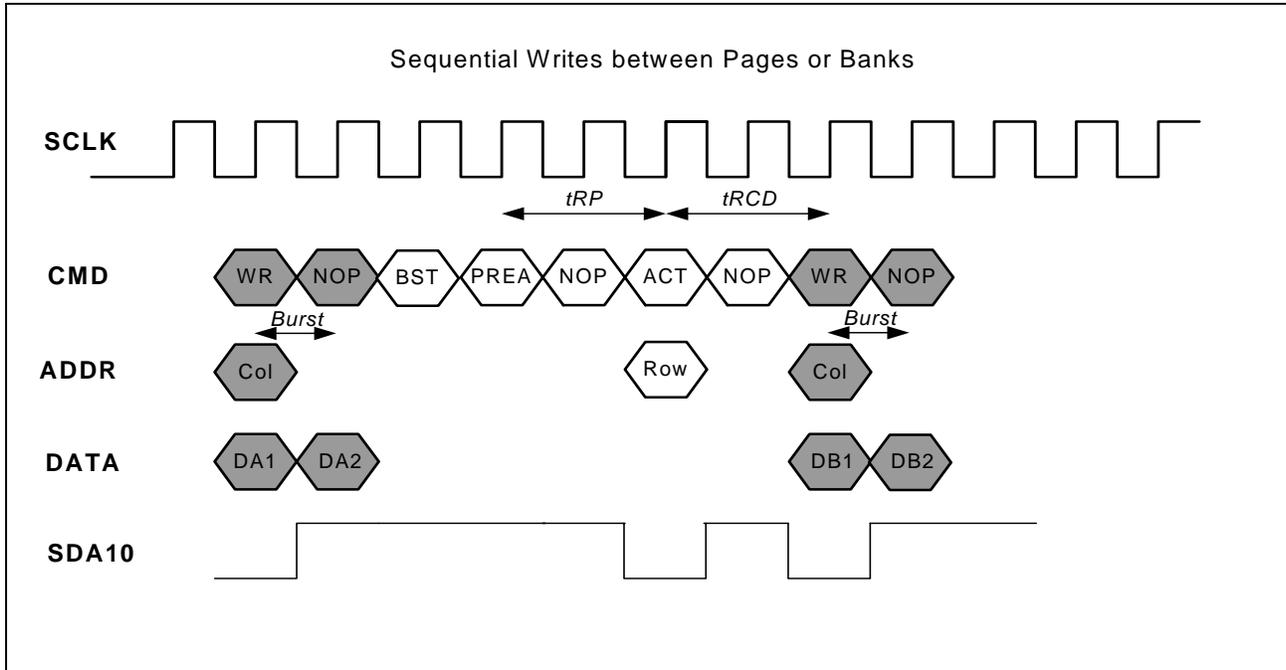
Core/IOP reads from SDRAM interrupted by off-bank/page accesses. Crossing bank/page boundaries have a great impact in performance, giving a 1 word per 12 cycles throughput.



Nr. Cycles	Core	Controller	Data
1	xr3:0=q[j1+=260];;	ACT	
2	int. ACK	NOP	
3	int. ACK	RD	
4	int. ACK	NOP	
5	int. ACK	BST	DA1
6	int. ACK	NOP	DA2
7	int. ACK	NOP	
8	int. ACK	NOP	
9	int. ACK	NOP	
10	int. ACK	NOP	
11	int. ACK	NOP	
12	int. ACK	PREA	
13	int. ACK	NOP	
14	xr7:4=q[j1+=j4];;	ACT	
15	int. ACK	NOP	
16	int. ACK	RD	
17	int. ACK;	NOP	
18	int. ACK;	BST	DB1

13.8 Writes between Page/Bank

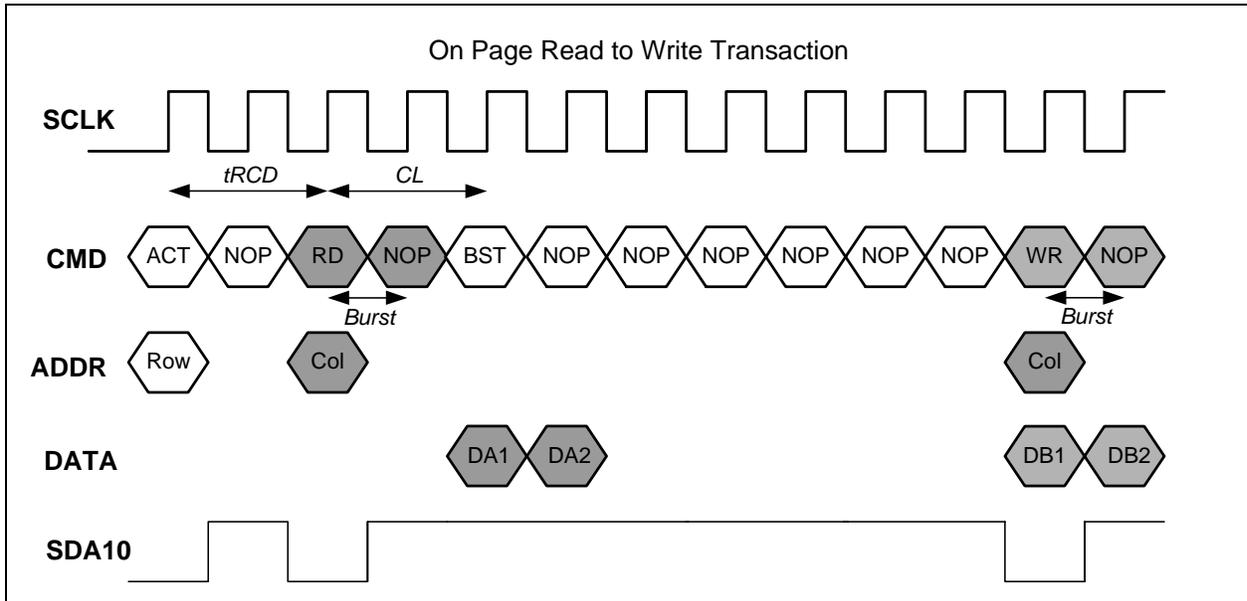
Core/IOP writes to SDRAM interrupted by off-bank/page accesses. Like for read accesses, although smaller, crossing bank/page boundaries have a great impact in performance, giving a 1 word per 6 cycles throughput.



Nr. Cycles	Core	Controller	Data
1	<code>q[j1+=260]=xr3:0;;</code>	ACT	
2	int. ACK	NOP	
3	int. ACK	WR	DA1
4	int. ACK	NOP	DA2
5	int. ACK	BST	
6	int. ACK	PREA	
7	int. ACK	NOP	
8	<code>q[j1+=j4]=xr7:4;;</code>	ACT	
9	int. ACK	NOP	
10	int. ACK	WR	DB1
11	int. ACK	NOP	DB2

13.9 Minimum Read to Write Interval

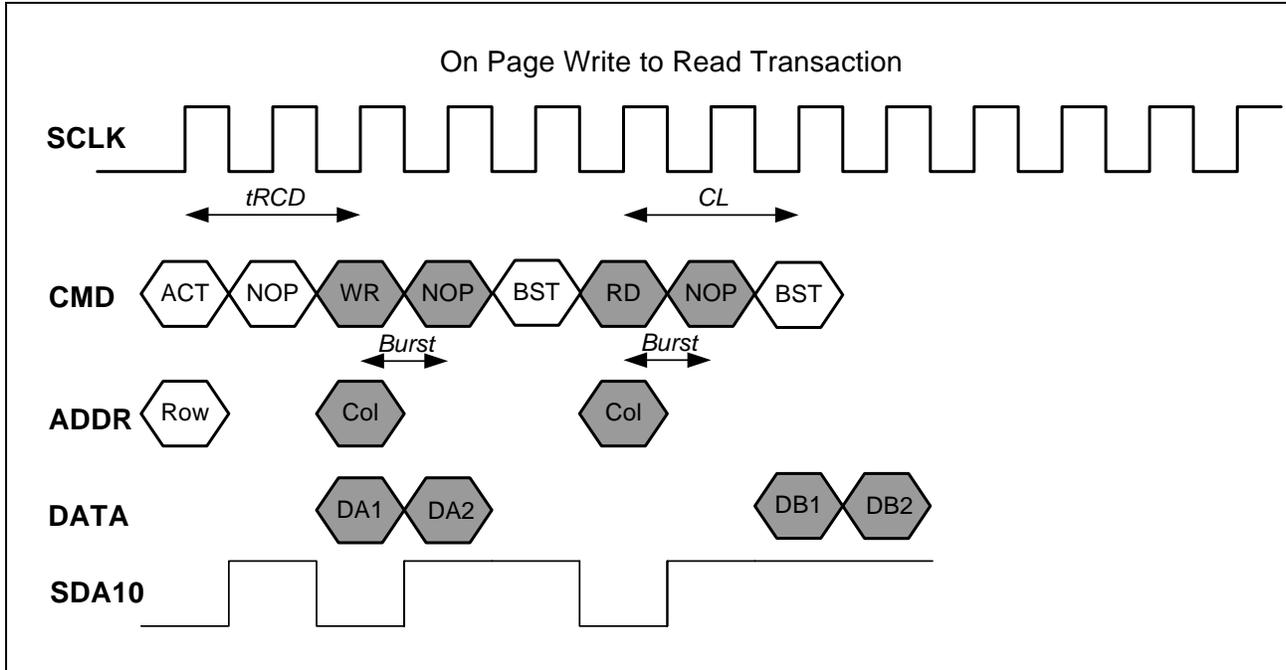
The following diagram shows an SDRAM read to write transaction. As shown below, the transition from one access to another introduces some cycles of overhead, resulting in a loss of performance, 1 word per 6 cycles.



Nr. Cycles	Core	Controller	Data
1	<code>xr3:0=q[j1+=j4];;</code>	ACT	
2	int. ACK	NOP	
3	int. ACK	RD	
4	int. ACK	NOP	
5	int. ACK	BST	DA1
6	int. ACK	NOP	DA2
7	int. ACK	NOP	
8	int. ACK	NOP	
9	int. ACK	NOP	
10	int. ACK	NOP	
11	int. ACK	NOP	
12	<code>q[j1+=j4]=xr7:4;;</code>	WR	DB1
13	int. ACK	NOP	DB2

13.10 Minimum Write to Read Interval

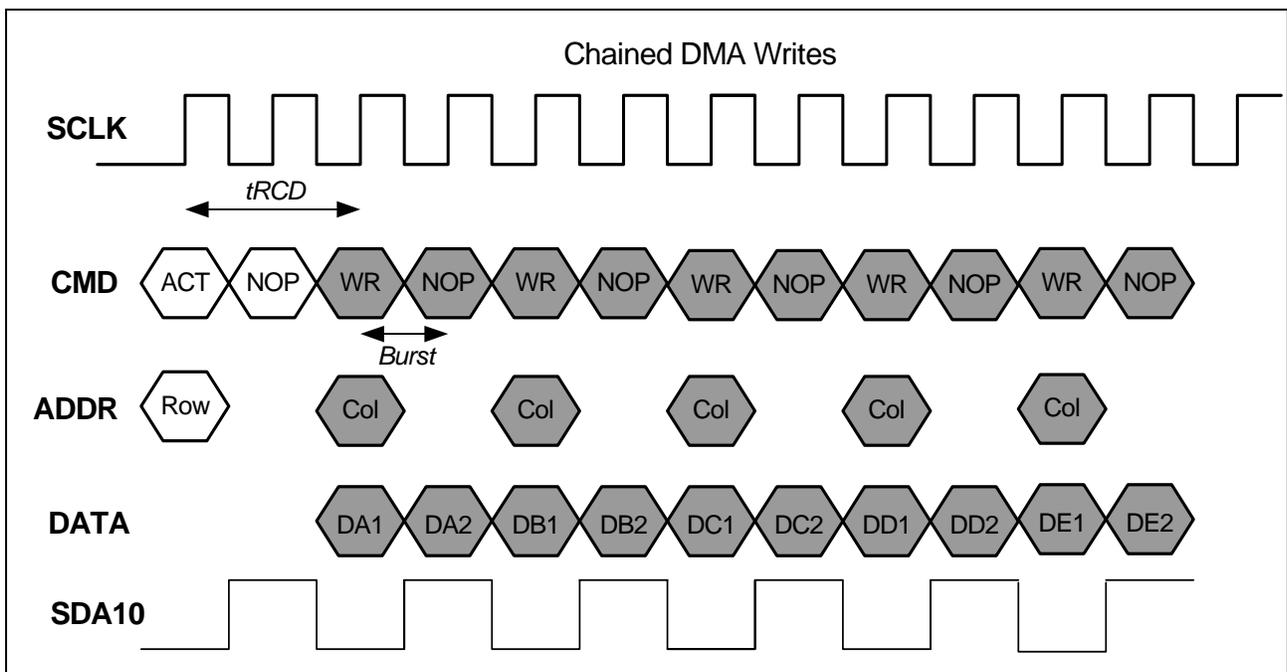
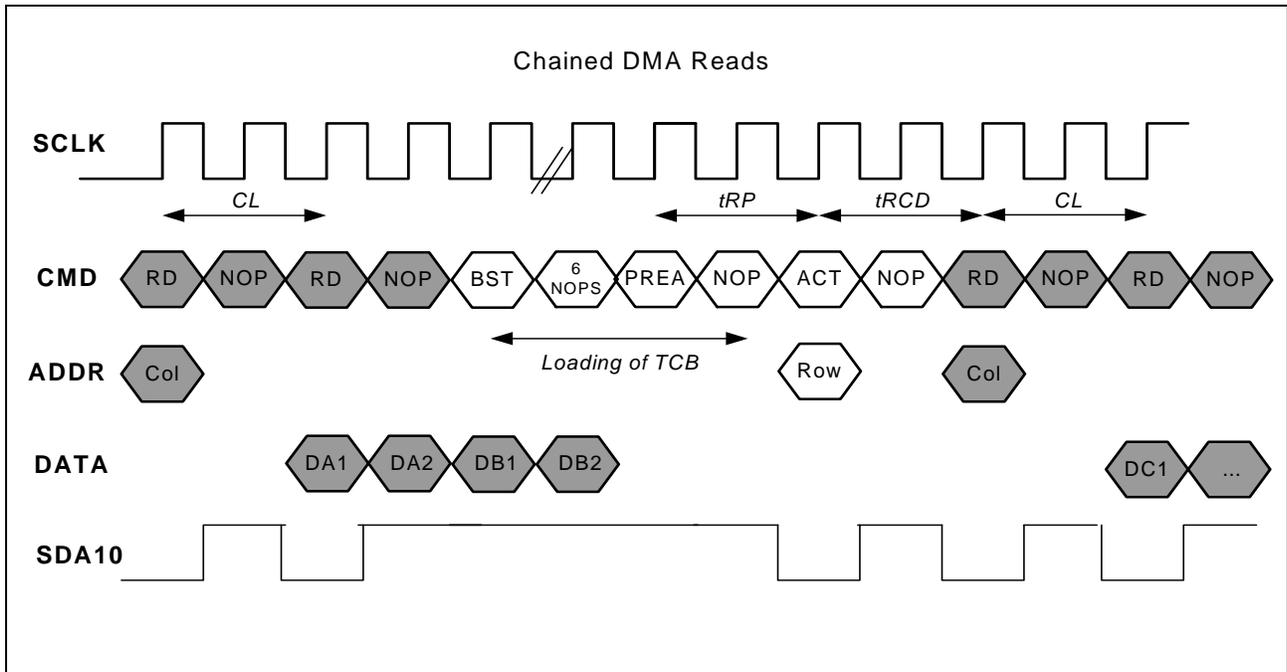
The following diagram shows an SDRAM write to read transaction. As for read to write transitions, some cycles of overhead are introduced, resulting in a loss of performance, 1 word per 4 cycles.



Nr. Cycles	Core	Controller	Data
1	<code>q[j1+=j4]=xr3:0;;</code>	ACT	
2	int. ACK	NOP	
3	int. ACK	WR	DA1
4	int. ACK	NOP	DA2
5	int. ACK	BST	
6	<code>xr7:4=q[j1+=j4];;</code>	RD	
7	int. ACK	NOP	
8	int. ACK	BST	DB1
9	int. ACK	NOP	DB2

13.11 Chained DMA Transfers

When using chained DMAs, higher throughput can be achieved. However, as shown below, a performance difference can be seen between chained DMA writes and reads to and from SDRAM. Loading of the TCB for chained read DMAs requires additional core cycles.

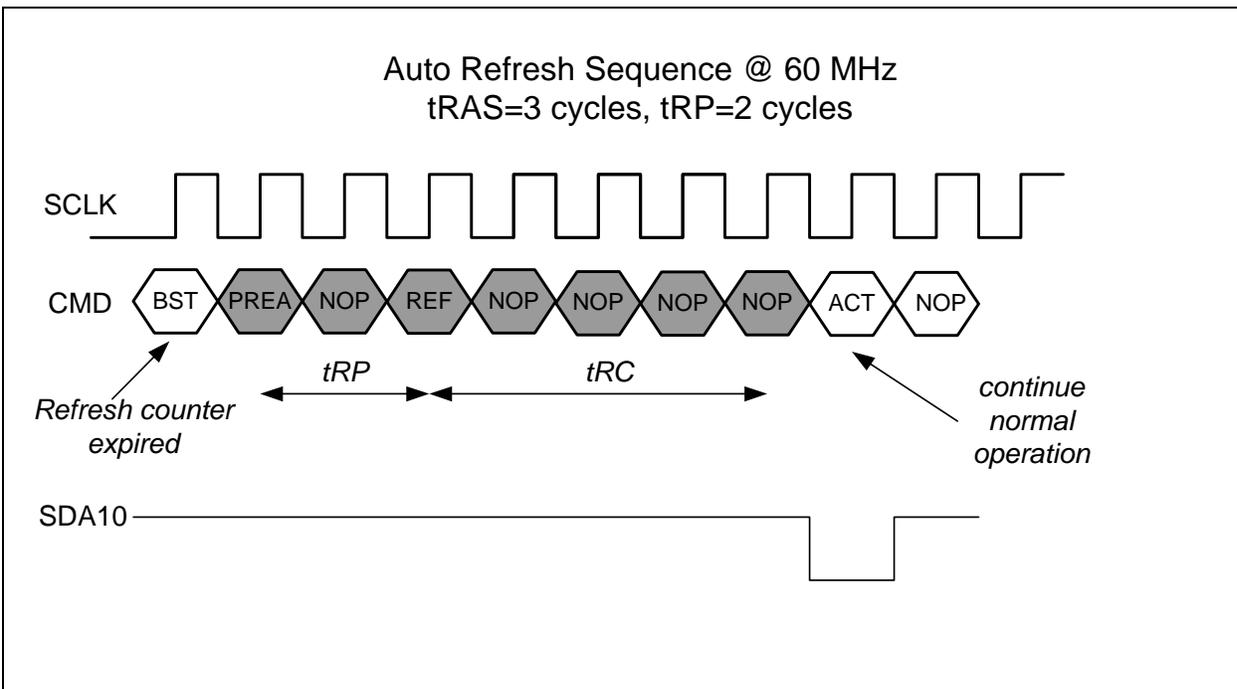


13.12 Auto-Refresh

The refresh counter triggers refresh requests when expired. The following diagram shows the refresh sequence. All banks are precharged (SDA10 asserted high) right after the internal counter issues the refresh. A REF command starts an internal row refresh with CAS before RAS. During refresh, no other commands can be executed. The ratio between application time and refresh time is given by:

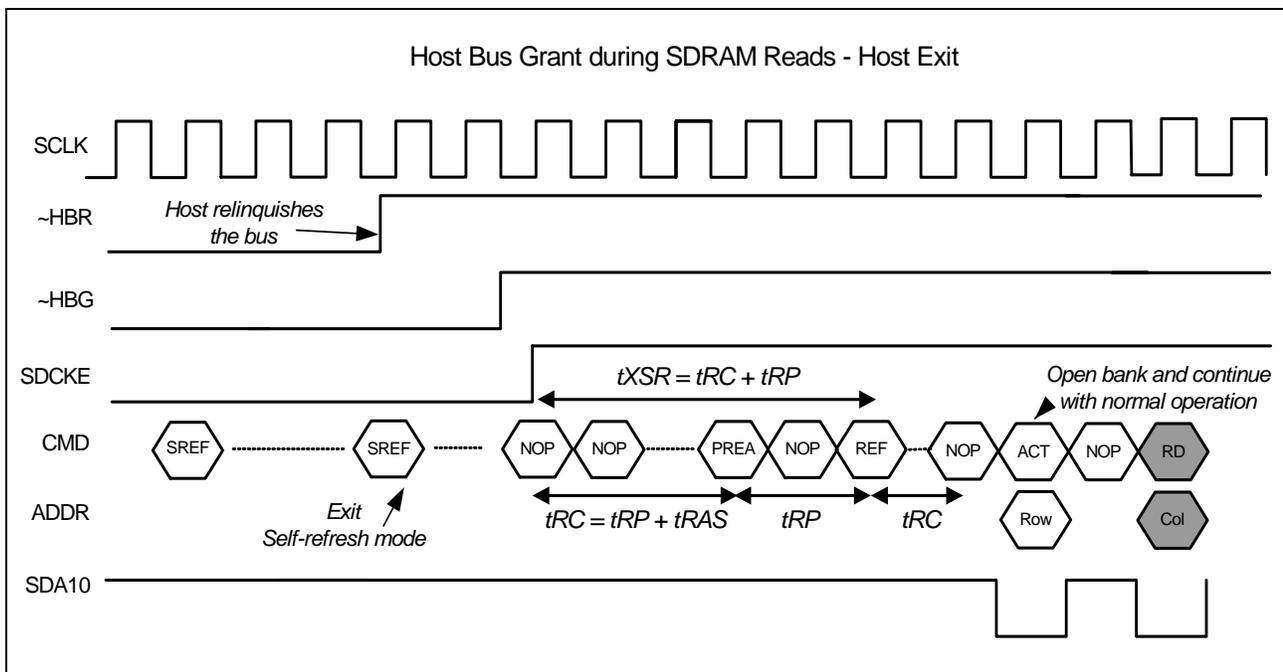
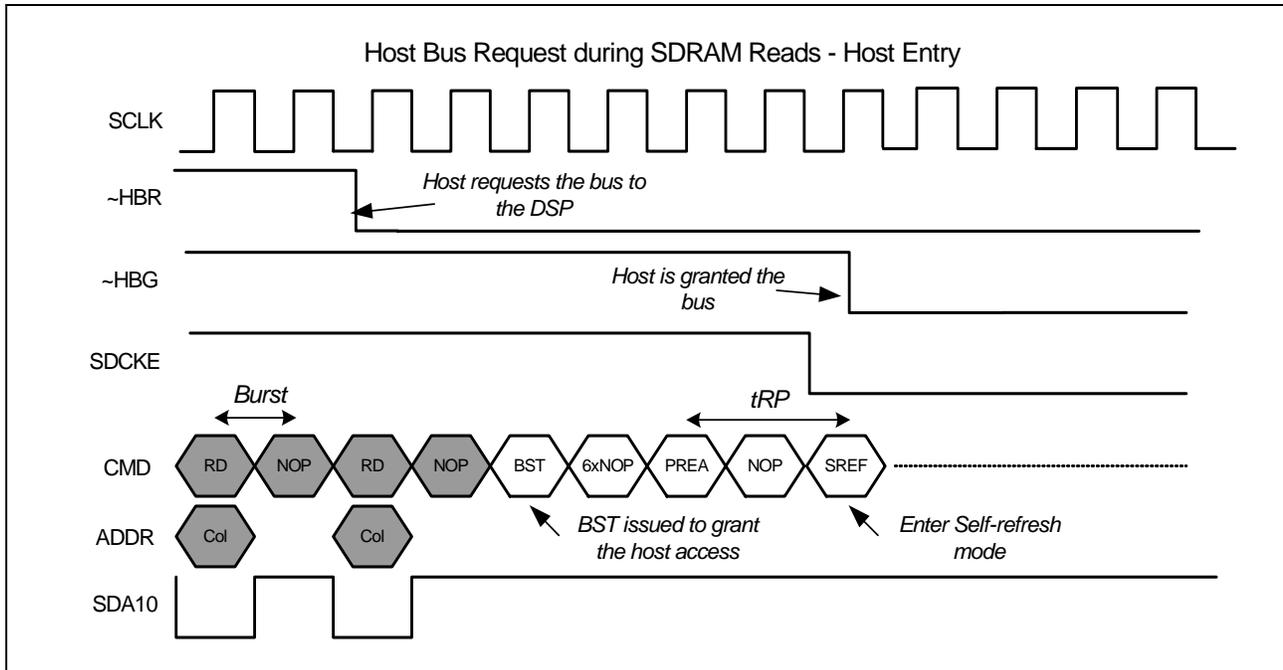
$$t_{RC} = 15,625 \mu\text{s}/\text{rows} \quad \text{e.g. } 8 \text{ cycles}/900 \text{ cycles} \times 100 = 0.88\%$$

This means that the refresh sequence requires 0.88 % of the whole performance at 60 MHz. Auto refresh with 15.625 $\mu\text{s}/\text{row}$ gives the best performance ratio. This ratio can be modified accordingly to the system needs by changing the refresh rate programmed in SDRCON.



13.13 Self-Refresh and Host Accesses

The following diagrams show a random host access during a read from SDRAM:



After the detection of a host request, the current burst operation of the SDRAM controller is interrupted and frozen. The TigerSHARC DSP enters self-refresh mode, in response to host bus request (\sim HBR), before the bus is relinquished to the host. As shown above, the controller issues a BST command,

stopping the current action. In order to provide current information, all banks are precharged. During the PREA and the SREF command, the SDCKE pin transitions from high to low, thus enabling the self-refresh mode.

After deassertion of \sim HBG, the controller ends the self refresh function and the SDCKE line transits high. Following, the controller issues a refresh sequence. The bank is then activated and it continues with normal operation, reading data from where it left off.

Note that *only the SDCKE pin keeps control of the device in self-refresh mode.*

13.14 SDRAM Performance Table

The following table gives a summary of all the different SDRAM accesses previously explored.

Accesses	Operations	Page	Throughput per SDRAM Clock (64-bit words) ¹
Sequential Uninterrupted	Read	Same	1 word/1 cycle
Non-sequential Uninterrupted	Read	Same	1 word/1 cycle
Sequential Interrupted	Read	Same	1 word/ 8 cycles (6 + CL)
Sequential Uninterrupted	Write	Same	1 word/1 cycle
Non-sequential Uninterrupted	Write	Same	1 word/1 cycle
Sequential Interrupted	Write	Same	1 word/1 cycle
Both	Read to Write	Same	1 word/6 cycles
Both	Write to Read	Same	1 word/4 cycles (2 + CL)
Nonsequential	Reads	Different	1 word/12 cycles (6+ t _{RP} + t _{RCD} + CL)
Nonsequential	Writes	Different	1 word/6 cycles (2+ t _{RP} + t _{RCD})
Autorefresh before read	Reads	Different	1 word/17 cycle (6+ 2t _{RP} + t _{RAS} + t _{RCD} + CL)
Autorefresh before write	Writes	Different	1 word/11 cycle (2+ 2t _{RP} + t _{RAS} + t _{RCD})

CAS latency (CL) = 2 cycles, Precharge (t_{RP}) = 2 cycles, t_{RAS} = 3 cycles, t_{RCD} = CL = 2 cycles

¹ SYSCON External bus width configuration: 64-bit bus.

14 Optimizing SDRAM Performance

There are several ways in which the SDRAM performance can be optimized depending on the system set up and application. Some of these are explained below:

14.1 External Buffering

In parallel connections, one address and control bus feeds all the system devices. In order to meet the timing requirements and help against capacitive load, the TigerSHARC controller provides a feature to cope with this situation.

The controller has the capability of pipelining the SDRAM accesses by making use of the internal delay buffer. In SDRCON, setting the pipe depth bit to 1 inserts this delay buffer, allowing address and command pipelining. It is recommended to set this bit if the capacitive load of 30 pF/pin is exceeded. The data sheet's timing is based on a nominal load of 30 pF/pin.

Figure 10 shows a possible system set-up where an external buffer is used in conjunction with the delay buffer provided by the TigerSHARC controller. The external buffer reduces capacitive load and power dissipation but increases the pipeline effects.

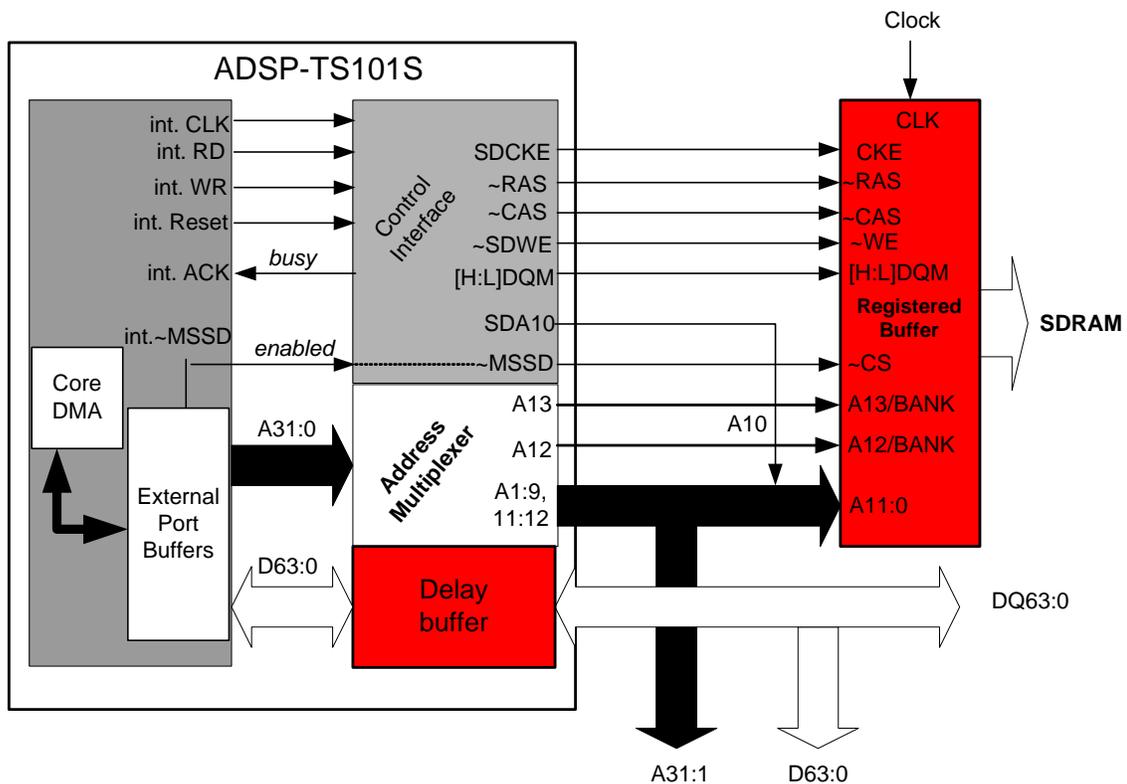


Figure 10 Signal chain: ADSP-TS101S to SDRAM Using External Buffer

14.2 Using PC Modules

The maximum addressable size is 64Mx32-bit or 32Mx64-bit. The use of unbuffered PC DIMM modules (typical I/O sizes x32, x64, or x72 bits) is efficient, because most vendors offer a size of x64 bits, which is commonly used for PCs. Moreover, depending on the size, the need for external address- and control buffers (pipelining) is required.

14.3 General Rules for Optimized Performance

Depending on the SDRAM's page size and number of banks, place your data segments to minimize off page and off bank accesses.

Use parallel connection for SDRAMs with big page size in order to obtain 32-bit or 64-bit I/O (SDRAM: the bigger the page size, the smaller the I/O structure).

Use the optimized settings for the controller's state machine (*tRAS*, *tRP*, *CL*) depending on the speed grade and on the application's speed

15 References

- [1] *ADSP-TS101S TigerSHARC Processor Hardware Reference Manual. Revision 1.0, April 2003. Analog Devices Inc.*
- [2] *ADSP-TS101S Data Sheet. Rev. A, February 2003. Analog Devices Inc*
- [3] *The ABC of SDRAMemory (EE-126). Rev 1, March 2002. Analog Devices Inc.*
- [4] *Understanding DMA on the ADSP-TS101S TigerSHARC (EE-143). October 2001. Analog Devices Inc.*
- [5] *ADSP-TS101S TigerSHARC Processor Boot Loader Kernels Operation (EE-174). April 2003. Analog Devices Inc.*

16 Document History

Revision	Description
<i>Rev 2 – December 19, 2003 by Maikel Kokaly-Bannourah and Robert Hoffmann</i>	General Update and modified Address Mapping Scheme
<i>Rev 1 – December 6, 2002 by Maikel Kokaly-Bannourah and Robert Hoffmann</i>	Initial Release